Intelligent Power Module (IPM) 600 V, 15 A

NFAQ1560R43T

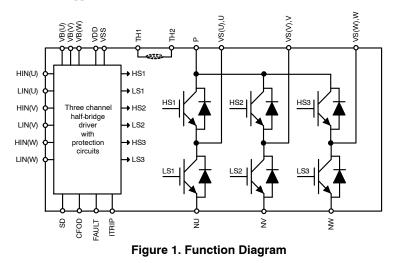
The NFAQ1560R43T is a fully-integrated inverter power stage consisting of a high-voltage driver, six IGBTs (FS4 RC IGBT technology) and a thermistor, suitable for driving permanent magnet synchronous motors (PMSM), brushless-DC (BLDC) motors and AC asynchronous motors. The IGBTs are configured in a 3-phase bridge with separate emitter connections for the lower legs for maximum flexibility in the choice of control algorithm. The power stage has a full range of protection functions including cross-conduction protection, external shutdown and under-voltage lockout functions. An internal comparator and reference connected to the over-current protection circuit allows the designer to set the over-current protection level.

Features

- Three-phase 15 A/600 V IGBT Module with Integrated Drivers
- Compact 29.6 mm x 18.2 mm Dual In-Line Package
- Built-in Under Voltage Protection
- Cross-conduction Protection
- ITRIP Input to Shut Down All IGBTs
- Integrated Bootstrap Diodes and Resistors
- Thermistor for Substrate Temperature Measurement
- Shut Down Pin
- UL1557 Certification (File Number: E339285)
- This is a Pb–Free Device

Typical Applications

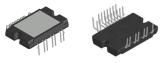
- Industrial Pumps
- Industrial Fans
- Industrial Automation
- Home Appliances





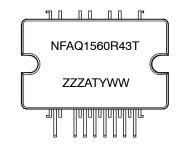
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DIP38 29.6x18.2 CASE 181AE

MARKING DIAGRAM



NFAQ1560R43T	= Specific Device Code
ZZZ	= Assembly Lot Code
А	= Assembly Location
Т	= Test Location
Y	= Year
WW	= Work Week

