

TC1130

32-Bit Single-Chip Microcontroller

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Microcontrollers



Never stop thinking.

Edition 2008-12

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TC1130

32-Bit Single-Chip Microcontroller

Microcontrollers



TC1130 Da	ta Sheet		
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77, 78, 79 ,81, 82	Added "O	perating Conditons apply" statement	

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32-Bit Single-Chip Microcontroller TriCore™ Family

TC1130

1 Summary of Features

- High Performance 32-bit TriCore[™] V1.3 CPU with 4-Stage Pipeline
- Floating Point Unit (FPU)
- Dual Issue super-scalar implementation
 - MAC Instruction maximum triple issue
- Circular Buffer and bit-reverse addressing modes for DSP algorithms
- Very fast interrupt response time
- Hardware controlled context switch for task switch and interrupts
- Memory Management Unit (MMU)
- On-chip Memory
 - 28-Kbyte Data Memory (SPRAM)
 - 32-Kbyte Code Memory (SPRAM)
 - 16-Kbyte Instruction Cache (ICACHE)
 - 4-Kbyte Data Cache (DCACHE)
 - 64-Kbyte SRAM Data Memory Unit (DMU)
 - 16-Kbyte Boot ROM
- On-chip Bus Systems
 - 64-bit High Performance Local Memory Bus (LMB) for fast access between caches and on-local memories and FPI Interface
 - On-chip Flexible Peripheral Interconnect Bus (FPI) for interconnections of functional units
- DMA Controller with 8 channels for data transfer operations between peripheral units and memory locations
- Two high speed Micro Link Interfaces (MLI0/1) for controller communication and emulation
- Flexible External Bus Interface Unit (EBU) to access external data memories
- One Multifunctional General Purpose Timer Unit (GPTU) with three 32-bit timer/ counters
- Two Capture and Compare units (CCU60/1) for PWM signal generation, each with
 - 3-channel, 16 bit Capture and Compare unit
 - 1-channel, 16 bit Compare unit
- Three Asynchronous/Synchronous Serial Channels (ASC0/1/2) with baud-rate generator, parity, framing and overrun error detection, support FIFO and IrDA data transmission
- Two High Speed Synchronous Serial Channels (SSC0/1) with programmable data length, FIFO support and shift direction

Data Sheet 1 V1.1, 2008-12



Summary of Features

- One MultiCAN module with four CAN nodes and 128 message buffers for high efficiency data handling
- Fast Ethernet Controller with 10/100 Mbit/sec MII-Based physical devices support
- USB module with compliance to USB Specification Revision 1.1, with support for 1.5 MBaud to 12 MBaud devices
- Inter-IC (IIC) module with two physical IIC buses
- Digital I/O ports with 3.3 V I/O capabilities
- Level 2 On-chip Debug Support
- Power Management System
- Clock Generation Unit with PLL
- Maximum CPU and Bus clock frequency at 150 MHz without MMU and 120 MHz with MMU
- Ambient temperature under bias: -40° to +85°C
- P-LBGA-208 package



2 General Device Information

2.1 Block Diagram

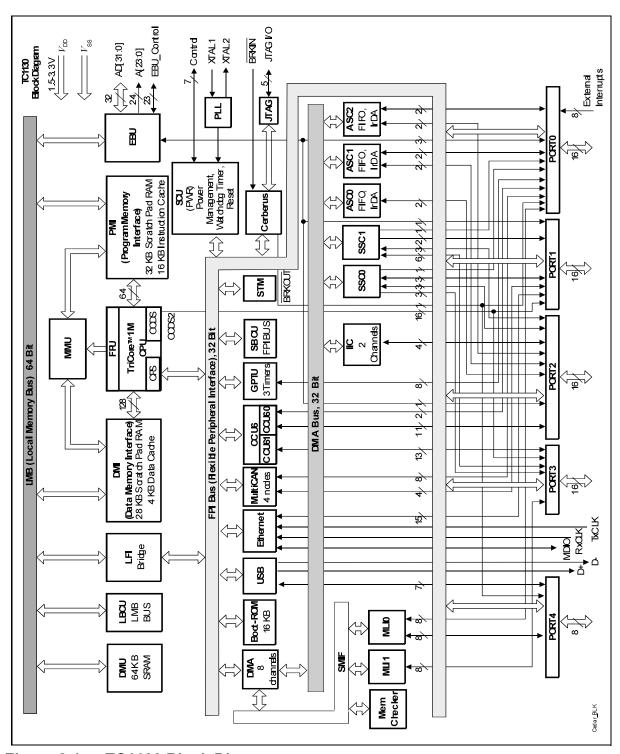


Figure 2-1 TC1130 Block Diagram



2.2 Logic Symbol

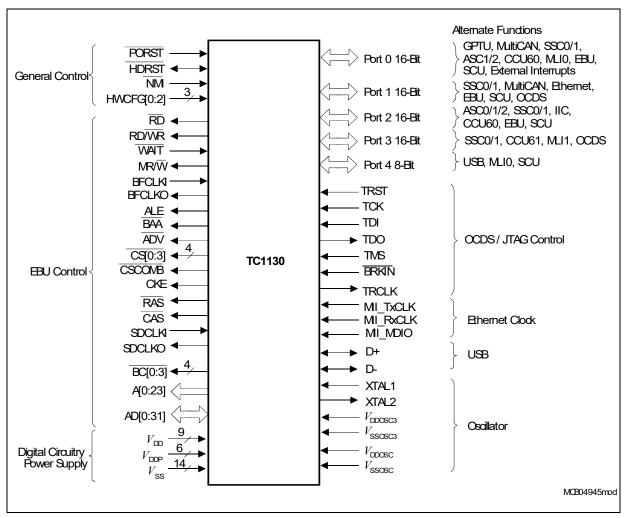


Figure 2-2 TC1130 Logic Symbol



2.3 Pin Configuration

	A	В	С	D		F	G	Н	J	K	L	M	N	P	R	т	
16	Reser ved	P3.10	P3.11	P3.12	P2.15	P2.14	P2.11	P2.9	P2.8	P2.7	V _{DDOS0}	XTAL1	XTAL2	V _{DD}	$V_{ m SS}$	Reser ved	16
15	P3.0	P3.1	P3.8	P3.2	P3.3	P3.6	P3.5	P3.9	P3.15	P2.12	$V_{ m SS}$	P0.3	P2.4	P0.1	P0.9	D-	15
14	P1.9	P1.10	P1.11	P1.14	P1.13	P1.15	P3.4	P3.7	P3.14	P2.13	HW ŒG1	HW CFG0	P2.5	P2.3	P0.10	D+	14
13	P1.8	P1.7	P1.5	V_{DDP}	$V_{\rm SS}$	P1.12	V_{DD}	V_{SS}	V_{DDP}	P3.13	P2.10	V_{SS}	V_{DDP}	P2.2	P0.8	TDI	13
12	P1.6	P1.3	P1.1	P1.2							ge Pi	n	P2.6	P2.0	P0.5	так	12
11	BAA	ADV	P1.4	P1.0		С	onfig f		ion (t C113		iew)		P0.0	P2.1	P0.4	TRST	11
10	A17	A18	A19	A20			V_{DD}	Vss	V_{SS}	V_{DD}			P0.7	P0.2	P0.6	ТΣ	10
9	A16	WAIT	CS2	<u>C</u> 80			V_{DD}	V _{SS}	V_{SS}	V_{DD}			P0.11	P0.12	P4.1	TMS	9
8	A15	css	AD0	CS1			V_{DD}	Vss	Vss	V_{DD}			P0.14	P0.13	P4.0	TRCLK	8
7	BC3	BC2	AD1	AD16			V_{DD}	Vss	V_{SS}	V_{DD}			P4.2	P0.15	P4.5	NMI	7
6	BC1	AD2	AD3	RAS		,					-		P4.3	P4.4	P4.6	HW CFG2	6
5	BC0	AD17	AD4	CAS									HDRS	P4.7	PORST	BRKIN	5
4	AD18	AD 19	AD20	V_{DDP}	V_{SS}	AD28	AD29	V_{DDP}	V_{SS}	A14	Œ	V_{DDP}	$V_{\rm SS}$	A23	A22	A21	4
3	AD5	AD21	AD7	AD25	AD11	AD12	AD15	AD30	A10	A11	A12	A13	<u>CS</u> COMB	MR/W	ALE	RD W F	3
2	AD6	AD22	AD8	AD9	AD26	AD27	AD31	AD14	A 5	A6	A7	A8	A9	RD	MII_ RXCLK	MII_ TXCLK	2
1	Reser ved	AD23	AD24		BFCLKO	AD10	AD13	acctiko	SDOLKI	AO	A1	A2	A3	A4	MII_ MDIO	Reser ved	1
	Α	В	С	D	E	F	G	Н	J	K	L	М	N	Р	R	Т	
														MCI	- 0495	0mod	

Figure 2-3 TC1130 Pins: P-BGA-208 Package (top view)



2.4 Pin Definitions and Functions

Table 2-1 Pin Definitions and Functions

Symbol	Pin	In Out	PU/ PD ¹⁾	Functions	
P0		I/O		Port 0	
				Port 0 is a 16-bi	it bi-directional general purpose I/O port
				which can be a	lternatively used for GPTU, MultiCAN,
				ASC1/2, SSC0	/1, MLI0, EBU and SCU.
P0.0	N11	I/O	PUC	GPTU_0	GPTU input/output line 0
		I/O		RXD1B	ASC1 receiver input/output B
P0.1	P15	I/O	PUC	GPTU_1	GPTU input/output line 1
		0		TXD1B	ASC1 transmitter output B
P0.2	P10	I/O	PUC	GPTU_2	GPTU input/output line 2
		I/O		RXD2B	ASC2 receiver input/output B
P0.3	M15	I/O	PUC	GPTU_3	GPTU input/output line 3
		0		TXD2B	ASC2 transmitter output B
P0.4	R11	I/O	PUC	GPTU_4	GPTU input/output line 4
				SLSI1	SSC1 Slave Select input
		0		BREQ	EBU Bus Request Output
P0.5	R12	I/O	PUC	GPTU_5	GPTU input/output line 5
				HOLD	EBU Hold Request Input
				l 	CCU60 Timer 12 hardware run
D 0.0	D.40	0	D. 10	BRKOUT_B	OCDS Break Out B
P0.6	R10	I/O	PUC	GPTU_6	GPTU input/output line 6
		I/O		HLDA	EBU Hold Acknowledge Input/Output
				_	CCU60 Timer 13 hardware run
D0.7	NIAO	0	DUG	SLSO0_0	SSC0 Slave Select output 0
P0.7	N10	I/O	PUC	GPTU_7	GPTU input/output line 7
D0 0	D40	0	DUG	SLSO1_0	SSC1 Slave Select output 0
P0.8	R13		PUC		
		0		REQ0	External Trigger Input 0 MLI0 transmit channel clock output A
P0.9	R15		PUC	TCLK0A	•
P0.9	KIS	0	PUC	TREADY0A	CAN node 0 transmitter output A
				REQ1	MLI0 transmit channel ready input A
P0.10	R14		PUC		External Trigger Input 1 CAN node 1 receiver input A
F 0. 10	1114		- 00	REQ2	External Trigger Input 2
		0		TVALID0A	MLI0 transmit channel valid output A
P0.11	N9	0	PUC	TXDCAN1_A	CAN node 1 transmitter output A
1 0.11	פוו		00	REQ3	External Trigger Input 3
		0		TDATA0A	MLI0 transmit channel data output A
				IDAIAUA	METO transmit oriannel data output A



Table 2-1 Pin Definitions and Functions (cont'd)

Symbol	Pin	In Out	PU/ PD ¹⁾	Functions	
P0.12	P9	I	PUC	RXDCAN2	CAN node 2 receiver input
		1		RCLK0A	MLI0 receive channel clock input A
		1		REQ4	External Trigger Input 4
P0.13	P8	0	PUC	TXDCAN2	CAN node 2 transmitter output
		1		REQ5	External Trigger Input 5
		0		RREADY0A	MLI0 receive channel ready output A
P0.14	N8	1	PUC	RXDCAN3	CAN node 3 receiver input
		1		REQ6	External Trigger Input 6
		1		RVALID0A	MLI0 receive channel valid input A
P0.15	P7	0	PUC	TXDCAN3	CAN node 3 transmitter output
		1		REQ7	External Trigger Input 7
		I		RDATA0A	MLI0 receive channel data input A



Table 2-1 Pin Definitions and Functions (cont'd)

Symbol	Pin	In Out	PU/ PD ¹⁾	Functions	
P1		I/O		Port 1	
				Port 1 serves	as 16-bit bi-directional general purpose
				I/O port which	can be used for input/output for Ethernet
				controller, Mul	tiCAN, CAN, OCDS L2, SSC0/1, EBU
				and SCU.	
P1.0	D11	0	PUC	MII_TXD0	Ethernet controller transmit data
					output line 0
		1		RXDCAN0_B	CAN node 0 receiver input B
		1		SWCFG0	Software configuration 0
		0		OCDSA_0	OCDS L2 Debug Line A0
P1.1	C12	0	PUC	MII_TXD1	Ethernet controller transmit data output line 1
		1		SWCFG1	Software configuration 1
		0		TXDCAN0_B	CAN node 0 transmitter output B
		0		OCDSA_1	OCDS L2 Debug Line A1
P1.2	D12	0	PUC	MII_TXD2	Ethernet controller transmit data output line 2
		1		RXDCAN1_B	CAN node 1 receiver input B
		1		SWCFG2	Software configuration 2
		0		OCDSA_2	OCDS L2 Debug Line A2
P1.3	B12	0	PUC	MII_TXD3	Ethernet controller transmit data output line 3
		0		TXDCAN1 B	CAN node 1 transmitter output B
		1		SWCFG3	Software configuration 3
		0		OCDSA_3	OCDS L2 Debug Line A3
P1.4	C11	0	PUC	MII TXER	Ethernet controller transmit error
				_	output line
		1		SWCFG4	Software configuration 4
		0		OCDSA_4	OCDS L2 Debug Line A4
P1.5	C13	0	PUC	MII_TXEN	Ethernet controller transmit enable output line
		1		SWCFG5	Software configuration 5
		0		OCDSA 5	OCDS L2 Debug Line A5
P1.6	A12	0	PUC	MII_MDC	Ethernet controller management data clock output line
		1		SWCFG6	Software configuration 6
		0		OCDSA_6	OCDS L2 Debug Line A6



Table 2-1 Pin Definitions and Functions (cont'd)

Symbol	Pin	In Out	PU/ PD ¹⁾	Functions	
P1.7	B13	I	PUC	MII_RXDV	Ethernet Controller receive data valid input line
		1		SWCFG7	Software configuration 7
		0		OCDSA_7	OCDS L2 Debug Line A7
P1.8	A13	1	PUC	MII_CRS	Ethernet Controller carrier input line
		1		SWCFG8	Software configuration 8
		0		OCDSA_8	OCDS L2 Debug Line A8
P1.9	A14	I	PUC	MII_COL	Ethernet Controller collision input line
		1		SWCFG9	Software configuration 9
		0		OCDSA_9	OCDS L2 Debug Line A9
P1.10	B14	I	PUC	MII_RXD0	Ethernet Controller receive data input line 0
		1		SWCFG10	Software configuration 10
		0		OCDSA_10	OCDS L2 Debug Line A10
P1.11	C14	I	PUC	MII_RXD1	Ethernet Controller receive data input line 1
		1		SWCFG11	Software configuration 11
		0		OCDSA_11	OCDS L2 Debug Line A1
		0		SLSO0_1	SSC0 Slave Select output 1
P1.12	F13	I	PUC	MII_RXD2	Ethernet Controller receive data input line 2
		1		SWCFG12	Software configuration 12
		0		OCDSA_12	OCDS L2 Debug Line A12
		0		SLSO1_1	SSC1 Slave Select output 1
P1.13	E14	I	PUC	MII_RXD3	Ethernet Controller receive data input line 3
		1		SWCFG13	Software configuration 13
		0		OCDSA_13	OCDS L2 Debug Line A13
		0		SLSO0_2	SSC0 Slave Select output 2
P1.14	D14	I	PUC	MII_RXER	Ethernet Controller receive error input line
		0		SLSO1 2	SSC1 Slave Select output 2
		1		SWCFG14	Software configuration 14
		0		OCDSA 14	OCDS L2 Debug Line A14
P1.15	F14	I	PUC	SLSI0	SSC0 Slave Select Input
		0		RMW	EBU Read Modify Write
		1		SWCFG15	Software configuration 15
		0		OCDSA_15	OCDS L2 Debug Line A15



Table 2-1 Pin Definitions and Functions (cont'd)