Device marking is on package top side

ORDERING INFORMATION

Device	Package	Shipping (Qty / Packing)
NFAQ1560R43T	DIP38 (Pb-Free)	400 / Box

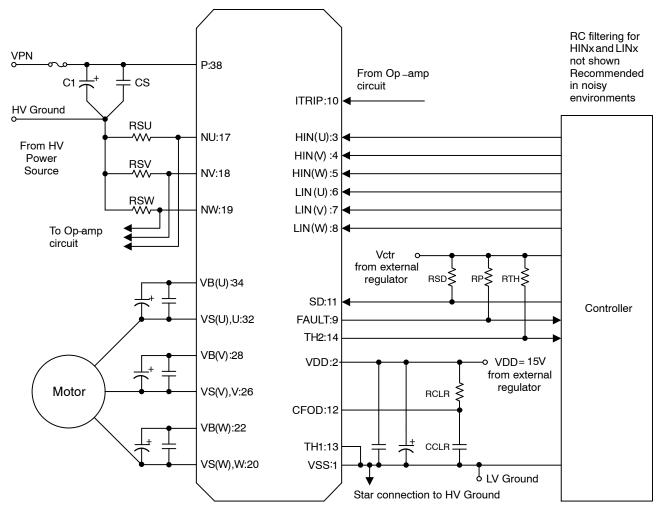


Figure 2. Application Schematic

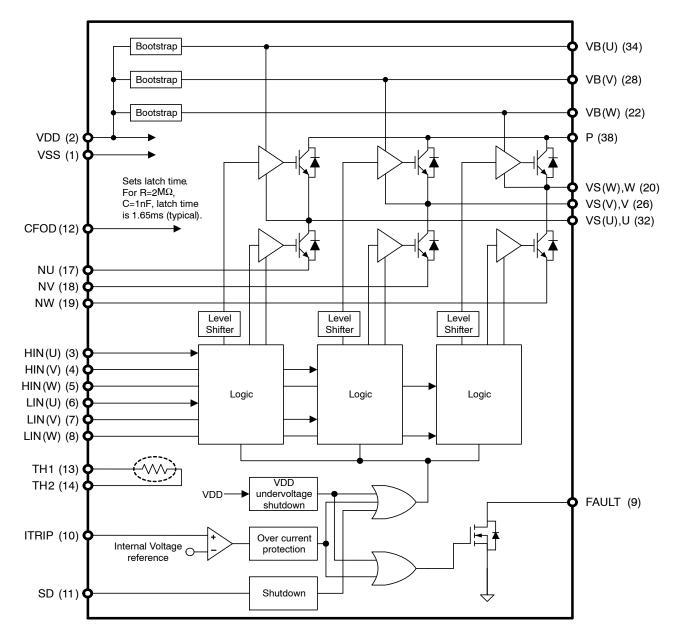


Figure 3. Simplified Block Diagram

PIN FUNCTION DESCRIPTION

Pin	Name	Description
1	VSS	Low-Side Common Supply Ground
2	VDD	Low-Side Bias Voltage for IC and IGBTs Driving
3	HIN(U)	Signal Input for High-Side U Phase
4	HIN(V)	Signal Input for High-Side V Phase
5	HIN(W)	Signal Input for High-Side W Phase
6	LIN(U)	Signal Input for Low-Side U Phase
7	LIN(V)	Signal Input for Low-Side V Phase
8	LIN(W)	Signal Input for Low-Side W Phase
9	FAULT	Fault output
10	ITRIP	Input for Over Current Protection
11	SD	Shut Down Input
12	CFOD	Capacitor and Resistor for Fault Output Duration Selection
13	TH1	Thermistor Bias Voltage
14	TH2	Series Resistor for Thermistor
17	NU	Negative DC-Link Input for U Phase
18	NV	Negative DC-Link Input for V Phase
19	NW	Negative DC-Link Input for W Phase
20	VS(W), W	High-Side Bias Voltage GND for W phase IGBT Driving, Output for W Phase
22	VB(W)	High-Side Bias Voltage for W phase IGBT Driving
26	VS(V), V	High-Side Bias Voltage GND for V phase IGBT Driving, Output for V Phase
28	VB(V)	High-Side Bias Voltage for V phase IGBT Driving
32	VS(U), U	High-Side Bias Voltage GND for U phase IGBT Driving, Output for U Phase
34	VB(U)	High-Side Bias Voltage for U phase IGBT Driving
38	Р	Positive DC-Link Input

 $NOTE: \quad \text{Pins 15, 16, 21, 23, 24, 25, 27, 29, 30, 31, 33, 35, 36 and 37 are not present}$

ABSOLUTE MAXIMUM RATINGS at T_C = 25°C (Note 1)

Parameter	Symbol	Conditions	Rating	Unit	
Supply Voltage	V _{PN}	P – NU, NV, NW, VPN (surge) < 500 V (Note 2)	450	V	
Collector-Emitter Voltage	V _{CES}	P – U, V, W; U – NU; V – NV; W – NW	600	V	
Each IGBT Collector Current	Ι _C	P, U, V, W, NU, NV, NW terminal current	±15	Α	
		P, U, V, W, NU, NV, NW terminal current, T _C = 100°C	±7.5	A	
Each IGBT Collector Current (Peak)	ch IGBT Collector Current (Peak) ICp T _C = 25°C, Under 1 ms Pulse Width		±30	Α	
Corrector Dissipation	P _C	T _C = 25°C, Per One Chip (IGBT Part)	50	W	
		T _C = 25°C, Per One Chip (FRD Part)	26.5	W	
High-Side Control Bias voltage	V _{BS}	VB(U) – VS(U), VB(V) – VS(V), VB(W) – VS(W) (Note 3)	-0.3 to +20.0	V	
Control Supply Voltage	V _{DD}	VDD – VSS	-0.3 to +20.0	V	
Input Signal Voltage	V _{IN}	HIN(U), HIN(V), HIN(W), LIN(U), LIN(V), LIN(W) – VSS	–0.3 to V _{DD}	V	
FAULT Terminal Voltage	V _{FAULT}	FAULT – VSS	-0.3 to V _{DD}	V	
CFOD Terminal Voltage	V _{CFOD}	CFOD – VSS	–0.3 to V _{DD}	V	
SD Terminal Voltage	V _{SD}	SD – VSS	–0.3 to V _{DD}	V	
Current Sensing Input Voltage	V _{ITRIP}	ITRIP – VSS	-0.3 to +10.0	V	
Operating Junction Temperature	Тj		150	°C	
Storage Temperature	T _{stg}		-40 to +125	°C	
Module Case Operation Temperature	T _c		-40 to +125	°C	
Tightening Torque	MT	Case mounting screws	0.6	Nm	
Isolation Voltage	V _{iso}	50 Hz sine wave AC 1 minute (Note 4)	2000	Vrms	

Stresses exceeding those listed in the Maximum Ratings table may damage the device. If any of these limits are exceeded, device functionality should not be assumed, damage may occur and reliability may be affected. 1. Refer to ELECTRICAL CHARACTERISTICS, RECOMMENDED OPERATING RANGES and/or APPLICATION INFORMATION for Safe

Operating parameters

2. This surge voltage developed by the switching operation due to the wiring inductance between P and NU, NV, NW terminal.

3. VBS = VB(U) - VS(U), VB(V) - VS(V), VB(W) - VS(W)

4. Test conditions : AC2500V, 1 s

RECOMMENDED OPERATING RANGES

Rating	Symbol	Conditions	Min	Тур	Max	Unit
Supply Voltage	V _{PN}	P – NU, NV, NW	0	280	450	V
High–Side Control Bias Voltage	V _{BS}	VB(U) – VS(U), VB(V) – VS(V), VB(W) – VS(W)	13.0	15	17.5	V
Control Supply Voltage	V _{DD}	VDD – VSS	14.0	15	16.5	V
ON-state Input Voltage	V _{IN(ON)}	HIN(U), HIN(V), HIN(W), LIN(U), LIN(V),	3.0	-	5.0	V
OFF-state Input Voltage	V _{IN(OFF)}	LIN(W) – VSS	0	-	0.3	V
PWM Frequency	f _{PWM}		1	-	20	kHz
Dead Time	DT	Turn-off to Turn-on (external)	1	-	-	μs
Allowable Input Pulse Width	PWIN	ON and OFF	1	-	-	μs
Tightening Torque		'M3' Type Screw	0.4	-	0.6	Nm

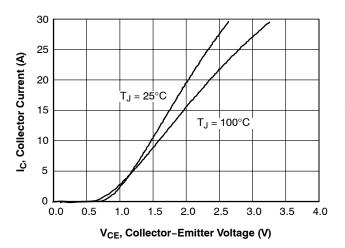
Functional operation above the stresses listed in the Recommended Operating Ranges is not implied. Extended exposure to stresses beyond the Recommended Operating Ranges limits may affect device reliability.

ELECTRICAL CHARACTERISTICS at T_C = 25 °C, V_{BIAS} (V_{BS} , V_{DD}) = 15 V unless otherwise noted.