Symbol	Pin	In Out	PU/ PD ¹⁾	Functions	
P2		I/O		Port 2	
				Port 2 is a 16-l	bit bi-directional general purpose I/O port
					alternatively used for ASC0/1/2, SSC0/1,
					EBU and SCU.
P2.0	P12	I/O	PUC	RXD0	ASC0 receiver input/output line
		0		CSEMU	EBU Chip Select Output for Emulator
					Region
P2.1	P11	0	PUC	TXD0	ASC0 transmitter output line
		1		TESTMODE	Test Mode Select Input
P2.2	P13	I/O	PUC	MRST0	SSC0 master receive/slave transmit
					input/output
P2.3	P14	I/O	PUC	MTSR0	SSC0 master transmit/slave receive input/output
P2.4	N15	I/O	PUC	SCLK0	SSC0 clock input/output line
P2.5	N14	0	PUC	COUT60_3	CCU60 compare channel 3 output
1 2.0	1,4,1,	1/0	00	MRST1A	SSC1 master receive/slave transmit
		"			input/output A
P2.6	N12	I/O	PUC	CC60_0	CCU60 input/output of capture
-					compare channel 0
		I/O		MTSR1A	SSC1 master transmit/slave receive
					input/output A
P2.7	K16	0	PUC	COUT60 0	CCU60 output of capture/compare
				_	channel 0
		I/O		SCLK1A	SSC1 clock input/output line A
P2.8	J16	I/O	PUC	CC60_1	CCU60 input/output of capture/
					compare channel 1
		I/O		RXD1A	ASC1 receiver input/output line A
P2.9	H16	О	PUC	COUT60_1	CCU60 output of capture/compare channel 1
		0		TXD1A	ASC1 transmitter output line A
P2.10	L13	I/O	PUC	CC60_2	CCU60 input/output of capture/
				_	compare channel 2
		I/O		RXD2A	ASC2 receiver input/output line A
P2.11	G16	0	PUC	COUT60 2	CCU60 output of capture/compare
				_	channel 2
		0		TXD2A	ASC2 transmitter output line A
P2.12	K15	I/O		SDA0	IIC Serial Data line 0
		1		CTRAP0	CCU60 trap input
		0		SLSO0_3	SSC0 Slave Select output 3



Table 2-1 Pin Definitions and Functions (cont'd)

Symbol	Pin	In Out	PU/ PD ¹⁾	Functions	
P2.13	K14	I/O		SCL0	IIC clock line 0
		1		CCPOS0_0	CCU60 Hall input signal 0
		0		SLSO1_3	SSC1 Slave Select output 3
P2.14	F16	1		CCPOS0_1	CCU60 Hall input signal 1
		I/O		SDA1	IIC Serial Data line 1
		0		SLSO0_4	SSC0 Slave Select output 4
P2.15	E16	1		CCPOS0_2	CCU60 Hall input signal 2
		I/O		SCL1	IIC clock line 1
		0		SLSO1_4	SSC1 Slave Select output 4



Table 2-1 Pin Definitions and Functions (cont'd)

Symbol	Pin	In Out	PU/ PD ¹⁾	Functions	
P3		I/O		Port 3	
				Port 3 is a 16-	-bit bi-directional general purpose I/O
				port which car	be alternatively used for MLI1, CCU61,
				SSC0/1 and C	OCDS Level 2 debug lines.
P3.0	A15	0	PUC	OCDSB_0	OCDS L2 Debug Line B0
		0		COUT61_3	CCU61 compare channel 3 output
P3.1	B15	0	PUC	OCDSB_1	OCDS L2 Debug Line B1
		I/O		CC61_0	CCU61 input/output of capture/
					compare channel 0
P3.2	D15	0	PUC	OCDSB_2	OCDS L2 Debug Line B2
		0		COUT61_0	CCU61 output of capture/compare
					channel 0
P3.3	E15	0	PUC	OCDSB_3	OCDS L2 Debug Line B3
		I/O		CC61_1	CCU61 input/output of capture/
			_		compare channel 1
P3.4	G14	0	PUC	OCDSB_4	OCDS L2 Debug Line B4
		0		COUT61_1	CCU61 output of capture/compare
					channel 1
P3.5	G15	0	PUC	OCDSB_5	OCDS L2 Debug Line B5
		I/O		CC61_2	CCU61 input/output of capture/
D0 0	E45		DITO	00000	compare channel 2
P3.6	F15	0	PUC	OCDSB_6	OCDS L2 Debug Line B6
		0		COUT61_2	CCU61 output of capture/compare channel 2
P3.7	H14	0	PUC	OCDSB 7	OCDS L2 Debug Line B7
		I		CTRAP1	CCU61 trap input
		0		SLSO0_5	SSC0 Slave Select output 5
P3.8	C15	0	PUC	OCDSB_8	OCDS L2 Debug Line B8
		I		CCPOS1_0	CCU61 Hall input signal 0
		0		TCLK1	MLI1 transmit channel clock output
		0		SLSO1_5	SSC1 Slave Select output 5
P3.9	H15	0	PUC	OCDSB_9	OCDS L2 Debug Line B9
		I		CCPOS1_1	CCU61 Hall input signal 1
		I		TREADY1	MLI1 transmit channel ready input
		0		SLSO0_6	SSC0 Slave Select output 6
P3.10	B16	0	PUC	OCDSB_10	OCDS L2 Debug Line B10
		I		CCPOS1_2	CCU61 Hall input signal 2
		0		TVALID1	MLI1 transmit channel valid output
		0		SLSO1_6	SSC1 Slave Select output 6



Table 2-1 Pin Definitions and Functions (cont'd)

Symbol	Pin	In Out	PU/ PD ¹⁾	Functions	
P3.11	C16	0	PUC	OCDSB_11	OCDS L2 Debug Line B11
		0		TDATA1	MLI1 transmit channel data output
		0		SLSO0_7	SSC0 Slave Select output 7
		1		CC61_T12HR	CCU61 Timer 12 hardware run
P3.12	D16	0	PUC	OCDSB_12	OCDS L2 Debug Line B12
		1		RCLK1	MLI1 receive channel clock input
		0		SLSO1_7	SSC1 Slave Select output 7
		1		CC61_T13HR	CCU61 Timer 13 hardware run
P3.13	K13	0	PUC	OCDSB_13	OCDS L2 Debug Line B13
		0		RREADY1	MLI1 receive channel ready output
		I/O		MRST1B	SSC1 master receive/slave
					transmit input/output B
P3.14	J14	0	PUC	OCDSB_14	OCDS L2 Debug Line B14
		1		RVALID1	MLI1 receive channel valid input
		I/O		MTSR1B	SSC1 master transmit/slave
					receive input/output B
P3.15	J15	0	PUC	OCDSB_15	OCDS L2 Debug Line B15
		1		RDATA1	MLI1 receive channel data input
		I/O		SCLK1B	SSC1 clock input/output line B



Table 2-1 Pin Definitions and Functions (cont'd)

Symbol	Pin	In Out	PU/ PD ¹⁾	Functions		
P4		I/O		Port 4 Port 4 is an 8-bit bi-directional general purpose I/O port which can be alternatively used for USB, MLI0 and		
P4.0	R8	I O	PUC	SCU. USBCLK TCLK0B	48 MHz input clock MLI0 transmit channel clock output B	
P4.1	R9	I I	PUC	RCVI TREADY0B	USB data input MLI0 transmit channel ready input B	
P4.2	N7	ĺ	PUC	VPI	USB D+ CMOS level mirror of differential signal	
P4.3	N6	O	PUC	TVALID0B VMI	MLI0 transmit channel valid output B USB D- CMOS level mirror of differential signal	
P4.4	P6	0	PUC	TDATA0B VPO RCLK0B	MLI0 transmit channel data output B USB D+ CMOS level output MLI0 receive channel clock input B	
P4.5	R7	0	PUC	VMO RREADY0B	USB D- CMOS level output MLI0 receive channel ready output B	
P4.6	R6	0	PUC	USBOE RVALID0B	Direction select for transmit or receive MLI0 receive channel valid input B	
P4.7	P5	I O	PUC	RDATA0B BRKOUT_A	MLI0 receive channel data input B OCDS Break Out A	
HDRST	N5	I/O	PUA	Assertion of the synchronous circuitry. This clock cycles. The internal reto a power-on wake-up rese	eset Input/Reset Indication Output his bi-directional open-drain pin causes a reset of the chip through external pin must be driven for a minimum $4f_{\rm CPU}$ eset circuitry drives this pin in response h, hardware, watchdog and power-down t for a specific period of time. For a t, activation of this pin is programmable.	
PORST	R5	I	PUC	Power-on Reset Input A low level on PORST causes an asynchronous reset of the entire chip. PORST is a fully asynchronous level sensitive signal.		
NMI	Т7	I	PUC	A high-to-low	le Interrupt Input transition on this pin causes an uest to the CPU.	



Table 2-1 Pin Definitions and Functions (cont'd)

Symbol	Pin	In Out	PU/ PD ¹⁾	Functions	
TRST	T11	I	PDC	JTAG Module Reset/Enable Input A low level at this pin resets and disables the JTAG module. A high level enables the JTAG module.	
TCK	T12	I	PUC	JTAG Module Clock Input	
TDI	T13	I	PUC	JTAG Module Serial Data Input	
TDO	T10	0		JTAG Module Serial Data Output	
TMS	T9	ı	PUC	JTAG Module State Machine Control Input	
TRCLK	T8	0	_	Trace Clock for OCDS_L2 Lines	
HWCFG0 HWCFG1 HWCFG2	M14 L14 T6		PUC PUC PDC	Hardware Configuration Inputs The Configuration Inputs define the boot options of the TC1130 after a hardware invoked reset operation.	
BRKIN	T5	I	PUC	OCDS Break Input A low level on this pin causes a break in the chip's execution when the OCDS is enabled. In addition, the level of this pin during power-on reset determines the boot configuration.	
MII_ TXCLK	T2	I	PDC	+	
MII_ RXCLK	R2	I	PDC	Ethernet Controller Receive Clock MII_RXCLK is a continuous clock. Its frequency is 25 MHz for 100 Mbit/sec operation, and 2.5 MHz for 10 Mbit/sec. MII_RXD[3:0], MII_RXDV and MII_EXER are driven by the PHY off the falling edge of MII_RXCLK and sampled on the rising edge of MII_RXCLK.	
MII_ MDIO	R1	I/O	PDA	Ethernet Controller Management Data Input/ Output When a read command is being executed, the data that is clocked out of the PHY will be presented on the input line. When the Core is clocking control or data onto the MII_MDIO line, the signal will carry the information.	
D+	T14	I/O		USB D+ Data Line	



Table 2-1 Pin Definitions and Functions (cont'd)

Symbol	Pin	In Out	PU/ PD ¹⁾	Functions
D-	T15	I/O	_	USB D- Data Line
CS0 CS1 CS2 CS3	D9 D8 C9 B8	0 0 0	PUC PUC PUC PUC	EBU Chip Select Output Line 0 EBU Chip Select Output Line 1 EBU Chip Select Output Line 2 EBU Chip Select Output Line 3 Each corresponds to a programmable region. Only one can be active at one time.
CSCOMB	N3	0	PUC	EBU Chip Select Output for combination function (Overlay Memory and Global)
SDCLKI	J1	I		SDRAM Clock Input (Clock Feedback)
SDCLKO	H1	0		SDRAM Clock Output Accesses to SDRAM devices are synchronized to this clock.
RAS	D6	0	PUC	EBU SDRAM Row Address Strobe Output
CAS	D5	0	PUC	EBU SDRAM Column Address Strobe Output
CKE	L4	0	PUC	EBU SDRAM Clock Enable Output
BFCLKI	D1	I		Burst Flash Clock Input (Clock Feedback)
BFCLKO	E1	0		Burst Flash Clock Output Accesses to Burst Flash devices are synchronized to this clock.
RD	P2	0	PUC	EBU Read Control Line Output in master mode Input in slave mode
RD/WR	Т3	0	PUC	EBU Write Control Line Output in master mode Input in slave mode
WAIT	B9	I	PUC	EBU Wait Control Line
ALE	R3	0	PDC	EBU Address Latch Enable Output
MR/W	P3	0	PUC	EBU Motorola-style Read/Write Output
BAA	A11	0	PUC	EBU Burst Address Advance Output
				For advancing address in a Burst Flash access
ADV	B11	0	PUC	EBU Burst Flash Address Valid Output



Table 2-1 Pin Definitions and Functions (cont'd)

Symbol	Pin	In Out	PU/ PD ¹⁾	Functions
				EBU Address/Data Bus Input/Output Lines
AD0	C8	I/O	PUC	EBU Address/Data Bus Line 0
AD1	C7	I/O	PUC	EBU Address/Data Bus Line 1
AD2	B6	I/O	PUC	EBU Address/Data Bus Line 2
AD3	C6	I/O	PUC	EBU Address/Data Bus Line 3
AD4	C5	I/O	PUC	EBU Address/Data Bus Line 4
AD5	A3	I/O	PUC	EBU Address/Data Bus Line 5
AD6	A2	I/O	PUC	EBU Address/Data Bus Line 6
AD7	C3	I/O	PUC	EBU Address/Data Bus Line 7
AD8	C2	I/O	PUC	EBU Address/Data Bus Line 8
AD9	D2	I/O	PUC	EBU Address/Data Bus Line 9
AD10	F1	I/O	PUC	EBU Address/Data Bus Line 10
AD11	E3	I/O	PUC	EBU Address/Data Bus Line 11
AD12	F3	I/O	PUC	EBU Address/Data Bus Line 12
AD13	G1	I/O	PUC	EBU Address/Data Bus Line 13
AD14	H2	I/O	PUC	EBU Address/Data Bus Line 14
AD15	G3	I/O	PUC	EBU Address/Data Bus Line 15
AD16	D7	I/O	PUC	EBU Address/Data Bus Line 16
AD17	B5	I/O	PUC	EBU Address/Data Bus Line 17
AD18	A4	I/O	PUC	EBU Address/Data Bus Line 18
AD19	B4	I/O	PUC	EBU Address/Data Bus Line 19
AD20	C4	I/O	PUC	EBU Address/Data Bus Line 20
AD21	B3	I/O	PUC	EBU Address/Data Bus Line 21
AD22	B2	I/O	PUC	EBU Address/Data Bus Line 22
AD23	B1	I/O	PUC	EBU Address/Data Bus Line 23
AD24	C1	I/O	PUC	EBU Address/Data Bus Line 24
AD25	D3	I/O	PUC	EBU Address/Data Bus Line 25
AD26	E2	I/O	PUC	EBU Address/Data Bus Line 26
AD27	F2	I/O	PUC	EBU Address/Data Bus Line 27
AD28	F4	I/O	PUC	EBU Address/Data Bus Line 28
AD29	G4	I/O	PUC	EBU Address/Data Bus Line 29
AD30	H3	I/O	PUC	EBU Address/Data Bus Line 30
AD31	G2	I/O	PUC	EBU Address/Data Bus Line 31
BC0	A5	0	PUC	EBU Byte Control Line 0
BC1	A6	0	PUC	EBU Byte Control Line 1
BC2	B7	0	PUC	EBU Byte Control Line 2
BC3	A7	0	PUC	EBU Byte Control Line 3



Table 2-1 Pin Definitions and Functions (cont'd)

Symbol	Pin	In Out	PU/ PD ¹⁾	Functions
				EBU Address Bus Input/Output Lines
A0	K1	0	PUC	EBU Address Bus Line 0
A1	L1	0	PUC	EBU Address Bus Line 1
A2	M1	0	PUC	EBU Address Bus Line 2
A3	N1	0	PUC	EBU Address Bus Line 3
A4	P1	0	PUC	EBU Address Bus Line 4
A5	J2	0	PUC	EBU Address Bus Line 5
A6	K2	0	PUC	EBU Address Bus Line 6
A7	L2	0	PUC	EBU Address Bus Line 7
A8	M2	0	PUC	EBU Address Bus Line 8
A9	N2	0	PUC	EBU Address Bus Line 9
A10	J3	0	PUC	EBU Address Bus Line 10
A11	K3	O	PUC	EBU Address Bus Line 11
A12	L3	0	PUC	EBU Address Bus Line 12
A13	M3	0	PUC	EBU Address Bus Line 13
A14	K4	0	PUC	EBU Address Bus Line 14
A15	A8	0	PUC	EBU Address Bus Line 15
A16	A9	O	PUC	EBU Address Bus Line 16
A17	A10	0	PUC	EBU Address Bus Line 17
A18	B10	0	PUC	EBU Address Bus Line 18
A19	C10	0	PUC	EBU Address Bus Line 19
A20	D10	0	PUC	EBU Address Bus Line 20
A21	T4	0	PUC	EBU Address Bus Line 21
A22	R4	0	PUC	EBU Address Bus Line 22
A23	P4	0	PUC	EBU Address Bus Line 23
XTAL1 XTAL2	M16 N16	0		Oscillator/PLL/Clock Generator Input/Output Pins XTAL1 is the input to the main oscillator amplifier and input to the internal clock generator. XTAL2 is the output of the main oscillator amplifier circuit. For clocking of the device from an external source, XTAL1 is driven with the clock signal while XTAL2 is left unconnected. For crystal oscillator operation, XTAL1 and XTAL2 are connected to the crystal with the appropriate recommended oscillator circuitry.
$\overline{V_{ extsf{DDOSC3}}}$	P16	_	_	Main Oscillator Power Supply (3.3 V)
$\overline{V_{\sf SSOSC3}}$	R16		_	Main Oscillator Ground
$\overline{V_{DDOSC}}$	L16	_	_	Main Oscillator Power Supply (1.5 V)



Table 2-1 Pin Definitions and Functions (cont'd)

Symbol	Pin	In Out	PU/ PD ¹⁾	Functions
$\overline{V_{SSOSC}}$	L15	_		Main Oscillator Ground
V_{DD}	G7 G8 G9 G10 G13 K7,K8 K9	_		Core and Logic Power Supply (1.5 V)
$\overline{V_{DDP}}$	D4 D13 H4 J13 M4 N13	_	_	Ports Power Supply (3.3 V)
$\overline{V_{SS}}$	E4 E13 H7 H8 H9 H10 H13 J4,J7 J8,J9 J10 M13 N4	_		Ground
N.C.	A1 A16 T1 T16		_	Not Connected These pins must not be connected.