Parameter	Test Conditions	Symbol	Min	Тур	Max	Unit
POWER OUTPUT SECTION						
Collector-Emitter Leakage Current	V _{CE} = 600 V	I _{CES}	-	-	100	μΑ
Collector-Emitter Saturation Voltage	$IN = 5 V, I_C = 15 A, T_j = 25 °C$	V _{CE(sat)}	-	1.8	2.5	V
	IN = 5 V, I _C = 7.5 A, T _j = 100 °C		-	1.4	-	V
FWDi Forward Voltage	$IN = 0 V, I_C = -15 A, T_i = 25 \degree C$	VF	-	2.0	2.5	V
	IN = 0 V, I _C = -7.5 A, T _j = 100 °C		-	1.6	-	V
Junction to Case Thermal Resistance	Inverter IGBT Part (per 1/6 Module)	R _{th(j-c)Q}	-	-	2.5	°C/W
	Inverter FRD Part (per 1/6 Module)	R _{th(j-c)F}	-	-	4.7	°C/W
SWITCHING CHARACTER					1	
Switching Time	I _C = 15 A, V _{PN} = 300 V, T _j = 25°C,	t _{ON}	-	0.4	1.1	μs
	Inductive Switching	t _{OFF}	-	0.5	1.2	μs
Turn-on Switching Loss	I _C = 15 A, V _{PN} = 300 V, T _j = 25°C	E _{ON}	-	360	-	μJ
Turn-off Switching Loss	-	E _{OFF}	-	260	-	μJ
Total Switching Loss	-	E _{TOT}	_	620	_	μJ
Turn-on Switching Loss	I _C = 7.5 A, V _{PN} = 300 V, T _i = 100°C	E _{ON}	_	180	_	Lμ
Turn-off Switching Loss		E _{OFF}	_	180	_	μJ
Total Switching Loss	-	E _{TOT}	_	360	_	μJ
Diode Reverse Recovery Energy	I _C = 7.5 A, V _{PN} = 300 V, T _i = 100°C,	E _{REC}	_	110	_	μJ
Diode Reverse Recovery Time	(di/dt set by internal driver)	t _{RR}	_	120	-	ns
Reverse Bias Safe Operating Area	I _C = 30 A, V _{CE} = 450 V	RBSOA		Full Squar	e	
Short Circuit Safe Operating Area	uit Safe Operating Area $V_{CE} = 400 \text{ V}, \text{ T}_i = 100^{\circ}\text{C}$		4		-	μs
DRIVER SECTION						· · ·
Quiescent VBS Supply Current	V _{BS} = 15 V, HIN = 0 V, per driver	I _{QBS}	-	0.08	0.4	mA
Quiescent VDD Supply Current	V _{DD} = 15 V, LIN = 0 V, V _{DD} – V _{SS}	I _{QDD}	_	0.95	3.0	mA
ON Threshold Voltage	HIN(U), HIN(V), HIN(W), LIN(U), LIN(V),	VIN(ON)	-	-	2.5	V
OFF Threshold Voltage	LIN(W) – VSS	V _{IN(OFF)}	0.8	-	-	V
Logic 1 Input Current	V _{IN} = +3.3 V	I _{IN+}	_	660	900	μΑ
Logic 0 Input Current	V _{IN} = 0 V	I _{IN-}	-	-	3	μΑ
Bootstrap ON Resistance	I _B = 1 mA	RB	-	500	-	Ω
FAULT Terminal Sink Current	FAULT: ON / VFAULT = 0.1 V	loSD	-	2	-	mA
Fault-Output Pulse Width	FAULT – VSS From time fault condition clear R = 2 M Ω , C = 1 nF	t _{FOD}	1.1	1.65	2.2	ms
CFOD Threshold	CFOD - VSS	V _{CFOD}	-	8	-	V
Shut Down Threshold	SD – VSS	V _{SD+}	-	-	2.5	V
		V _{SD-}	0.8	-	-	V
ITRIP Trip Level	ITRIP – VSS	VITRIP	0.44	0.49	0.54	V
ITRIP to Shutdown Propagation Delay		t _{ITRIP}	-	1.1	-	μs
ITRIP Blanking Time		t _{ITRIPBL}	250	350	-	ns
High-Side Control Bias Voltage	Reset Level	UVBSR	10.3	11.1	11.9	V
Under-Voltage Protection	Detection Level	UVBSD	10.1	10.9	11.7	V
	Hysteresis	UVBSH	_	0.2	_	V

Parameter	Test Conditions	Symbol	Min	Тур	Max	Unit
DRIVER SECTION						
Supply Voltage Under-Voltage Protection	Reset Level	UVDDR	10.3	11.1	11.7	V
	Detection Level	UVDDD	10.1	10.9	11.5	V
	Hysteresis	UVDDH	-	0.2	-	V

Product parametric performance is indicated in the Electrical Characteristics for the listed test conditions, unless otherwise noted. Product performance may not be indicated by the Electrical Characteristics if operated under different conditions.



TYPICAL CHARACTERISTICS

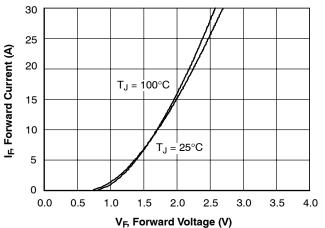


Figure 4. V_{CE} versus I_C for Different Temperatures (V_{DD} = 15 V)

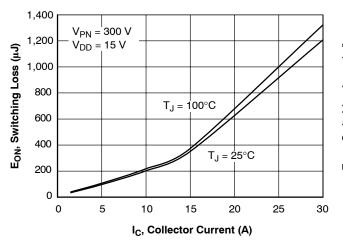


Figure 6. E_{ON} versus I_C for Different Temperatures

Figure 5. V_F versus I_F for Different Temperatures

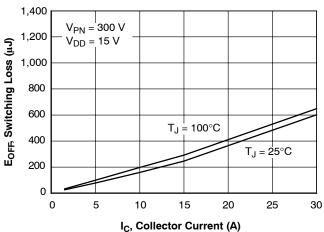


Figure 7. E_{OFF} versus I_{C} for Different Temperatures

TYPICAL CHARACTERISTICS (Continued)

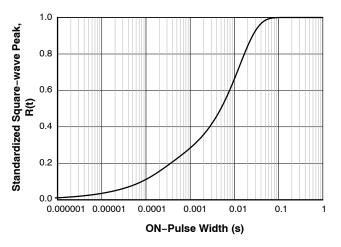


Figure 8. Thermal Impedance Plot (IGBT)

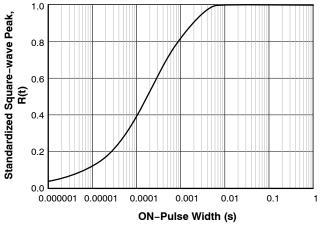


Figure 9. Thermal Impedance Plot (FRD)

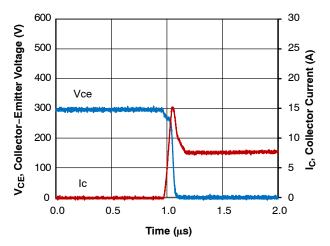


Figure 10. Turn–on Waveform Tj = 100°C, V_{CC} = 300 V

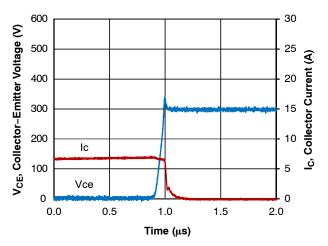
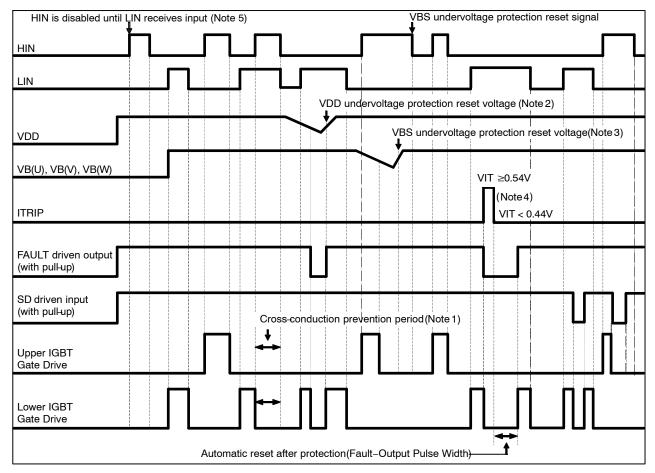