¹⁾ Refers to internal pull-up or pull-down device connected and corresponding type. The notation '—' indicates that the internal pull-up or pull-down device is not enabled.

Note: P2.12 to P2.15 are always configured as open drain.



3 Functional Description

3.1 On-Chip Memories

The TC1130 provides the following on-chip memories:

- · Program Memory Interface (PMI) with
 - 32-Kbyte Scratch-pad Code RAM (SPRAM)
 - 16-Kbyte Instruction Cache Memory (ICACHE)
- · Data Memory Interface (DMI) with
 - 28-Kbyte Scratch-pad Data RAM (SPRAM)
 - 4-Kbyte Data Cache Memory (DCACHE)
- · Data Memory Unit (DMU) with
 - 64-Kbyte SRAM
- 16-Kbyte Boot ROM (BROM)



3.2 Address Map

Table 3-1 defines the specific segment oriented address blocks of the TC1130 with its address range, size, and PMI/DMI access view. **Table 3-2** shows the block address map of the Segment 15 which includes on-chip peripheral units and ports.

Table 3-1 TC1130 Block Address Map

Address Range	Size	Description	DMI Acc.	PMI Acc.	
0000 0000 _H – 7FFF FFFF _H	2 GB	MMU Space	via FPI	via FPI	c a
8000 0000 _H – 8FFF FFFF _H	256 MB	External Memory Space mapped from Segment 10	via LMB	via LMB	c h
9000 0000 _H – 9FDF FFFF _H	256 MB	Reserved	via FPI	via FPI	e d
A000 0000 _H – AFBF FFFF _H	252 MB	External Memory Space	via LMB	via LMB	n o
AFC0 0000 _H – AFC0 FFFF _H	64 KB	DMU Space			n- c
AFC1 0000 _H – AFFF FFFF _H	~4 MB	Reserved			∣a c - h
B000 0000 _H – BFFF FFFF _H	256 MB	Reserved	via FPI	via FPI	e d
C000 0000 _H – C000 FFFF _H	64 KB	DMU	via LMB	via LMB	c a
C001 0000 _H – CFFF FFFF _H	~ 256 MB	Reserved			c h e d
	Range 0000 0000 _H - 7FFF FFFF _H 8000 0000 _H - 8FFF FFFF _H 9000 0000 _H - 9FDF FFFF _H A000 0000 _H - AFBF FFFF _H AFC0 0000 _H - AFC0 FFFF _H AFC1 0000 _H - AFFF FFFF _H B000 0000 _H - BFFF FFFF _H C000 0000 _H - C000 FFFF _H	$\begin{array}{c ccccccccccccccccccccccccccccccccccc$	Range Image Image <th< td=""><td>Range . Acc. 0000 0000_H - 2 GB MMU Space via FPI 8000 0000_H - 256 MB External Memory Space mapped from Segment 10 via LMB 9000 0000_H - 256 MB Reserved via FPI A000 0000_H - 252 MB External Memory Space via LMB AFC0 0000_H - 64 KB DMU Space LMB AFC1 0000_H - ~4 MB Reserved via FPI B000 0000_H - 256 MB Reserved via FPI C000 0000_H - 64 KB DMU via FPI C000 0000_H - 64 KB DMU via LMB C000 FFFF_H 64 KB DMU Via LMB C001 0000_H - ~ 256 Reserved LMB</td><td>Range . Acc. Acc. 0000 0000_H − 7FFF FFFH 2 GB MMU Space via FPI via FPI 8000 0000_H − 8FFF FFFH 256 MB External Memory Space mapped from Segment 10 via LMB LMB 9000 0000_H − 9FDF FFFH 256 MB Reserved via FPI via FPI A000 0000_H − AFBF FFFH 252 MB External Memory Space via LMB LMB AFC0 0000_H − AFC0 FFFH 64 KB DMU Space via LMB LMB AFC1 0000_H − AFFF FFFH 256 MB Reserved via FPI FPI B000 0000_H − BFFF FFFH 256 MB Reserved via FPI Via FPI C000 0000_H − C000 FFFFH 64 KB DMU via LMB LMB C001 0000_H − C000 FFFFH ~ 256 Reserved Reserved Via LMB</td></th<>	Range . Acc. 0000 0000 _H - 2 GB MMU Space via FPI 8000 0000 _H - 256 MB External Memory Space mapped from Segment 10 via LMB 9000 0000 _H - 256 MB Reserved via FPI A000 0000 _H - 252 MB External Memory Space via LMB AFC0 0000 _H - 64 KB DMU Space LMB AFC1 0000 _H - ~4 MB Reserved via FPI B000 0000 _H - 256 MB Reserved via FPI C000 0000 _H - 64 KB DMU via FPI C000 0000 _H - 64 KB DMU via LMB C000 FFFF _H 64 KB DMU Via LMB C001 0000 _H - ~ 256 Reserved LMB	Range . Acc. Acc. 0000 0000 _H − 7FFF FFFH 2 GB MMU Space via FPI via FPI 8000 0000 _H − 8FFF FFFH 256 MB External Memory Space mapped from Segment 10 via LMB LMB 9000 0000 _H − 9FDF FFFH 256 MB Reserved via FPI via FPI A000 0000 _H − AFBF FFFH 252 MB External Memory Space via LMB LMB AFC0 0000 _H − AFC0 FFFH 64 KB DMU Space via LMB LMB AFC1 0000 _H − AFFF FFFH 256 MB Reserved via FPI FPI B000 0000 _H − BFFF FFFH 256 MB Reserved via FPI Via FPI C000 0000 _H − C000 FFFFH 64 KB DMU via LMB LMB C001 0000 _H − C000 FFFFH ~ 256 Reserved Reserved Via LMB



Table 3-1 TC1130 Block Address Map (cont'd)

Seg- ment	Address Range	Size	Description	DMI Acc.	PMI Acc.	
	D000 0000 _H – D000 6FFF _H	28 KB DMI Local Data RAM (LDRAM)		DMI local	via LMB	
	D000 7000 _H – D3FF FFFF _H	~ 64 MB	Reserved			
	D400 0000 _H – D400 7FFF _H	32 KB	PMI Local Code Scratch Pad RAM (SPRAM)	via LMB	PMI local	
13	D400 8000 _H – D7FF FFFF _H	~64 MB	Reserved			
	D800 0000 _H – DDFF FFFF _H	96 MB	External Memory Space	via	via	
	DE00 0000 _H – DEFF FFFF _H	16 MB	Emulator Memory Space	LMB LMB		
	DF00 0000 _H – DFFF BFFF _H	~16 MB	Reserved	_	_	
	DFFF C000 _H – DFFF FFFF _H	16 KB	Boot ROM Space	via FPI	via FPI	ped
	E000 0000 _H – 128 MB E7FF FFFF _H		External Memory Space	via LMB	via LMB	non-cached
14	E800 0000 _H – E83F FFFF _H	4 MB	Reserved for mapped space for lower 4 Mbytes of Local Memory in Segment 12 (Transformed by LFI bridge to C000 0000 _H – C03F FFFF _H)	access only from FPI bus side of LFI	access only from FPI bus side of LFI	lou
	E840 0000 _H – E84F FFFF _H	1 MB	Reserved for mapped space for lower 1 Mbyte of Local Memory in Segment 13 (Transformed by LFI bridge to D000 0000 _H – D00F FFFF _H)	access only from FPI bus	access only from FPI bus	
	E850 0000 _H – E85F FFFF _H	1 MB	Reserved for mapped space for 1 Mbyte of Local Memory in Segment 13 (Transformed by LFI bridge to D400 0000 _H – D40F FFFF _H)	side of LFI	side of LFI	



Table 3-1 TC1130 Block Address Map (cont'd)

Seg- ment	Address Range	Size	Description	DMI Acc.	PMI Acc.	
14	E860 0000 _H – EFFF FFFF _H	122 MB	Reserved	_	_	n o
15	F000 0000 _H – FFFF FFFF _H	256 MB	See Table 3-2	via LMB or via FPI	via LMB or via FPI	n- c a c h e d

Table 3-2 Block Address Map of Segment 15

Symbol	Description	Address Range	Size				
System Peripheral Bus (SPB)							
SCU	System Control Unit (incl. WDT)	F000 0000 _H - F000 00FF _H	256 Bytes				
SBCU	FPI Bus Control Unit	F000 0100 _H - F000 01FF _H	256 Bytes				
STM	System Timer	F000 0200 _H - F000 02FF _H	256 Bytes				
OCDS	On-Chip Debug Support (Cerberus)	F000 0300 _H - F000 03FF _H	256 Bytes				
_	Reserved	F000 0400 _H - F000 04FF _H	256 Bytes				
_	Reserved	F000 0500 _H - F000 05FF _H	256 Bytes				
GPTU	General Purpose Timer Unit	F000 0600 _H - F000 06FF _H	256 Bytes				
_	Reserved	F000 0700 _H - F000 07FF _H	256 Bytes				
_	Reserved	F000 0800 _H - F000 08FF _H	256 Bytes				
_	Reserved	F000 0900 _H - F000 09FF _H	256 Bytes				
_	Reserved	F000 0A00 _H - F000 0AFF _H	256 Bytes				
_	Reserved	F000 0B00 _H - F000 0BFF _H	256 Bytes				
P0	Port 0	F000 0C00 _H - F000 0CFF _H	256 Bytes				
P1	Port 1	F000 0D00 _H - F000 0DFF _H	256 Bytes				
P2	Port 2	F000 0E00 _H - F000 0EFF _H	256 Bytes				
P3	Port 3	F000 0F00 _H - F000 0FFF _H	256 Bytes				
P4	Port 4	F000 1000 _H - F000 10FF _H	256 Bytes				
_	Reserved	F000 1100 _H - F000 11FF _H	256 Bytes				



Table 3-2 Block Address Map of Segment 15 (cont'd)

Symbol	Description	Address Range	Size
_	Reserved	F000 1200 _H - F000 12FF _H	256 Bytes
_	Reserved	F000 1300 _H - F000 13FF _H	256 Bytes
_	Reserved	F000 1400 _H - F000 14FF _H	256 Bytes
_	Reserved	F000 1500 _H - F000 15FF _H	256 Bytes
_	Reserved	F000 1600 _H - F000 16FF _H	256 Bytes
_	Reserved	F000 1700 _H - F000 17FF _H	256 Bytes
_	Reserved	F000 1800 _H - F000 18FF _H	256 Bytes
_	Reserved	F000 1900 _H - F000 19FF _H	256 Bytes
CCU60	Capture/Compare Unit 0	F000 2000 _H - F000 20FF _H	256 Bytes
CCU61	Capture/Compare Unit 1	F000 2100 _H - F000 21FF _H	256 Bytes
_	Reserved	F000 2200 _H - F000 3BFF _H	_
DMA	Direct Memory Access Controller	F000 3C00 _H - F000 3EFF _H	3 × 256 Bytes
_	Reserved	F000 3F00 _H - F000 3FFF _H	_
CAN	MultiCAN Controller	F000 4000 _H - F000 5FFF _H	8 Kbytes
_	Reserved	F000 6000 _H - F00E 1FFF _H	_
USB	USB RAM based Registers	F00E 2000 _H - F00E 219F _H	416 Bytes
USB	USB RAM	F00E 21A0 _H - F00E 27FF _H	1.6 Kbytes
USB	USB Registers	F00E 2800 _H - F00E 28FF _H	256 Bytes
_	Reserved	F00E 2900 _H - F00F FFFF _H	_
Units on	SMIF Interface of DMA Controller		
_	Reserved	F010 0000 _H - F010 00FF _H	256 Bytes
SSC0	Synchronous Serial Interface 0	F010 0100 _H - F010 01FF _H	256 Bytes
SSC1	Synchronous Serial Interface 1	F010 0200 _H - F010 02FF _H	256 Bytes
ASC0	Async./Sync. Serial Interface 0	F010 0300 _H - F010 03FF _H	256 Bytes
ASC1	Async./Sync. Serial Interface 1	F010 0400 _H - F010 04FF _H	256 Bytes
ASC2	Async./Sync. Serial Interface 2	F010 0500 _H - F010 05FF _H	256 Bytes
I2C	Inter IC	F010 0600 _H - F010 06FF _H	256 Bytes
_	Reserved	F010 0700 _H - F010 BFFF _H	_
MLI0	Micro Link Interface 0	F010 C000 _H - F010 C0FF _H	256 Bytes



 Table 3-2
 Block Address Map of Segment 15 (cont'd)

Symbol	Description	Address Range	Size
MCHK	Memory Checker	F010 C200 _H - F010 C2FF _H	256 Bytes
_	Reserved	F010 C300 _H - F01D FFFF _H	_
MLI0_ SP0	MLI0 Small Transfer Window 0	F01E 0000 _H - F01E 1FFF _H	8 Kbytes
MLI0_ SP1	MLI0 Small Transfer Window 1	F01E 2000 _H - F01E 3FFF _H	8 Kbytes
MLI0_ SP2	MLI0 Small Transfer Window 2	F01E 4000 _H - F01E 5FFF _H	8 Kbytes
MLI0_ SP3	MLI0 Small Transfer Window 3	F01E 6000 _H - F01E 7FFF _H	8 Kbytes
MLI1_ SP0	MLI1 Small Transfer Window 0	F01E 8000 _H - F01E 9FFF _H	8 Kbytes
MLI1_ SP1	MLI1 Small Transfer Window 1	F01E A000 _H - F01E BFFF _H	8 Kbytes
MLI1_ SP2	MLI1 Small Transfer Window 2	F01E C000 _H - F01E DFFF _H	8 Kbytes
MLI1_ SP3	MLI1 Small Transfer Window 3	F01E E000 _H - F01E FFFF _H	8 Kbytes
_	Reserved	F01F 0000 _H - F01F FFFF _H	_
MLI0_ LP0	MLI0 Large Transfer Window 0	F020 0000 _H - F020 FFFF _H	64 Kbytes
MLI0_ LP1	MLI0 Large Transfer Window 1	F021 0000 _H - F021 FFFF _H	64 Kbytes
MLI0_ LP2	MLI0 Large Transfer Window 2	F022 0000 _H - F022 FFFF _H	64 Kbytes
MLI0_ LP3	MLI0 Large Transfer Window 3	F023 0000 _H - F023 FFFF _H	64 Kbytes
MLI1_ LP0	MLI1 Large Transfer Window 0	F024 0000 _H - F024 FFFF _H	64 Kbytes
MLI1_ LP1	MLI1 Large Transfer Window 1	F025 0000 _H - F025 FFFF _H	64 Kbytes
MLI1_ LP2	MLI1 Large Transfer Window 2	F026 0000 _H - F026 FFFF _H	64 Kbytes



Table 3-2 Block Address Map of Segment 15 (cont'd)

Table 0-2 Block Address Map of Deginent To (cont a)			
Symbol	Description	Address Range	Size
MLI1_ LP3	MLI1 Large Transfer Window 3	F027 0000 _H - F027 FFFF _H	64 Kbytes
_	Reserved	F028 0000 _H - F200 00FF _H	_
ECU	Ethernet Controller Unit	F200 0100 _H - F200 05FF _H	1280Bytes
_	Reserved	F200 0600 _H - F7E0 FEFF _H	_
CPU (Part of System Peripheral Bus)			
CPU SFRs	CPU Slave Interface	F7E0 FF00 _H - F7E0 FFFF _H	256 Bytes
	Reserved	F7E1 0000 _H - F7E1 7FFF _H	_
	MMU	F7E1 8000 _H - F7E1 80FF _H	256 Bytes
	Reserved	F7E1 8100 _H - F7E1 BFFF _H	_
	Memory Protection Registers	F7E1 C000 _H -F7E1 EFFF _H	12 Kbytes
	Reserved	F7E1 F000 _H - F7E1 FCFF _H	_
	Core Debug Register (OCDS)	F7E1 FD00 _H -F7E1 FDFF _H	256 Bytes
	Core Special Function Registers (CSFRs)	F7E1 FE00 _H -F7E1 FEFF _H	256 Bytes
	General Purpose Register (GPRs)	F7E1 FF00 _H - F7E1 FFFF _H	256 Bytes
_	Reserved	F7E2 0000 _H - F7FF FFFF _H	_
Local Memory Buses (LMB)			
EBU	External Bus Interface Unit	F800 0000 _H - F800 03FF _H	1 Kbyte
DMU	Data Memory Unit	F800 0400 _H - F800 04FF _H	256 Bytes
_	Reserved	F800 0500 _H - F87F FBFF _H	_
DMI	Data Memory Interface Unit	F87F FC00 _H - F87F FCFF _H	256 Bytes
PMI	Program Memory Interface Unit	F87F FD00 _H - F87F FDFF _H	256 Bytes
LBCU	Local Memory Bus Control Unit	F87F FE00 _H - F87F FEFF _H	256 Bytes
LFI	LMB to FPI Bus Bridge	F87F FF00 _H - F87F FFFF _H	256 Bytes
_	Reserved	F880 0000 _H - FFFF FFFF _H	_



3.3 Memory Protection System

The TC1130 memory protection system specifies the addressable range and read/write permissions of memory segments available to the currently executing task. The memory protection system controls the position and range of addressable segments in memory. It also controls the types of read and write operations allowed within addressable memory segments. Any illegal memory access is detected by the memory protection hardware, which then invokes the appropriate Trap Service Routine (TSR) to handle the error. Thus, the memory protection system protects critical system functions against both software and hardware errors. The memory protection hardware can also generate signals to the Debug Unit to facilitate tracing illegal memory accesses.

In TC1130, TriCore™ supports two address spaces: the virtual address space and the physical address space. Both address spaces are 4 Gbytes in size and are divided into 16 segments with each segment being 256 Mbytes. The upper 4 bits of the 32-bit address are used to identify the segment. Virtual segments are numbered 0 - 15. But a virtual address is always translated into a physical address before accessing memory. The virtual address is translated into a physical address using one of two translation mechanisms: (a) direct translation, and (b) Page Table Entry (PTE) based translation. If the virtual address belongs to the upper half of the virtual address space then the virtual address is directly used as the physical address (direct translation). If the virtual address belongs to the lower half of the address space, then the virtual address is used directly as the physical address if the processor is operating in physical mode (direct translation) or translated using a Page Table Entry if the processor is operating in Virtual mode (PTE translation). These are managed by Memory Management Unit (MMU).

Memory protection is enforced using separate mechanisms for the two translation paths.

3.3.1 Protection for Direct translation

Memory protection for addresses that undergo direct translation is enforced using the range based protection that has been used in the previous generation of the TriCore™ architecture. The range based protection mechanism provides support for protecting memory ranges from unauthorized read, write, or instruction fetch accesses. The TriCore™ architecture provides up to four protection register sets with the PSW.PRS field controlling the selection of the protection register set. Because the TC1130 uses a Harvard-style memory architecture, each Memory Protection Register Set is broken down into a Data Protection Register Set and a Code Protection Register Set. Each Data Protection Register Set can specify up to four address ranges to receive particular protection modes. Each Code Protection Register Set can specify up to two address ranges to receive particular protection modes.

Each of the Data Protection Register Sets and Code Protection Register Sets determines the range and protection modes for a separate memory area. Each contains register pairs which determine the address range (the Data Segment Protection Registers and Code Segment Protection Registers) and one register (Data Protection



Mode Register) which determines the memory access modes which apply to the specified range.

3.3.2 Protection for PTE based translation

Memory protection for addresses that undergo PTE based translation is enforced using the PTE used for the address translation. The PTE provides support for protecting a process from unauthorized read, write, or instruction fetches by other processes. The PTE has the following bits that are provided for the purpose of protection:

- Execute Enable (XE) enables instruction fetch to the page
- Write Enable (WE) enables data writes to the page
- Read Enable (RE) enables data reads from the page

Furthermore, User-0 accesses to virtual addresses in the upper half of the virtual address space are disallowed when operating in virtual mode. In physical mode, User-0 accesses are disallowed only to segments 14 and 15. Any User-0 access to a virtual address that is restricted to User-1 or supervisor mode will cause a Virtual Address Protection (VAP) Trap in both the physical and virtual modes.

3.3.3 Memory Checker

The Memory Checker module (MCHK) makes it possible to check the data consistency of memories. It uses DMA moves to read from the selected address area and to write the value read in a memory checker input register (the moves should be 32-bit moves). A polynomial checksum calculation is done with each write operation to the memory checker input register.



3.4 On-Chip Bus System

The TC1130 includes two bus systems:

- Local Memory Bus (LMB)
- Flexible Peripheral Interface Bus (FPI)

The LMB-to-FPI (LFI) bridge interconnects the FPI bus and LMB Bus.

3.4.1 Local Memory Bus (LMB)

The Local Memory Bus interconnects the memory units and functional units, such as CPU and DMU. The main objective of the LMB bus is to support devices with fast response time. This allows the DMI and PMI fast access to local memory and reduces load on the FPI bus. The TriCore™ system itself is located on the LMB bus. Via External Bus Unit, it interconnects TC1130 and external components.

The Local Memory Bus is a synchronous, pipelined, split bus with variable block size transfer support. It supports 8, 16, 32 and 64 bits single beat transactions and variable length 64 bits block transfers.

Features:

The LMB provides the following features:

- Synchronous, Pipelined, Multimaster, 64-bit high performance bus
- · Optimized for high speed and high performance
- 32-bit address, 64-bit data buses
- Central, simple per cycle arbitration
- · Slave controlled wait state insertion
- Address pipelining (max depth 2)
- Supports Split transactions
- Supports Variable block size transfer
- Supports Locked transaction (read-modify-write)

3.4.2 Flexible Peripheral Interconnect Bus (FPI)

The FPI Bus is an on-chip bus that is used in modular and highly integrated microprocessors and microcontrollers (**systems-on-chips**). FPI Bus is designed for memory mapped data transfers between its bus agents. Bus agents are on-chip function blocks (modules), equipped with an FPI Bus interface and connected via FPI Bus signals. An FPI Bus agent acts as an FPI Bus master when it initiates data read or data write operations once bus ownership has been granted to the agent. An FPI Bus agent that is addressed by an FPI Bus operation acts as an FPI Bus slave when it performs the requested data read or write operation.



Features:

The FPI Bus is designed with the requirements of high-performance systems in mind. The features are:

- Core independent
- Multimaster capability (up to 16 masters)
- Demultiplexed operation
- Clock synchronous
- Peak transfer rate of up to 800 Mbytes/sec (@ 100 MHz bus clock)
- Address and data bus scalable (address bus up to 32 bits, data bus up to 64 bits)
- 8-/16-/32- and 64-bit data transfers
- Broad range of transfer types from single to multiple data transfers
- · Split transaction support for agents with long response time
- Burst transfer capability
- · EMI and power consumption minimized

3.4.3 LFI

The LMB-to-FPI Interface (LFI) block provides the circuitry to interface (bridge) the FPI bus and the Local Memory Bus (LMB).

LFI Features:

- Full support for bus transactions found within current TriCore™ 1.3 based systems:
 - Single 8/16/32-bit Write/Read transfers from FPI to LMB
 - Single 8/16/32/64-bit Write/Read transfers from LMB to FPI
 - Read-Modify-Write transfers of 8/16/32-bit in both directions
 - Burst transactions of 2, 4 or 8 data beats from the FPI to the LMB
 - Burst transactions of 2 or 4 data beats from the LMB to the FPI
- Address decoding and translation as required by TriCore™ 1.3 implementation
- FPI master interface supports full pipelining on FPI bus
- LMB master interface supports pipelining on LMB within the scope of the LMB specification
- FPI master interface can act as default master on FPI bus
- Programmable support for split LMB to FPI read transactions
- Retry generation on both FPI and LMB buses
- Full support for abort, retry, error and FPI timeout conditions
- Flexible LMB/FPI clock ratio support including dynamic clock switching support
- LFI core clock may be shut down when no transactions are being issued to LFI from either bus and the LFI has no transactions in progress, thus saving power.



3.5 LMB External Bus Unit

The LMB External Bus Control Unit (EBU) of the TC1130 is the interface between external resources, like memories and peripheral units, and the internal resources connected to on-chip buses if enabled. The basic structure and external interconnections of the EBU are shown in **Figure 3-1**.

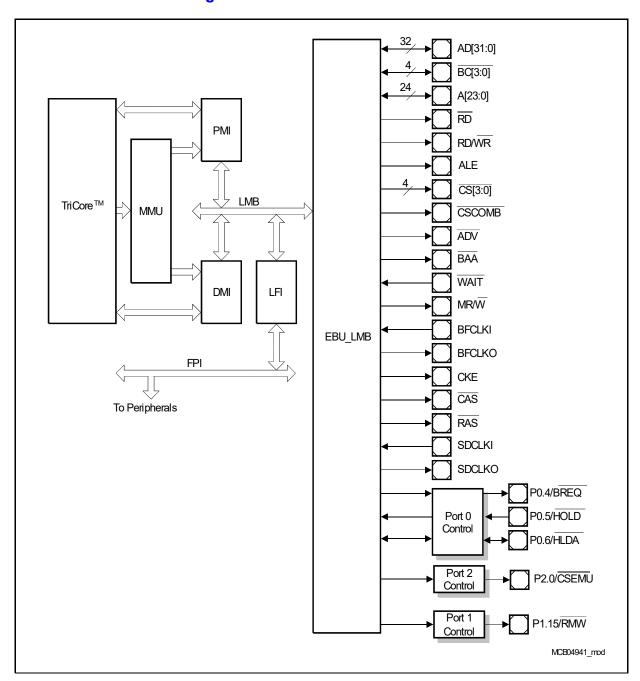


Figure 3-1 EBU Structure and Interface



The EBU is used primarily for any Local Memory Bus (LMB) master accessing external memories. The EBU controls all transactions required for this operation and in particular handles the arbitration between the internal EBU master and the external EBU master.

The types of external devices/bus modes controlled by the EBU are:

- Intel-style peripherals (separate RD and WR signals)
- ROMs, EPROMs
- Static RAMs
- PC100 and PC133 SDRAMs (Burst Read/Write Capacity/Multi-Bank/Page support)
- Specific types of Burst Mode Flash devices
- Special support for external emulator/debug hardware

- Supports 64-bit Local Memory Bus (LMB)
- Supports external bus frequency: internal LMB frequency = 1:1 or 1:2
- Provides highly programmable access parameters
- Supports Intel-style peripherals/devices
- Supports PC100 and PC133 (runs in maximum 120 MHz) SDRAM (burst access, multibanking, precharge, refresh)
- Supports 16- and 32-bit SDRAM data bus and 64-,128-, and 256-Mbit devices
- Supports Burst Flash devices
- Supports Multiplexed access (address and data on the same bus) when PC100 and PC133 SDRAM are not presented on the external bus
- Supports data buffering: Code Prefetch Buffer, Read/Write Buffer
- External master arbitration compatible to C166 and other TriCore™ devices
- Provides 4 programmable address regions (1 dedicated for emulator)
- Provides a CSGLB signal, bit programmable to combine one or more CS lines for buffer control
- Provides RMW signal reflecting read-modify-write action
- · Supports Little Endian byte ordering
- Provides signal for controlling data flow of slow-memory buffer



3.6 Direct Memory Access (DMA)

The Direct Memory Access Controller executes DMA transactions from a source address location to a destination address location, without intervention of the CPU. One DMA transaction is controlled by one DMA channel. Each DMA channel has assigned its own channel register set. The total of 8 channels are provided by one DMA sub-block.

The DMA module is connected to 3 bus interfaces in TC1130, the Flexible Peripheral Interconnect Bus (FPI), the DMA Bus and the Micro Link Bus. It can do transfers on each of the buses as well as between the buses.

In addition, it bridges accesses from the Flexible Peripheral Interconnect Bus to the peripherals on the DMA Bus, allowing easy access to these peripherals by CPU. Clock control, address decoding, DMA request wiring, and DMA interrupt service request control are implementation specific and managed outside the DMA controller kernel.

- 8 independent DMA channels
 - Up to 8 selectable request inputs per DMA channel
 - Programmable priority of DMA channels within a DMA sub-block (2 levels)
 - Software and hardware DMA request generation
 - Hardware requests by selected peripherals and external inputs
- Programmable priority of the DMA sub-block on the bus interfaces
- Buffer capability for move actions on the buses (min. 1 move per bus is buffered).
- Individually programmable operation modes for each DMA channel
 - Single mode: stops and disables DMA channel after a predefined number of DMA transfers
 - Continuous mode: DMA channel remains enabled after a predefined number of DMA transfers; DMA transaction can be repeated.
 - Programmable address modification
- Full 32-bit addressing capability of each DMA channel
 - 4-Gbyte address range
 - Support of circular buffer addressing mode
- Programmable data width of a DMA transaction: 8-bit, 16-bit, or 32-bit
- Micro Link supported
- Register set for each DMA channel
 - Source and destination address register
 - Channel control and status register
 - Transfer count register
- Flexible interrupt generation (the service request node logic for the MLI channels is also implemented in the DMA module)
- All buses/interfaces connected to the DMA module must work at the same frequency
- Read/write requests of the FPI Bus Side to the Remote Peripherals are bridged to the DMA Bus (only the DMA is master on the DMA bus)



The basic structure and external interconnections of the DMA are shown in Figure 3-2.

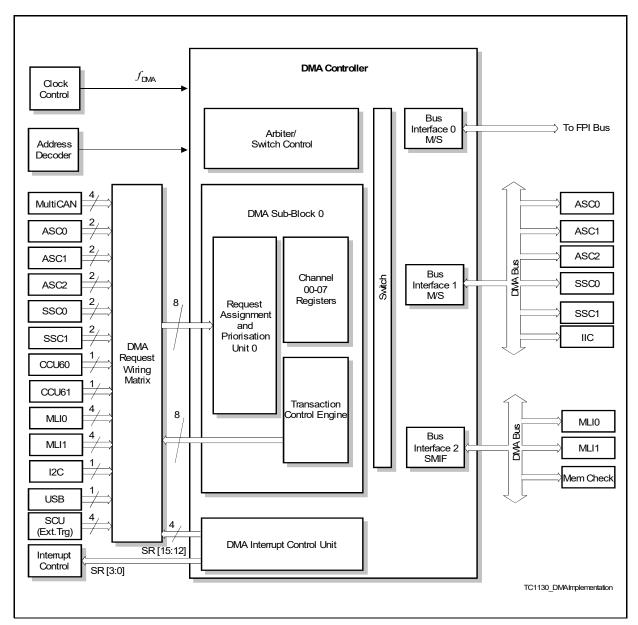


Figure 3-2 DMA Controller Structure and Interconnections



3.7 Interrupt System

An interrupt request can be serviced by the CPU, which is called "Service Provider". Interrupt requests are referred to as "Service Requests" in this document.

Each peripheral in the TC1130 can generate service requests. Additionally, the Bus Control Unit, the Debug Unit, the DMA Controller and even the CPU itself can generate service requests to the Service Provider. As shown in **Figure 3-3**, each unit that can generate service requests is connected to one or multiple Service Request Nodes (SRN). Each SRN contains a Service Request Control Register mod_SRC, where "mod" is the identifier of the unit requesting service. The SRNs are connected to the Interrupt Control Unit (ICU) via the CPU Interrupt Arbitration Bus. The ICU arbitrates service requests for the CPU and administers the Interrupt Arbitration Bus.

Units that can generate service requests are:

- Asynchronous/Synchronous Serial Interfaces (ASC0, ASC1 and ASC2) with 4 SRNs each
- High-Speed Synchronous Serial Interfaces (SSC0 and SSC1) with 3 SRNs each
- Inter IC Interface (IIC) with 3 SRNs
- Universal Serial Bus (USB) with 8 SRNs
- Micro Link Interface MLI0 with 4 SRNs and MLI1 with 2 SRNs
- General Purpose Timer Unit (GPTU) with 8 SRNs
- Capture/Compare Unit (CCU60 and CCU61) with 4 SRNs each
- MultiCAN (CAN) with 16 SRNs
- Ethernet Controller with 9 SRNs
- External Interrupts with 4 SRNs
- Direct Memory Access Controller (DMA) with 4 SRNs
- DMA Bus with 1 SRN
- System Timer (STM) with 2 SRNs
- Bus Control Units (SBCU and LBCU) with 1 SRN each
- Central Processing Unit (CPU) with 4 SRNs
- Floating Point Unit (FPU) with 1 SRN
- Debug Unit (OCDS) with 1 SRN

The CPU can make service requests directly to itself (via the ICU). The CPU Service Request Nodes are activated through software.