Figure 11. Turn–off Waveform Tj = 100°C, V_{CC} = 300 V

APPLICATIONS INFORMATION

Input / Output Timing Chart



NOTES:

- 1. This section of the timing diagram shows the effect of cross-conduction prevention.
- 2. This section of the timing diagram shows that when the voltage on VDD decreases sufficiently all gate output signals will go low, switching off all six IGBTs. When the voltage on VDD rises sufficiently, normal operation will resume.
- 3. This section shows that when the bootstrap voltage on VB(U) (VB(V), VB(W)) drops, the corresponding high side output U (V, W) is switched off. When the voltage on VB(U) (VB(V), VB(W)) rises sufficiently, normal operation will resume.
- This section shows that when the voltage on ITRIP exceeds the threshold, all IGBTs are turned off. Normal operation resumes later after the over-current condition is removed.
- 5. After VDD has risen above the threshold to enable normal operation, the driver waits to receive an input signal on the LIN input before enabling the driver for the HIN signal.

Figure 12. Input / Output Timing Chart

INPUT				OUTPUT			
HIN	LIN	ITRIP	SD	High side IGBT	Low side IGBT	U,V,W	FAULT
Н	L	L	Н	ON (Note 5)	OFF	Р	OFF
L	Н	L	Н	OFF	ON	NU,NV,NW	OFF
L	L	L	Н	OFF	OFF	High Impedance	OFF
Н	Н	L	Н	OFF	OFF	High Impedance	OFF
Х	Х	Н	Х	OFF	OFF	High Impedance	ON
Х	Х	L	L	OFF	OFF	High Impedance	OFF

Table 1. INPUT / OUTPUT LOGIC TABLE

Parameter	Symbol	Condition	Min	Тур	Max	Unit
Resistance	R ₂₅	$T_{th} = 25^{\circ}C$	99	100	101	kΩ
	R ₁₀₀	$T_{th} = 100^{\circ}C$	5.2	5.4	5.6	kΩ
B-Constant (25 to 50°C)	В		4208	4250	4293	К
Temperature Range			-40	-	+125	°C

Table 2. THERMISTOR CHARACTERISTICS

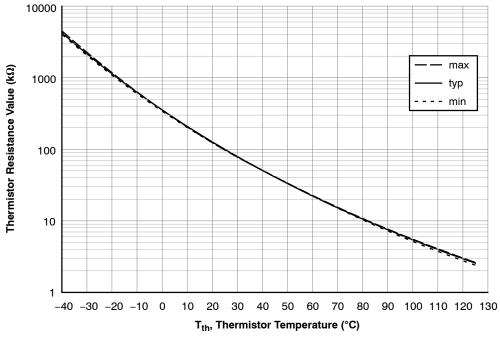
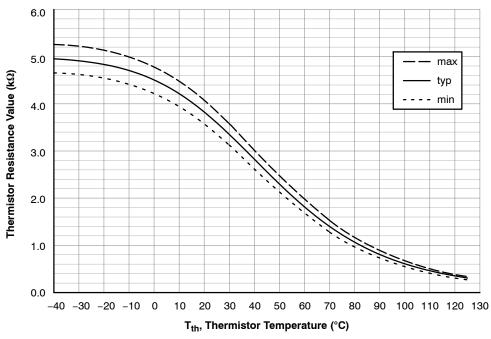
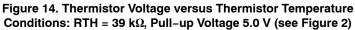


Figure 13. Thermistor Resistance versus Thermistor Temperature





FAULT Pin

The FAULT output is an open drain output requiring a pull-up resistor. If the pull-up voltage is 5 V, use a pull-up resistor with a value of 6.8 k Ω or higher. If the pull-up voltage is 15 V, use a pull-up resistor with a value of 20 k Ω or higher. The FAULT output is triggered if there is a VDD undervoltage or an overcurrent condition.