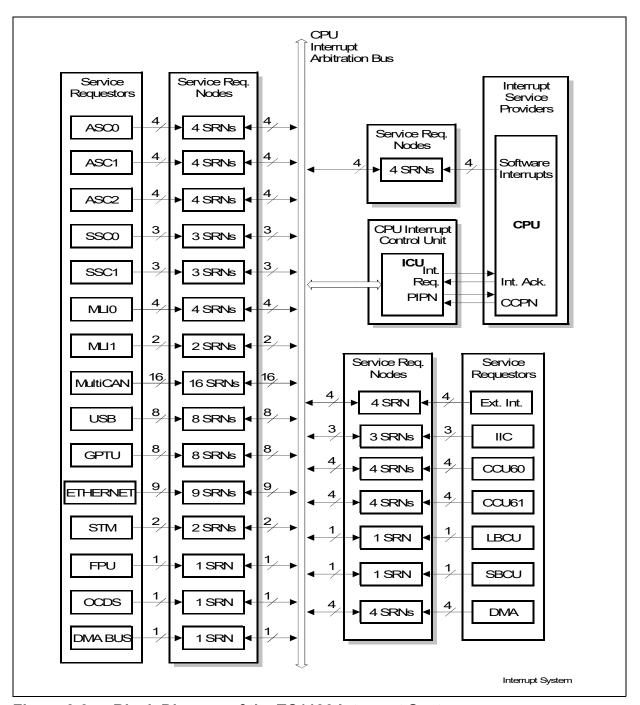


Figure 3-3 Block Diagram of the TC1130 Interrupt System



3.8 Parallel Ports

The TC1130 has 72 digital input/output port lines, which are organized into four parallel 16-bit ports and one parallel 8-bit port, Port P0 to Port P4 with 3.3 V nominal voltage.

The digital parallel ports can be used as general purpose I/O lines or they can perform input/output functions for the on-chip peripheral units. An overview on the port-to-peripheral unit assignment is shown in **Figure 3-4**.

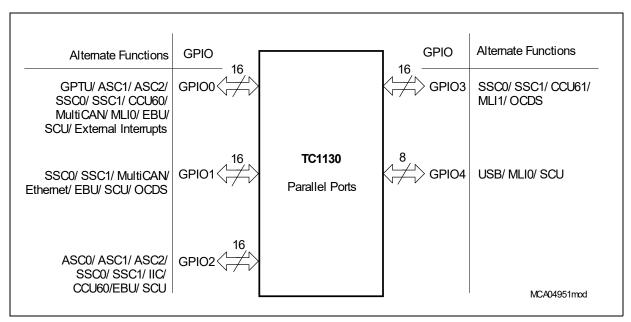


Figure 3-4 Parallel Ports of the TC1130



3.9 Asynchronous/Synchronous Serial Interface (ASC)

Figure 3-5 shows a global view of the functional blocks of three Asynchronous/ Synchronous Serial interfaces (ASC0, ASC1 and ASC2).

Each ASC module (ASC0/ASC1/ASC2) communicates with the external world via one pair of I/O lines. The RXD line is the receive data input signal (in synchronous mode also output). TXD is the transmit output signal. Clock control, address decoding, and interrupt service request control are managed outside the ASC module kernel.

The Asynchronous/Synchronous Serial interfaces provide serial communication between the TC1130 and other microcontrollers, microprocessors or external peripherals.

Each ASC supports full-duplex asynchronous communication and half-duplex synchronous communication. In synchronous mode, data is transmitted or received synchronous to a shift clock which is generated by the ASC internally. In asynchronous mode, 8-bit or 9-bit data transfer, parity generation, and the number of stop bits can be selected. Parity, framing, and overrun error detection are provided to increase the reliability of data transfers. Transmission and reception of data is double-buffered. For multiprocessor communication, a mechanism is included to distinguish address bytes from data bytes. Testing is supported by a loop-back option. A 13-bit baud-rate generator provides the ASC with a separate serial clock signal that can be accurately adjusted by a prescaler implemented as a fractional divider.



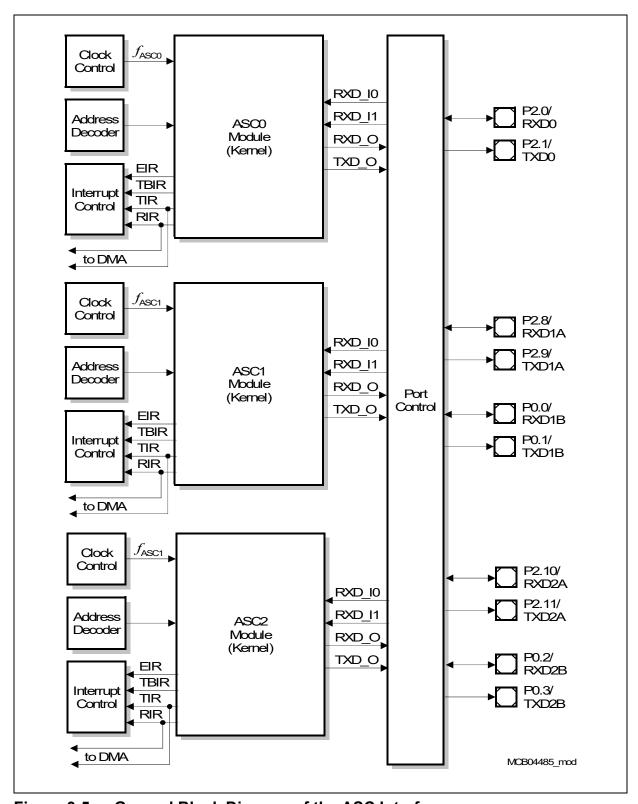


Figure 3-5 General Block Diagram of the ASC Interfaces



- Full-duplex asynchronous operating modes
 - 8-bit or 9-bit data frames, LSB first
 - Parity bit generation/checking
 - One or two stop bits
 - Baud rate from 4.6875 MBaud to 1.1 Baud (@ 75 MHz clock)
- Multiprocessor mode for automatic address/data byte detection
- Loop-back capability
- · Half-duplex 8-bit synchronous operating mode
 - Baud rate from 9.375 MBaud to 762.9 Baud (@ 75 MHz clock)
- Support for IrDA data transmission up to 115.2 kBaud maximum
- Double buffered transmitter/receiver
- Interrupt generation
 - On a transmitter buffer empty condition
 - On a transmit last bit of a frame condition
 - On a receiver buffer full condition
 - On an error condition (frame, parity, overrun error)
- FIFO
 - 8-byte receive FIFO (RXFIFO)
 - 8-byte transmit FIFO (TXFIFO)
 - Independent control of RXFIFO and TXFIFO
 - 9-bit FIFO data width
 - Programmable Receive/Transmit Interrupt Trigger Level
 - Receive and Transmit FIFO filling level indication
 - Overrun error generation
 - Underflow error generation



3.10 High-Speed Synchronous Serial Interface (SSC)

Figure 3-6 shows a global view of the functional blocks of two High-Speed Synchronous Serial interfaces (SSC0 and SSC1).

Each SSC supports full-duplex and half-duplex serial synchronous communication up to 37.5 MBaud (@ 75 MHz module clock) with receive and transmit FIFO support. The serial clock signal can be generated by the SSC itself (master mode) or can be received from an external master (slave mode). Data width, shift direction, clock polarity and phase are programmable. This allows communication with SPI-compatible devices. Transmission and reception of data is double-buffered. A shift clock generator provides the SSC with a separate serial clock signal. Eight slave select inputs are available for slave mode operation. Eight programmable slave select outputs (chip selects) are supported in master mode.

- Master and slave mode operation
 - Full-duplex or half-duplex operation
 - Automatic pad control possible
- Flexible data format
 - Programmable number of data bits: 2 to 16 bits
 - Programmable shift direction: LSB or MSB shift first
 - Programmable clock polarity: idle low or high state for the shift clock
 - Programmable clock/data phase: data shift with leading or trailing edge of the shift clock
- Baud rate generation minimum at 572.2 Baud (@ 75 MHz module clock)
- Interrupt generation
 - On a transmitter empty condition
 - On a receiver full condition
 - On an error condition (receive, phase, baud rate, transmit error)
- Four-pin interface
- Flexible SSC pin configuration
- Up to eight slave select inputs in slave mode
- Up to eight programmable slave select outputs SLSO in master mode
 - Automatic SLSO generation with programmable timing
 - Programmable active level and enable control
- 4-stage receive FIFO (RXFIFO) and 4-stage transmit FIFO (TXFIFO)
 - Independent control of RXFIFO and TXFIFO
 - 2- to 16-bit FIFO data width
 - Programmable receive/transmit interrupt trigger level
 - Receive and transmit FIFO filling level indication
 - Overrun error generation
 - Underflow error generation



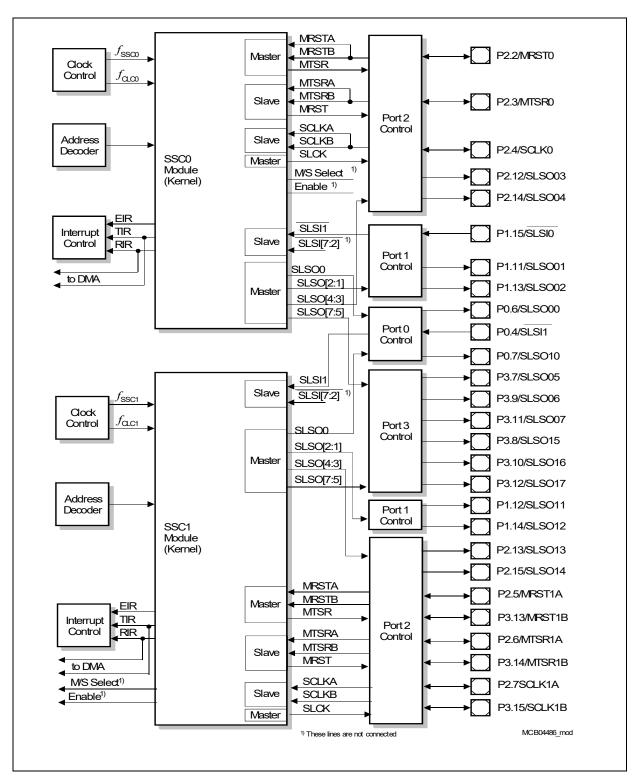


Figure 3-6 General Block Diagram of the SSC Interfaces



3.11 Inter IC Serial Interface (IIC)

Figure 3-7 shows a global view of the functional blocks of the Inter IC Serial interface (IIC).

The IIC module has four I/O lines, located at Port 2. The IIC module is further supplied with clock control, interrupt control and address decoding logic. One DMA request can be generated by IIC module.

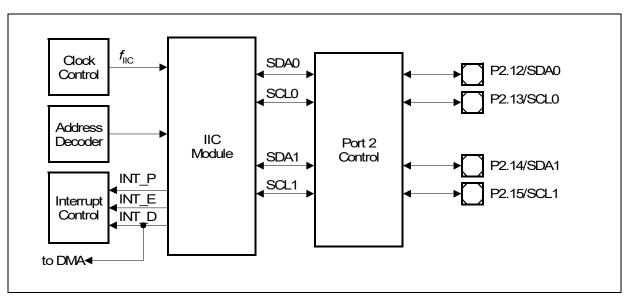


Figure 3-7 General Block Diagram of the IIC Interface

The on-chip IIC bus module connects the platform buses to other external controllers and/or peripherals via the two-line serial IIC interface. One line is responsible for clock transfer and synchronization (SCL), the other is responsible for the data transfer (SDA). The IIC bus module provides communication at data rates of up to 400 kbit/sec and features 7-bit addressing as well as 10-bit addressing. This module is fully compatible to the IIC bus protocol.

The module can operate in three different modes:

Master mode, where the IIC controls the bus transactions and provides the clock signal.

Slave mode, where an external master controls the bus transactions and provides the clock signal.

Multimaster mode, where several masters can be connected to the bus, i.e. the IIC can be master or slave.

The on-chip IIC bus module allows efficient communication via the common IIC bus. The module unloads the CPU of low level tasks such as:

- (De)Serialization of bus data
- Generation of start and stop conditions
- Monitoring the bus lines in slave mode



- · Evaluation of the device address in slave mode
- · Bus access arbitration in multimaster mode

- Extended buffer allows up to 4 send/receive data bytes to be stored
- Selectable baud rate generation
- Support of standard 100 kBaud and extended 400 kBaud data rates
- Operation in 7-bit addressing mode or 10-bit addressing mode
- · Flexible control via interrupt service routines or by polling
- Dynamic access to up to 2 physical IIC buses



3.12 Universal Serial Bus Interface (USB)

Figure 3-8 shows a global view of the functional blocks of the Universal Serial Bus interface (USB).

The USB module is further supplied with clock control, interrupt control, address decoding, and port control logic. One DMA request can be generated by USB module.

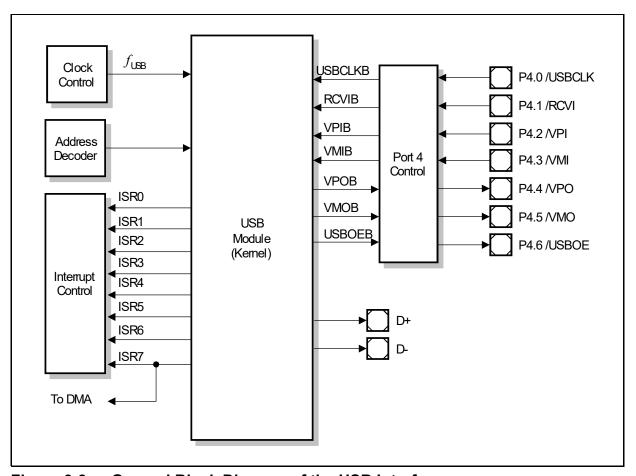


Figure 3-8 General Block Diagram of the USB Interface

The USB handles all transactions between the serial USB bus and the internal (parallel) bus of the microcontroller. The USB module includes several units which are required to support data handling with the USB bus: the on-chip USB transceiver (optionally), the flexible USB buffer block with a 32-bit wide RAM, the buffer control unit with sub modules for USB and CPU memory access control, the UDC_IF device interface for USB protocol handling, the microcontroller interface unit (MCU) with the USB specific special function registers and the interrupt generation unit. A clock generation unit provides the clock signal for the USB module for full speed and low speed USB operation.



- USB1.1 Device Standard Interface
- On-chip transceiver
- Differential I/O allow cable length up to 5m without additional hardware at target's end.
- Hot attach
- USB1.1 full speed device
- USB protocol handling in hardware
- Clock and data recovery from USB
- · Bit stripping and bit stuffing functions
- CRC5 checking, CRC16 generation and checking
- · Serial to parallel data conversion
- Maintenance of data synchronization bits (DATA0/DATA1 Toggle Bits)
- · Supports multiple configurations, interfaces and alternate settings
- 11 endpoints with user configurable endpoint information
- Flexible intermediate buffering of transmission data
- Powerful data handling capability, FIFO-support
- Back-to-back transfers fully supported by module automatism
- · Multi-packet transfer without CPU load
- Handles data transfer with minimum CPU load
- Auto increment and single address modes selectable for easy data access
- Powerful interrupt generation
- · Meets suspend power consumption restrictions in power-down mode
- Remote wakeup from USB bus activity
- Explicit support of setup information
- Enhanced status monitoring

3.13 MultiCAN

Figure 3-9 shows a global view of the functional blocks of the MultiCAN module.

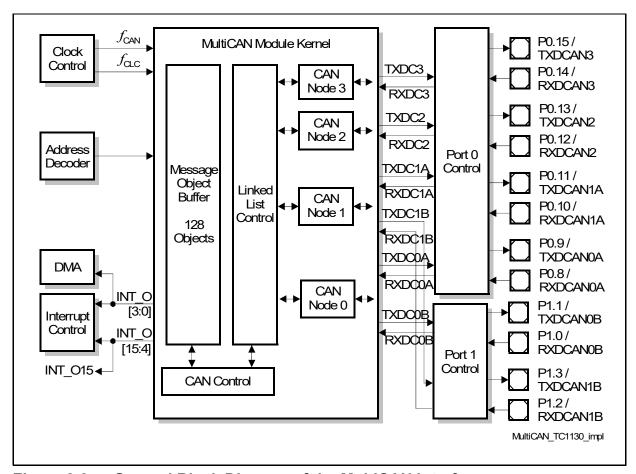


Figure 3-9 General Block Diagram of the MultiCAN Interface

The MultiCAN module contains 4 Full-CAN nodes operating independently or exchanging data and remote frames via a gateway function. Transmission and reception of CAN frames is handled in accordance to CAN specification V2.0 B (active). Each CAN node can receive and transmit standard frames with 11-bit identifiers as well as extended frames with 29-bit identifiers.

All CAN nodes share a common set of message objects, where each message object may be individually allocated to one of the CAN nodes. Besides serving as a storage container for incoming and outgoing frames, message objects may be combined to build gateways between the CAN nodes or to setup a FIFO buffer.

The message objects are organized in double chained lists, where each CAN node has its own list of message objects. A CAN node stores frames only into message objects that are allocated to the list of the CAN node. It only transmits messages from objects of this list.

A powerful, command driven list controller performs all list operations.



The bit timings for the CAN nodes are derived from the peripheral clock (f_{CAN}) and are programmable up to a data rate of 1 MBaud. A pair of receive and transmit pins connects each CAN node to a bus transceiver.

- Compliant to ISO 11898
- CAN functionality according to CAN specification V2.0 B (active)
- Dedicated control registers are provided for each CAN node
- A data transfer rate up to 1 MBaud is supported
- Flexible and powerful message transfer control and error handling capabilities are implemented
- Advanced CAN bus bit timing analysis and baud rate detection can be performed for each CAN node via the frame counter
- Full-CAN functionality: A set of 128 message objects can be individually
 - allocated (assigned) to any CAN node
 - configured as transmit or receive object
 - setup to handle frames with 11-bit or 29-bit identifier
 - counted or assigned a timestamp via a frame counter
 - configured to remote monitoring mode
- Advanced Acceptance Filtering:
 - Each message object provides an individual acceptance mask to filter incoming frames.
 - A message object can be configured to accept only standard or only extended frames or to accept both standard and extended frames.
 - Message objects can be grouped into 4 priority classes.
 - The selection of the message to be transmitted first can be performed on the basis of frame identifier, IDE bit and RTR bit according to CAN arbitration rules.
- Advanced Message Object Functionality:
 - Message objects can be combined to build FIFO message buffers of arbitrary size, which is only limited by the total number of message objects.
 - Message objects can be linked to form a gateway to automatically transfer frames between two different CAN buses. A single gateway can link any two CAN nodes. An arbitrary number of gateways may be defined.
- Advanced Data Management:
 - The Message objects are organized in double chained lists.
 - List reorganizations may be performed at any time, even during full operation of the CAN nodes.
 - A powerful, command driven list controller manages the organization of the list structure and ensures consistency of the list.
 - Message FIFOs are based on the list structure and can easily be scaled in size during CAN operation.



- Static Allocation Commands offer compatibility with TwinCAN applications, which are not list based.
- Advanced Interrupt Handling:
 - Up to 16 interrupt output lines are available. Most interrupt requests can be individually routed to one of the 16 interrupt output lines.
 - Message postprocessing notifications can be flexibly aggregated into a dedicated register field of 256 notification bits.



3.14 Micro Link Serial Bus Interface (MLI)

Figure 3-10 shows a global view of the functional blocks of two Micro Link Serial Bus interfaces (MLI0 and MLI1).

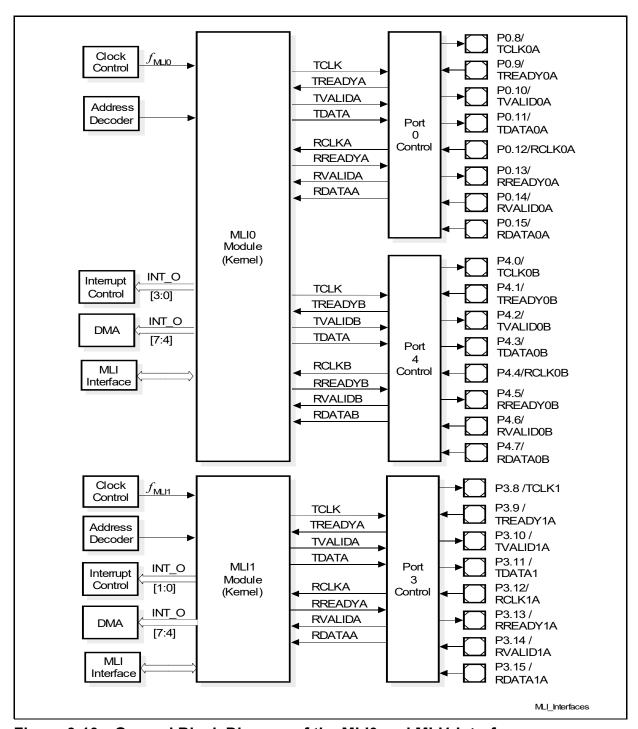


Figure 3-10 General Block Diagram of the MLI0 and MLI1 Interfaces



The Micro Link Serial Bus Interface is dedicated to the serial communication between the other Infineon 32-bit controllers with MLI. The communication is intended to be fast due to an address translation system, and it is not necessary to have any special program in the second controller.

- Serial communication from the MLI transmitter to MLI receiver of another controller
- Module supports connection of each MLI with up to four MLI from other controllers
- Fully transparent read/write access supported (= remote programming)
- Complete address range of target controller available
- Special protocol to transfer data, address offset, or address offset and data
- Error control using a parity bit
- 32-bit, 16-bit, and 8-bit data transfers
- · Address offset width: from 1- to 16-bit
- Baud rate: f_{MLI} / 2 (symmetric shift clock approach), baud rate definition by the corresponding fractional divider



3.15 General Purpose Timer Unit (GPTU)

Figure 3-11 shows a global view of the functional blocks of the General Purpose Timer Unit (GPTU).

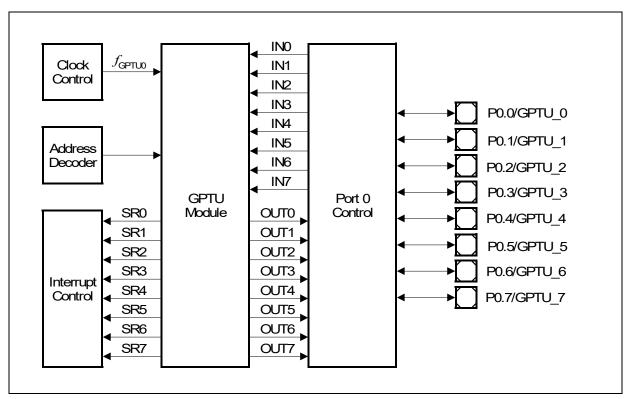


Figure 3-11 General Block Diagram of the GPTU Interface

The GPTU consists of three 32-bit timers designed to solve such application tasks as event timing, event counting, and event recording. The GPTU communicates with the external world via eight I/O lines located at Port 0.

The three timers of GPTU module, T0, T1 and T2, can operate independently of each other or can be combined:

General Features:

- All timers are 32-bit precision timers with a maximum input frequency of f_{GPTU}
- Events generated in T0 or T1 can be used to trigger actions in T2
- Timer overflow or underflow in T2 can be used to clock either T0 or T1
- T0 and T1 can be concatenated to form one 64-bit timer

Features of T0 and T1:

- Each timer has a dedicated 32-bit reload register with automatic reload on overflow
- Timers can be split into individual 8-, 16-, or 24-bit timers with individual reload registers



- Overflow signals can be selected to generate service requests, pin output signals, and T2 trigger events
- Two input pins can define a count option

Features of T2:

- Count up or down is selectable
- Operating modes:
 - Timer
 - Counter
 - Quadrature counter (incremental/phase encoded counter interface)
- Options:
 - External start/stop, one-shot operation, timer clear on external event
 - Count direction control through software or an external event
 - Two 32-bit reload/capture registers
- · Reload modes:
 - Reload on overflow or underflow
 - Reload on external event: positive transition, negative transition, or both transitions
- Capture modes:
 - Capture on external event: positive transition, negative transition, or both transitions
 - Capture and clear timer on external event: positive transition, negative transition, or both transitions
- Can be split into two 16-bit counter/timers
- Timer count, reload, capture, and trigger functions can be assigned to input pins. To and T1 overflow events can also be assigned to these functions.
- Overflow and underflow signals can be used to trigger T0 and/or T1 and to toggle output pins
- T2 events are freely assignable to the service request nodes



3.16 Capture/Compare Unit 6 (CCU6)

Figure 3-12 shows a global view of the functional blocks of two Capture/Compare Units (CCU60 and CCU61).

Both of the CCU6 modules are further supplied with clock control, interrupt control, address decoding, and port control logic. One DMA request can be generated by each CCU6 module.

Each CCU6 provides two independent timers (T12, T13), which can be used for PWM generation, especially for AC-motor control. Additionally, special control modes for block commutation and multi-phase machines are supported.

Timer 12 Features:

- Three capture/compare channels, each channel can be used either as capture or as compare channel
- Generation of a three-phase PWM supported (six outputs, individual signals for highside and lowside switches)
- 16-bit resolution, maximum count frequency = peripheral clock
- Dead-time control for each channel to avoid short-circuits in the power stage
- Concurrent update of the required T12/13 registers
- Center-aligned and edge-aligned PWM can be generated
- Single-shot mode supported
- · Many interrupt request sources
- · Hysteresis-like control mode

Timer 13 Features:

- One independent compare channel with one output
- 16-bit resolution, maximum count frequency = peripheral clock
- Can be synchronized to T12
- Interrupt generation at period-match and compare-match
- Single-shot mode supported

Additional Features:

- Block commutation for Brushless DC-drives implemented
- · Position detection via Hall-sensor pattern
- Automatic rotational speed measurement for block commutation
- Integrated error handling
- Fast emergency stop without CPU load via external signal (CTRAP)
- Control modes for multi-channel AC-drives
- Output levels can be selected and adapted to the power stage



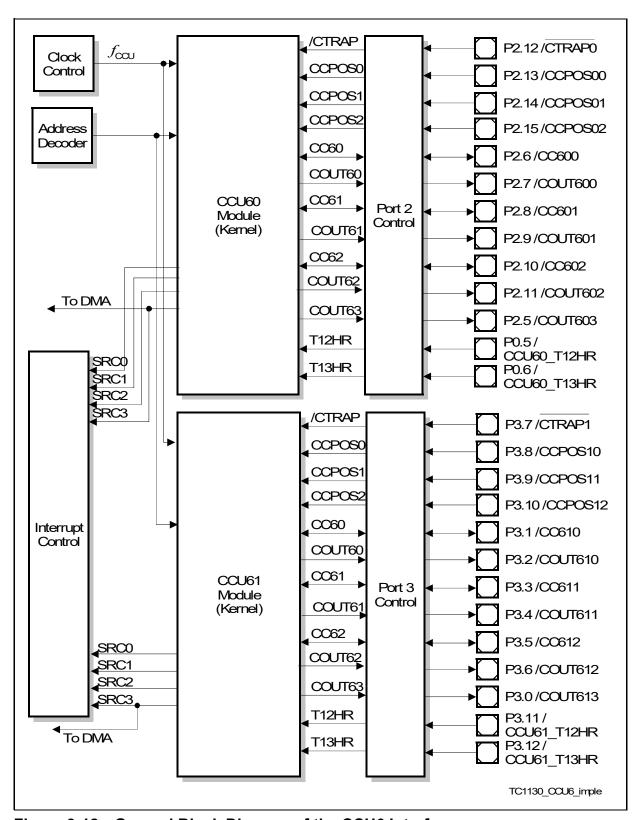


Figure 3-12 General Block Diagram of the CCU6 Interfaces



3.17 Ethernet Controller

The MAC controller implements the IEEE 802.3 and operates either at 100 Mbit/sec or 10 Mbit/sec. Figure 3-13 shows a global view of the Ethernet Controller module with the module specific interface connections.

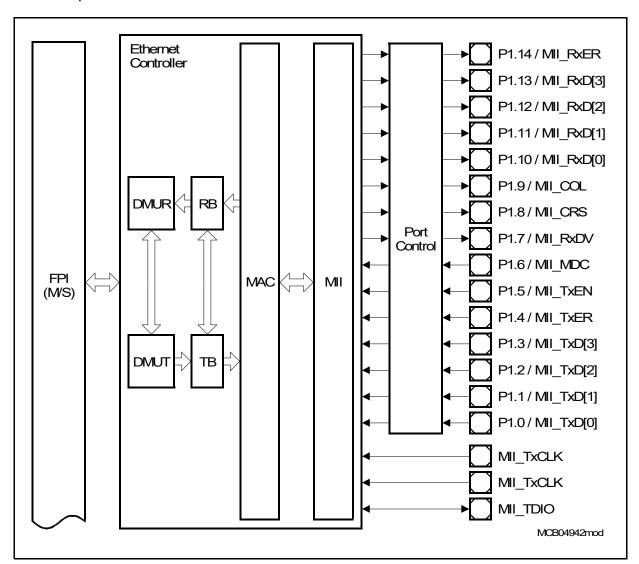


Figure 3-13 General Block Diagram of the Ethernet Controller

The Ethernet controller comprises the following functional blocks:

- Media Access Controller (MAC)
- · Receive Buffer (RB)
- Transmit Buffer (TB)
- Data Management Unit in Receive Direction (DMUR)
- Data Management Unit in Transmit Direction (DMUT)



RB and TB provide on-chip data buffering whereas DMUR and DMUT perform data transfer from/to the shared memory.

Two interfaces are provided by the Ethernet Controller module:

- MII interface for connection of Ethernet PHYs via 18 Input/Output lines
- Master/slave FPI bus interface for connection to the on-chip system bus for data transfer as well as configuration

Features:

- Media Independent Interface (MII) according to IEEE 802.3
- Supports 10 or 100 Mbit/sec MII-based Physical devices
- Supports Full Duplex Ethernet
- Supports data transfer between Ethernet Controller and COM-DRAM
- Supports data transfer between Ethernet Controller and SDRAM via EBU
- 256 x 32 bit Receive buffer and Transmit buffer each
- Supports burst transfers up to 8 x 32 Bytes

Media Access Controller (MAC)

- 100/10 Mbit/sec operations
- Full IEEE 802.3 compliance
- Station management signaling
- Large on-chip CAM (Content Addressable Memory)
- Full duplex mode
- 80-byte transmit FIFO
- 16-byte receive FIFO
- PAUSE Operation
- Flexible MAC Control Support
- Supports Long Packet mode and Short Packet mode
- PAD generation

Media Independent Interface (MII)

- Media independence
- · Multi-vendor point of interoperability
- Supports connection of MAC layer and Physical (PHY) layer devices
- Capable of supporting both 100 Mbit/sec and 10 Mbit/sec data rates
- Data and delimiters are synchronous to clock references
- Provides independent four bits wide transmit and receive data paths
- Supports connection of PHY layer and Station Management (STA) devices
- · Provides a simple management interface
- Capable of driving a limited length of shielded cable



3.18 System Timer

The STM within the TC1130 is designed for global system timing applications requiring both high precision and long range. The STM provides the following features:

- Free-running 56-bit counter
- All 56 bits can be read synchronously
- Different 32-bit portions of the 56-bit counter can be read synchronously
- Flexible interrupt generation on partial STM content compare match
- Driven by clock f_{STM} after reset (default after reset is $f_{STM} = f_{SYS} = 150 \text{ MHz}$)
- Counting starts automatically after a reset operation
- STM is reset under following reset causes:
 - Wake-up reset (PMG CON.DSRW must be set)
 - Software reset (RST_REQ.RRSTM must be set)
 - Power-on reset
- STM (and the clock divider) is not reset at watchdog reset and hardware reset (HDRST = 0)

The STM is an upward counter, running with the system clock frequency $f_{\rm SYS}$ (after reset $f_{\rm STM} = f_{\rm SYS}$). It is enabled per default after reset, and immediately starts counting up. Other than via reset, it is not possible to affect the contents of the timer during normal operation of the application; it can only be read, but not written to. Depending on the implementation of the clock control of the STM, the timer can optionally be disabled or suspended for power-saving and debugging purposes via a clock control register.

The maximum clock period is $2^{56}/f_{STM}$. At f_{STM} = 150 MHz (maximum), for example, the STM counts 15.2 years before overflowing. Thus, it is capable of continuously timing the entire expected product lifetime of a system without overflowing.