Under-voltage Protection

If VDD goes below the VDD supply under-voltage lockout falling threshold, the FAULT output is switched on. The FAULT output stays on until VDD rises above the VDD supply under-voltage lockout rising threshold. After VDD has risen above the threshold to enable normal operation, the driver waits to receive an input signal on the LIN input before enabling the driver for the HIN signal.

Overcurrent Protection

An over-current condition is detected if the voltage on the ITRIP pin is larger than the reference voltage. There is a blanking time of typically 350 ns to improve noise immunity. After a shutdown propagation delay of typically 1.1 μ s, the FAULT output is switched on. The FAULT output is held on for a time determined by the resistor and capacitor connected to the CFOD pin. If RCLR = 2 M Ω and CCLR = 1 nF, the FAULT output is switched on for 1.65 ms (typ.) because the FAULT pin goes back to high impedance when CFOD is higher than 8 V (typ.).

The over-current protection threshold should be set to be equal or lower to 2 times the module rated current (Io).

An additional fuse is recommended to protect against system level or abnormal over-current fault conditions.

Capacitors on High Voltage and VDD Supplies

Both the high voltage and VDD supplies require an electrolytic capacitor and an additional high frequency capacitor. The recommended value of the high frequency capacitor is between 100 nF and 10 μ F.

SD Pin

The SD terminal pin is used to enable or shut down the built-in driver. If the voltage on the SD pin rises above the VSD+ voltage, the output drivers are enabled. If the voltage on the SD pin falls below the VSD- voltage, the drivers are disabled.

Minimum Input Pulse Width

When input pulse width is less than 1 μ s, an output may not react to the pulse. (Both ON signal and OFF signal)

Calculation of Bootstrap Capacitor Value

The bootstrap capacitor value CB is calculated using the following approach. The following parameters influence the choice of bootstrap capacitor:

- VBS: Bootstrap power supply. 15 V is recommended.
- QG: Total gate charge of IGBT at VBS = 15 V. 8 nC.
- UVLO: Falling threshold for UVLO. Specified as 12 V.
- IDMAX: High side drive power dissipation. Specified as 0.4 mA.
- TONMAX: Maximum ON pulse width of high side IGBT.

Capacitance calculation formula:

```
CB = (QG + IDMAX * TONMAX)/(VBS – UVLO)
```

CB is recommended to be approximately 3 times the value calculated above. The recommended value of CB is in the range of 1 to 47 μ F, however, the value needs to be verified prior to production. When not using the bootstrap circuit, each high side driver power supply requires an external independent power supply.

The internal bootstrap circuit uses a MOSFET. The turn on time of this MOSFET is synchronized with the turn on of the low side IGBT. The bootstrap capacitor is charged by turning on the low side IGBT.

If the low side IGBT is held on for a long period of time (more than one second for example), the bootstrap voltage on the high side MOSFET will slowly discharge.

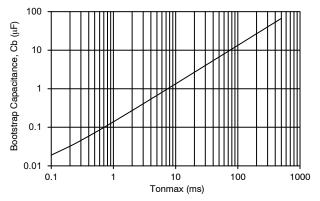


Figure 15. Bootstrap Capacitance versus Tonmax

TEST CIRCUITS

• ICES

	U+	V+	W+	U–	V–	W–
А	38	38	38	32	26	20
В	32	26	20	17	18	19

U+, V+, W+ : High side phase U-, V-, W- : Low side phase

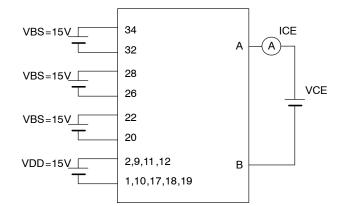


Figure 16. Test Circuit for I_{CE}

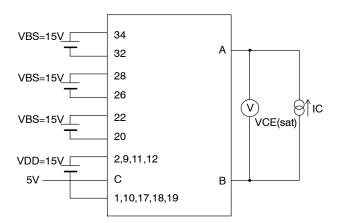


Figure 17. Test Circuit for V_{CE(SAT)}

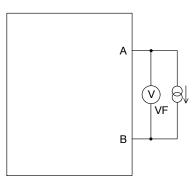


Figure 18. Test Circuit for V_F

• VCE(sat) (Test by pulse)