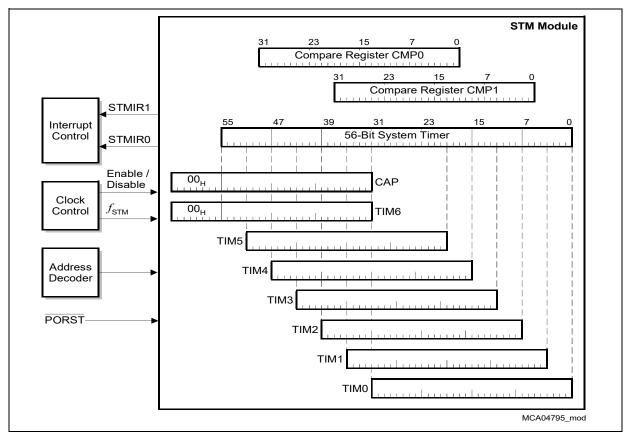


Figure 3-14 Block Diagram of the STM Module



3.19 Watchdog Timer

The Watchdog Timer (WDT) provides a highly reliable and secure way to detect and recover from software or hardware failure. The WDT helps to abort an accidental malfunction of the TC1130 in a user-specified time period. When enabled, the WDT will cause the TC1130 system to be reset if the WDT is not serviced within a user-programmable time period. The CPU must service the WDT within this time interval to prevent the WDT from causing a TC1130 system reset. Hence, routine service of the WDT confirms that the system is functioning properly.

In addition to this standard "Watchdog" function, the WDT incorporates the ENDINIT feature and monitors its modifications. A system-wide line is connected to the ENDINIT bit implemented in a WDT control register, serving as an additional write-protection for critical registers (besides supervisor mode protection). Registers protected via this line can be modified only when supervisor mode is active and bit ENDINIT = 0.

A further enhancement in the TC1130's Watchdog Timer is its reset prewarning operation. Instead of immediately resetting the device upon detection of an error, the WDT first issues a Non-Maskable Interrupt (NMI) to the CPU before finally resetting the device at a specified time period later. This gives the CPU a chance to save system state to memory for later examination of the cause of the malfunction, thus providing an important aid in debugging.

- 16-bit Watchdog counter
- Selectable input frequency: $f_{SYS}/256$ or $f_{SYS}/16384$
- 16-bit user-definable reload value for normal Watchdog operation, fixed reload value for time-out and prewarning modes
- Incorporation of the ENDINIT bit and monitoring of its modifications
- Sophisticated password access mechanism with fixed and user-definable password fields
- Proper access always requires two write accesses. The time between the two accesses is monitored by the WDT.
- Access Error Detection: Invalid password (during first access) or invalid guard bits (during second access) trigger the Watchdog reset generation.
- Overflow Error Detection: An overflow of the counter triggers the Watchdog reset generation.
- Watchdog function can be disabled; access protection and ENDINIT monitor function remain enabled.
- Double Reset Detection: If a Watchdog induced reset occurs twice without a proper access to its control register in between, a severe system malfunction is assumed and the TC1130 is held in reset until a power-on reset or a hardware reset occurs. This prevents the device from being periodically reset if, for instance, connection to the



- external memory has been lost such that even system initialization could not be performed.
- Important debugging support is provided through the reset prewarning operation by first issuing an NMI to the CPU before finally resetting the device after a certain period of time.

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3.20 System Control Unit

The System Control Unit (SCU) of the TC1130 handles the system control tasks. All of these system functions are tightly coupled; thus, they are conveniently handled by one unit, the SCU. The system tasks of the SCU are:

- Clock Control
 - Clock generation
 - Oscillator and PLL control
- Reset and Boot Control
 - Generation of all internal reset signals
 - Generation of external hardware and software reset signal
- Power Management Control
 - Enabling of several power management modes
- · Configuration input sampling
- Ethernet Interrupts
- FPU interrupts
- External Request Unit
- Parity Error Control
- Fault SRAM Fuse Box
- CSCOMB Control
- EBU Pull-Up Control
- · NMI Control and Status
- DMA Request Signal Selection



3.21 Boot Options

The TC1130 booting schemes provides a number of different boot options for the start of code execution. **Table 3-3** shows the boot options available in the TC1130.

Table 3-3 Boot Selections

BRKIN ¹⁾	TM ¹⁾	HWCFG [2:0]	Type of Boot	PC Start Value (User Entry)
1	1	000	Bootstrap Loader Serial boot from ASC to PMI scratch pad, run loaded program	DFFF FFFC _H ²⁾ (D400 0000 _H)
		001	Bootstrap Loader Serial boot from CAN to PMI scratch pad, run loaded program	
		010	Bootstrap Loader Serial boot from SSC to PMI scratch pad, run loaded program	
		011	External memory, EBU as master	DFFF FFFC _H ²⁾ (A000 0000 _H)
		100	External memory, EBU as slave	DFFF FFFC _H ²⁾ (A000 0000 _H)
		101	Reserved (STOP)	
		110	PMI scratch pad	D400 0000 _H
		111	Reserved (STOP)	DFFF FFFC _H ²⁾
1	0	000-111	Reserved (STOP)	DFFF FFFC _H ²⁾
0	1	000	Tristate chip	
		001	Go to external emulator space	DFFF FFFC _H ²⁾ (DE00 0000 _H)
		010	Reserved (STOP)	
		011	OSC and PLL Bypass	
		100-111	Reserved (STOP)	DFFFFFFC _H ²⁾
0	0	000-111	Reserved (STOP)	DFFFFFC _H ²⁾

¹⁾ This input signal is active low.

²⁾ This is the BootROM entry address; the start address of user program in parentheses.



3.22 Power Management System

The TC1130 power management system allows software to configure the various processing units to adjust automatically in order to draw the minimum necessary power for the application.

There are four power management modes:

- Run Mode
- Idle Mode
- Sleep Mode
- Deep Sleep Mode

Table 3-4 describes the features of the power management modes.

Table 3-4 Power Management Mode Summary

Mode	Description	
Run	The system is fully operational. All clocks and peripherals are enabled, as determined by software.	
Idle	The CPU clock is disabled, waiting for a condition to return it to run mode. Idle mode can be entered by software when the processor has no active tasks to perform. All peripherals remain powered and clocked. Processor memory is accessible to peripherals. A reset, Watchdog Timer event, a falling edge on the NMI pin, or any enabled interrupt event will return the system to run mode.	
Sleep	The system clock continues to be distributed only to those peripherals programmed to operate in sleep mode. The other peripheral modules will be shut down by the suspend signal. Interrupts from operating peripherals, the Watchdog Timer, a falling edge on the NMI pin, or a reset event will return the system to run mode. Entering this state requires an orderly shut-down controlled by the Power Management State Machine.	
Deep Sleep	The system clock is shut off; only an external signal will restart the system. Entering this state requires an orderly shut-down controlled by the Power Management State Machine (PMSM).	

Besides these explicit software-controlled power-saving modes, special attention has been paid in the TC1130 to automatic power-saving in operating units that are currently not required or idle. In this case, they are shut off automatically until their operation is required again.



3.23 On-Chip Debug Support

The On-Chip Debug Support of the TC1130 consists of the following building blocks:

- OCDS L1 module of TriCore™
- OCDS L2 interface of TriCore™
- OCDS L1 module in the BCU of the FPI Bus
- OCDS L1 facilities within the DMA
- OCDS L2 interface of DMA
- OCDS System Control Unit (OSCU)
- Multi Core Break Switch (MCBS)
- JTAG based Debug Interface (Cerberus JDI)
- Suspend functionality of peripherals

Features:

- TriCore™ L1 OCDS:
 - Hardware event generation unit
 - Break by DEBUG instruction or break signal
 - Full Single-Step support in hardware, possible also with software break
 - Access to memory, SFRs, etc. on the fly
- DMA L1 OCDS:
 - Output break request on errors
 - Suspension of pre-selected channels
- Level 2 trace port with 16 pins that outputs either TriCore[™], or DMA trace
- OCDS System Control Unit (Cerberus OSCU)
 - Minimum number of pins required (no OCDS enable pin)
 - Hardware allows hot attach of a debugger to a running system.
 - System is secure (can be locked from internal)
- Multi Core Break Switch (Cerberus MCBS):
 - TriCore™, DMA, break pins, and BCUs as break sources
 - TriCore[™] as break targets; other parts can in addition be suspended
 - Synchronous stop and restart of the system
 - Break to Suspend converter

Figure 3-15 shows a basic block diagram of the building blocks.



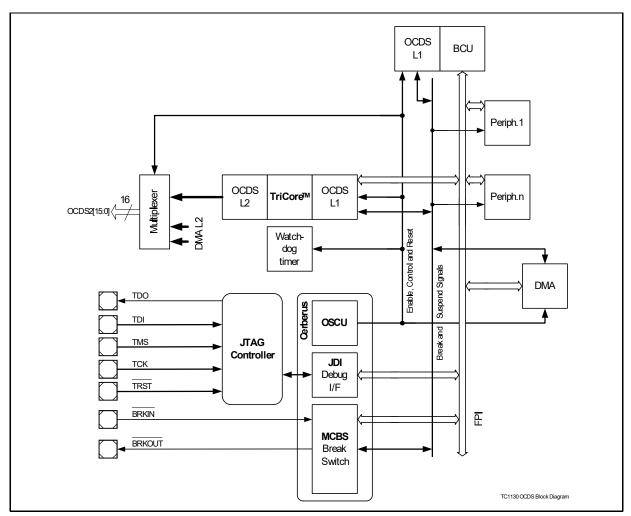


Figure 3-15 OCDS Support Basic Block Diagram



3.24 Clock Generation Unit

The Clock Generation Unit (CGU) allows a flexible clock generation for TC1130. The power consumption is indirectly proportional to the frequency, whereas the performance of the microcontroller is directly proportional to the frequency. During user program execution, the frequency can be programmed for an optimal ratio between performance and power consumption. Therefore, the power consumption can be adapted to the actual application state.

Features:

The Clock Generation Unit serves several purposes:

- · PLL feature for multiplying clock source by different factors
- Direct Drive for direct clock input
- Comfortable state machine for secure switching between basic PLL, direct, or prescaler operation
- Sleep and Power-down mode support
- USB Clock source and control

The Clock Generation Unit in the TC1130, shown in **Figure 3-16**, consists of an oscillator circuit and one Phase-Locked Loop (PLL). The PLL can convert a low-frequency external clock signal to a high-speed internal clock for maximum performance. The PLL also has fail-safe logic that detects degenerate external clock behavior such as abnormal frequency deviations or a total loss of the external clock. It can execute emergency actions if it looses the lock on the external clock.

In general, the Clock Generation Unit (CGU) is controlled through the System Control Unit (SCU) module of the TC1130.

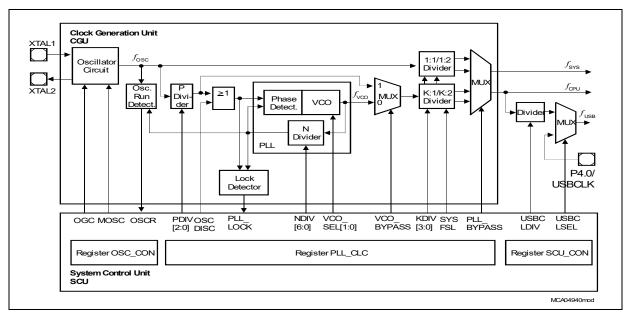


Figure 3-16 Clock Generation Unit Block Diagram



The oscillator circuit, which is designed to work with an external crystal oscillator or an external stable clock source, consists of an inverting amplifier with XTAL1 as input and XTAL2 as output.

Figure 3-17 shows the recommended external oscillator circuitries for both operating modes, i.e. external crystal mode and external input clock mode.

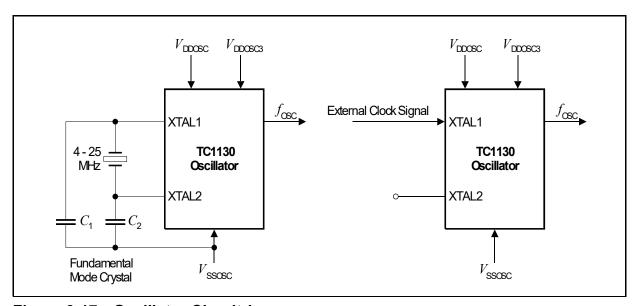


Figure 3-17 Oscillator Circuitries

When using an external clock signal, it must be connected to XTAL1 and XTAL2 is left open (unconnected). When supplying the clock signal directly, not using a crystal and the oscillator, the input frequency can be in the range of 0 - 40 MHz if the PLL is not used, 4 - 40 MHz in case the PLL is used.

When using a crystal, its frequency can be within the range of 4 MHz to 25 MHz. An external oscillator load circuitry must be used, connected to both pins, XTAL1 and XTAL2. It consists normally of the two load capacitances, C1 and C2. For some crystals, a series damp resistor may be necessary. The exact values and related operating range are dependant on the crystal and have to be determined and optimized together with the crystal vendor using the negative resistance method. As starting point for the evaluation and for non-productive systems, the following load capacitor values might be used.

Table 3-5 Load Capacitors Select

Fundamental Mode Crystal Frequency (approx., MHz)	Load Capacitors C1, C2 (pF)
4	33
8	18
12	12
16	10



Table 3-5 Load Capacitors Select (cont'd)

Fundamental Mode Crystal Frequency (approx., MHz)	Load Capacitors C1, C2 (pF)
20	10
24	10

A block capacitor between $V_{\rm DDOSC3}$ and $V_{\rm SSOSC}$, $V_{\rm DDOSC}$ and $V_{\rm SSOSC}$ is recommended, too.



3.25 Power Supply

The TC1130 provides an ingenious power supply concept in order to improve the EMI behavior as well as to minimize the crosstalk within on-chip modules.

Figure 3-18 shows the TC1130's power supply concept, where certain logic modules are individually supplied with power. This concept improves the EMI behavior by reduction of the noise cross coupling.

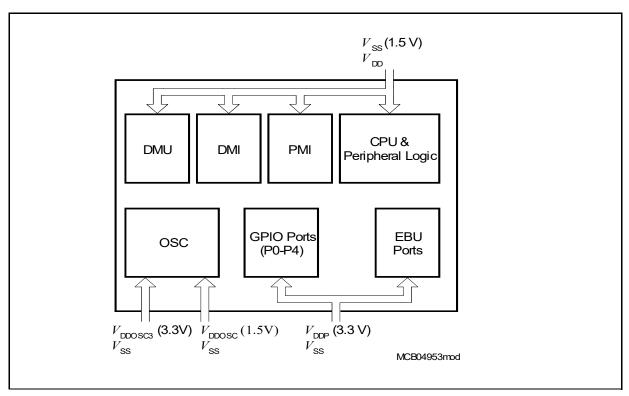


Figure 3-18 TC1130 Power Supply Concept



3.26 Power Sequencing

During power-up, reset pin PORST has to be held active until both power supply voltages have reached at least their minimum values.

During the power-up time (rising of the supply voltages from 0 to their regular operating values), it must be ensured that the core V_{DD} power supply reaches its operating value first, and then followed by the GPIO V_{DDP} power supply. During the rising time of the core voltage, it must be ensured that $0 < V_{DD} - V_{DDP} < 0.5 \text{ V}$.

During power-down, the core power supply V_{DD} and GPIO power supply V_{DDP} must be switched off completely until all capacitances are discharged to zero before the next power-up.

Note: The state of the pins are undefined when only the port voltage V_{DDP} is switched on.



3.27 Identification Register Values

Table 3-6 TC1130 Identification Registers

Short Name	Address	Value
SCU ID	F000 0008 _H	002C C001 _H
MANID	F000 0070 _H	0000 1820 _H
CHIPID	F000 0074 _H	0000 8C01 _H
RTID	F000 0078 _H	0000 0000 _H
SBCU_ID	F000 0108 _H	0000 6A0A _H
STM_ID	F000 0208 _H	0000 C005 _H
CBS_JDPID	F000 0308 _H	0000 6307 _H
GPTU_ID	F000 0608 _H	0001 C002 _H
CCU60_ID	F000 2008 _H	0042 C004 _H
CCU61_ID	F000 2108 _H	0042 C004 _H
DMA_ID	F000 3C08 _H	001A C011 _H
CAN_ID	F000 4008 _H	002B C022 _H
USB_ID	F00E 2808 _H	0000 4A00 _H
SSC0_ID	F010 0108 _H	0000 4530 _H
SSC1_ID	F010 0208 _H	0000 4530 _H
ASC0_ID	F010 0308 _H	0000 44E2 _H
ASC1_ID	F010 0408 _H	0000 44E2 _H
ASC2_ID	F010 0508 _H	0000 44E2 _H
IIC_ID	F010 0608 _H	0000 4604 _H
MLI0_ID	F010 C008 _H	0025 C004 _H
MLI1_ID	F010 C108 _H	0025 C004 _H
MCHK_ID	F010 C208 _H	001B C001 _H
CPS_ID	F7E0 FF08 _H	0015 C006 _H
MMU_ID	F7E1 8008 _H	0009 C002 _H
CPU_ID	F7E1 FE18 _H	000A C005 _H
EBU_ID	F800 0008 _H	0014 C004 _H
DMU_ID	F800 0408 _H	002D C001 _H
DMI_ID	F87F FC08 _H	0008 C004 _H
PMI_ID	F87F FD08 _H	000B C004 _H



Table 3-6 TC1130 Identification Registers (cont'd)

Short Name	Address	Value
LBCU_ID	F87F FE08 _H	000F C005 _H
LFI_ID	F87F FF08 _H	000C C005 _H



4 Electrical Parameters

4.1 General Parameters

4.1.1 Parameter Interpretation

The parameters listed in this section represent partly the characteristics of the TC1130 and partly its requirements on the system. To aid interpreting the parameters easily when evaluating them for design purposes, they are indicated by the abbreviations in the "Symbol" column:

· cc

These parameters indicate **C**ontroller **C**haracteristics, which are distinctive features of the TC1130 and must be considered for system design.