• VF (Test by pulse)

А

В

U+

38

32

V+

38

26

	U+	V+	W+	U–	V–	W–
А	38	38	38	32	26	20
В	32	26	20	17	18	19
С	3	4	5	6	7	8

W+

38

20

U–

32

17

V–

26

18

W-

20

19

• RB (Test by pulse)

	U+	V+	W+
А	2	2	2
В	34	28	22
С	6	7	8

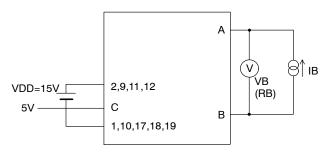
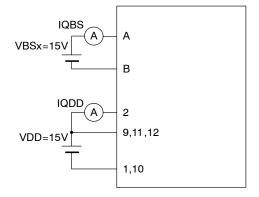


Figure 19. Test Circuit for RB

• IQBS, IQDD

	VBS U+	VBS V+	VBS W+	V _{DD}
А	34	28	22	2
В	32	26	20	1





• Switching Time (The circuit is a representative example of the Inverter Low side U phase.)

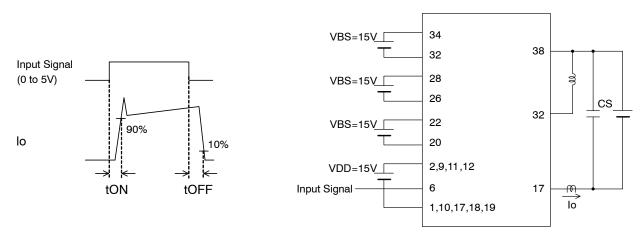
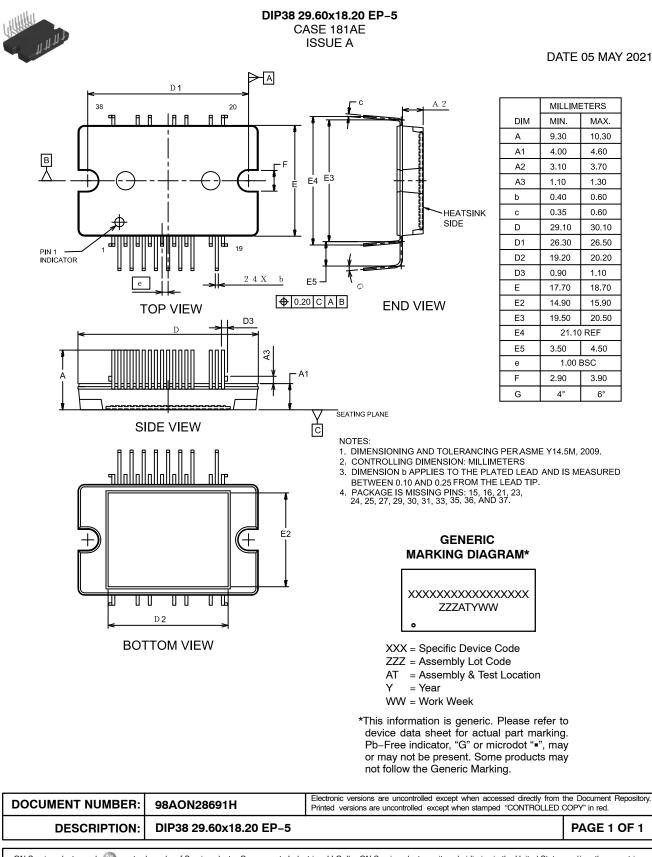


Figure 21. Test Circuit for Switching Time





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