· SR

These parameters indicate **S**ystem **R**equirements, which must be provided by the microcontroller system in which the TC1130 is included.



4.1.2 Absolute Maximum Rating

Parameter	Symbol	Lim	it Values	Unit	Notes
		min.	max.		
Ambient temperature	T_{A}	-40	85	°C	under bias
Storage temperature	T_{ST}	-65	150	°C	_
Junction temperature	T_{J}	-40	125	°C	under bias
Voltage at 1.5 V power supply pins with respect to $V_{\rm SS}^{-1}$	V_{DD}	-0.5	1.7	V	_
Voltage at 3.3 V power supply pins with respect to $V_{\rm SS}^{2)}$	V_{DDP}	-0.5	4.0	V	_
Voltage on any pin with respect to $V_{\rm SS}^{2)}$	V_{IN}	-0.5	4.0	V	_
Input current on any pin during overload condition	I_{IN}	-10	10	mA	_
Absolute sum of all input currents during overload condition	ΣI_{IN}	_	100	mA	_
CPU & LMB Bus Frequency	$f_{\sf SYS}$	_	150	MHz	_
FPI Bus Frequency	f_{FPI}	_	100	MHz	_

¹⁾ Applicable for $V_{\rm DD}$ and $V_{\rm DDOSC}$.

²⁾ Applicable for $V_{\rm DDP}$ and $V_{\rm DDOSC3}$. The maximum voltage difference must not exceed 4.0 V in any case (i.e. Supply Voltage = 4.0 V and Input Voltage = -0.5 V is not allowed).



4.1.3 Operating Condition

The following operating conditions must be complied with in order to ensure correct operation of the TC1130. All parameters specified in the following table refer to these operating conditions, unless otherwise indicated.

Parameter	Symbol	Limi	it Values	Unit	Notes	
		min.	max.		Conditions	
Digital supply voltage	V_{DD}	1.43	1.58	V	_	
	V_{DDP}	3.14	3.47	V	_	
Digital ground voltage	V_{SS}		0	V	_	
Digital core supply current	I_{DD}	_	525	mA	_	
Ambient temperature under bias	T_{A}	-40	+85	°C	_	
CPU clock	$f_{\sf SYS}$	_1)	150	MHz	_	
Overload current	I_{OV}	-1	1	mA	2)3)	
		-3	3		duty cycle ≤ 25%	
Short circuit current	I_{SC}	-1	1	mA	4)	
		-3	3		duty cycle ≤ 25%	
Absolute sum of overload +	$\Sigma I_{OV} $ +	_	[50]	mA	3)	
short circuit currents	$ I_{SC} $		100		duty cycle ≤ 25%	
Inactive device pin current (V _{DD} = V _{DDP} = 0)	$I_{ m ID}$	-1	1	mA	_	
External load capacitance	C_{L}	-	50	pF	_	
ESD strength	_	2000	_	V	Human Body Model (HBM)	

¹⁾ The TC1130 uses a static design, so the minimum operation frequency is 0 MHz. However, due to test time restriction no lower frequency boundary is tested.

²⁾ Overload conditions occur if the standard operating conditions are exceeded, i.e. the voltage on any pin exceeds the specified range (i.e. $V_{\text{OV}} > V_{\text{DDP}} + 0.5 \,\text{V}$ or $V_{\text{OV}} < V_{\text{SS}}$ - 0.5 V). The absolute sum of input overload currents on all digital I/O pins may not exceed **50 mA**. The supply voltage must remain within the specified limits.

³⁾ Not subject to production test, verified by design/characterization.

⁴⁾ Applicable for digital inputs.



4.2 DC Parameters

4.2.1 Input/Output Characteristics

 Table 4-1
 Input/Output DC-Characteristics (Operating Conditions apply)

Parameter	Symbol	Limit	Values	Unit	Test Condition
		min.	max.		
GPIO pins, Dedicated pin	s and EBU p	ins		•	
Input low voltage	V_{IL} SR	-0.3	0.8	V	LvTTL
Input high voltage	V _{IH} SR	2.0	V _{DDP} + 0.3	V	LvTTL
Output low voltage	V_{OL} CC	_	0.4	V	I _{OL} = 2mA
Output high voltage	V _{OH} CC	2.4	_	V	I _{OH} = -2mA
Pull-up current 1)	$ I_{PUA} CC$	_	149	μΑ	V _{IN} = 0V
	$ I_{PUC} CC$	_	7.2	μΑ	$V_{IN} = OV$
Pull-down current ²⁾	$ I_{PDA} CC$	_	156	μΑ	$V_{IN} = V_{DDP}$
	$ I_{PDC} CC$	_	15.7	μΑ	$V_{IN} = V_{DDP}$
Input leakage current 3)	I _{OZ1} CC	_	±350	nA	$0 < V_{\text{IN}} < V_{\text{DDP}}$
Pin Capacitance ⁴⁾	C_{IO} CC	_	10	pF	f = 1 MHz T _A = 25 °C

¹⁾ The current is applicable to the pins, for which a pull-up has been specified. Refer to **Table 2-1**. I_{PUX} refers to the pull-up current for type x in absolute values.

²⁾ The current is applicable to the pins, for which a pull-down has been specified. Refer to **Table 2-1**. I_{PDX} refers to the pull-down current for type x in absolute values.

³⁾ Excluded following pins: NMI, TRST, TCK, TDI, TMS, MII_TXCLK, MII_RXCLK, MII_MDIO, ALE, P2.1,HWCFG0, HWCFG1, HWCFG2, BRKIN, PORST, HDRST.

⁴⁾ Not subject to production test, verified by design/characterization.



4.2.2 Oscillator Characteristics

 Table 4-2
 Oscillator Pins Characteristics (Operating Conditions apply)

Parameter	Symbol Limit Values		Values	Unit	Test Condition
		min.	max.	1	
Oscillator Pins				'	
Input low voltage at XTAL1	V_{ILX} SR	-0.3	_	V	1)
Input high voltage at XTAL1	V_{IHX} SR	_	3	V	1)
Quartz oscillation peak-peak amplitude at oscillator Input	$V_{ m PPOSC}$ SR	0.6	_	V	1)
Input low voltage at XTAL1	V_{ILX} SR	-0.3	0.1	V	2)
Input high voltage at XTAL1	V _{IHX} SR	1.4	V _{DDC} + 0.3V	V	2)
Oscillator input current	I_{OSCIN}	_	25	μΑ	-

¹⁾ Quartz mode: using a quartz crystal

²⁾ Bypass mode: using an external clock



4.2.3 USB Characteristics

4.2.3.1 DC Electrical Characteristics

 Table 4-3
 USB Electrical Characteristics (Operating Conditions apply)

Parameter	Symbol	Li	mit Val	/alues Unit	Unit	Test Conditions
		min.	typ.	max.		
Input Level			•		•	
Differential Input Level	V_{DI}	0.2	_	_	V	V(D+) - V(D-)
Differential Common Mode Range	V_{DCM}	0.8	-	2.5	V	Range of Sensitivity
Single Ended Receiver Threshold	V_{SE}	<i>low</i> < 0.8	_	high > 2.0	V	_
Output Levels		•				
Static Output Low	V_{SOL}	_	_	< 0.3	V	with 1.5 kΩ to 3.6 V
Static Output High	V_{SOH}	2.8	3.3	3.6	V	with 15 $k\Omega$ to ground
Leakage Current		•	•	•	•	•
Hi_Z State Data Line Leakage	V_{HIZ}	-10	-	10	μА	0 < V _{in} < 3.3 V

4.2.3.2 Full Speed Electrical Characteristics

Table 4-4 Full Speed Characteristics (Operating Conditions apply)

Parameter	rameter Symbol Limit Values I		Unit	Test Conditions		
		min.	typ.	max.		
Driver Characteristic	S		_			
Rise/Fall Time	t _{RF}	4	_	20	ns	Capacitive load 50 pF
Rise/Fall Time Matching	t _{RFM}	90	100	110	%	Capacitive load 50 pF
Crossover Voltage of Differential Signals	V_{CV}	1.3	_	2.0	V	Capacitive load 50 pF



Parameter	Symbol	Limit Values		Unit	Test Conditions	
		min.	typ.	max.		
Driver Output Impedance	R_{DO}	28	_	44	Ω	Steady State Driver
Termination Impedance	R_{T}	1.425	1.5	1.575	kΩ	_

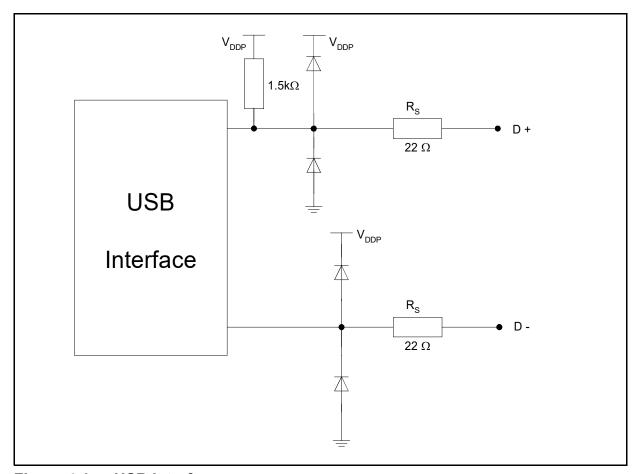


Figure 4-1 USB Interface



4.2.4 IIC Characteristics

Each IIC Pin is an open drain output pin with different characteristics than other pins. The related characteristics are given in the following table.

Table 4-5 IIC Pin Characteristics (Operating Conditions Apply)

Parameter	Sym	bol	Limit Valu	es	Unit	Test Conditions
			min.	max.		
Output low voltage	V _{OL}	CC	-	0.4	V	3 mA sink current 6 mA sink current
Input high voltage ¹⁾	V _{IH}	SR	0.7V _{DDP}	V _{DDP} +0.5	V	_
Input low voltage ¹⁾	V _{IL}	SR	-0.5	0.3V _{DDP}	V	_

¹⁾ Not subject to production test, verified by design/characterization.

Note: No 5 V IIC interface is supported with these pads. Only voltages lower than 3.63 V must be applied to these pads.

Note: IIC pins have no pull-up and pull-down devices.



4.2.5 Power Supply Current

Table 4-6 Power Supply Currents (Operating Conditions apply)

Parameter	Symbol	Limit	Limit Values		Limit Values		Test Conditions
		typ. 1)	max.				
Active mode supply	I_{DD}	314	679	mA	Sum of $I_{\rm DDS}^{-2)}$		
current		153	345	mA	$I_{\rm DD}$ at $V_{\rm DD}$ ³⁾		
		156	322	mA	I_{DD} at V_{DDP}		
Idle mode supply current	I_{ID}	74	154	mA	Sum of $I_{\rm DDS}^{2)4)}$		
		66	130	mA	$I_{\rm DD}$ at $V_{\rm DD}^{3)4)}$		
		6	15	mA	$I_{\rm DD}$ at $V_{\rm DDP}^{4)}$		
Deep sleep mode supply	I_{DS}	2	19	mA	Sum of $I_{\rm DDS}^{2)5)}$		
current		2	19	mA	$I_{\rm DD}$ at $V_{\rm DD}^{3)5)}$		
		3.6	58	μΑ	$I_{\rm DD}$ at $V_{\rm DDP}^{5)}$		

Typical values are measured at 25°C, CPU clock at 150 MHz, and nominal supply voltage that is 3.3 V for V_{DDP}, V_{DDOSC3} and 1.5 V for V_{DD}, V_{DDOSC}. These currents are measured using a typical application pattern. The power consumption of modules can increase or decrease using other application programs.

These power supply currents are defined as the sum of all currents at the V_{DD} power supply lines: $V_{DD} + V_{DDD} + V_{DDOSC3} + V_{DDOSC}$

 $^{^{\}rm 3)}$ This measurement includes the TriCore $^{\rm TM}$ and Logic power supply lines.

⁴⁾ CPU is in idle state, input clocks to all peripherals are enabled.

⁵⁾ Clock generation is disabled at the source.



4.3 AC Parameters

4.3.1 Power, Pad and Reset Timing

Parameter		bol	Limit	Unit	
			min.	max.	
$\overline{\text{Min. } V_{\text{DDP}} \text{ voltage to ensure defined pad}}$ states $^{1)}$	$V_{ m DDP}$	PA CC	0.6	_	V
Oscillator start-up time ²⁾	toscs	_S CC	_	30	ms
Minimum PORST active time after power supplies are stable at operating levels	t _{POA}	CC	50	-	ms
HDRST pulse width	t_{HD}	CC	1024 cycles ³⁾		f_{SYS}
Ports inactive after any reset active ²⁾	t_{PI}	CC	_	30	ns

¹⁾ This parameter is valid under assumption that PORST signal is constantly at low level during the power-up/ power-down of the $V_{\rm DDP}$.

 $^{^{2)}\,\,}$ Not subject to production test, verified by design/characterization.

³⁾ Any $\overline{\text{HDRST}}$ activation is internally prolonged to 1024 FPI bus clock cycles.



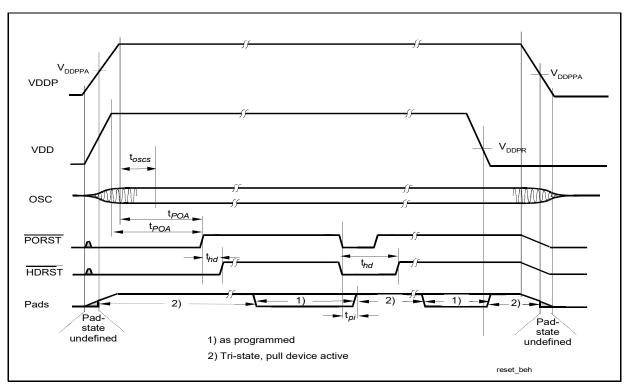


Figure 4-2 Power and Reset Timing



4.3.2 PLL Parameters

When PLL operation is configured ($PLL_CLC.LOCK = 1$), the on-chip phase locked loop is enabled and provides the master clock. The PLL multiplies the input frequency by the factor \mathbf{F} ($f_{MC} = f_{OSC} \times \mathbf{F}$) which results from the input divider, the multiplication factor (N Factor), and the output divider ($\mathbf{F} = NDIV + 1 / (PDIV + 1 \times KDIV + 1)$). The PLL circuit synchronizes the master clock to the input clock. This synchronization is done smoothly, i.e. the master clock frequency does not change abruptly.

Due to this adaptation to the input clock, the frequency of $f_{\rm MC}$ is constantly adjusted so it is locked to $f_{\rm OSC}$. The slight variation causes a jitter of $f_{\rm MC}$ which also affects the duration of individual TCMs.

The timing listed in the AC Characteristics refers to TCPs. Because f_{CPU} is derived from f_{MC} , the timing must be calculated using the minimum TCP possible under the respective circumstances.

The actual minimum value for TCP depends on the jitter of the PLL. As the PLL is constantly adjusting its output frequency in order to correspond to the applied input frequency (crystal or oscillator), the relative deviation for periods of more than one TCP is lower than for one single TCP (see formula and Figure 4-3).

This is especially important for bus cycles using waitstates and for the operation of timers, serial interfaces, etc. For all slower operations and longer periods (e.g. pulse train generation or measurement, lower baud rates, etc.) the deviation caused by the PLL jitter is negligible.

The value of the accumulated PLL jitter depends on the number of consecutive VCO output cycles within the respective timeframe. The VCO output clock is divided by the output prescaler (K = KDIV+1) to generate the master clock signal f_{MC} . Therefore, the number of VCO cycles can be represented as K × N, where N is the number of consecutive f_{MC} cycles (TCM).

For a period of $N \times TCM$, the accumulated PLL jitter is defined by the corresponding deviation D_N :

 D_N [ns] = $\pm (1.5 + 6.32 \times N / f_{MC})$; f_{MC} in [MHz], N = number of consecutive TCMs.

So, for a period of 3 TCMs @ 20 MHz and K = 12: D₃ = \pm (1.5 + 6.32 \times 3 / 20) = 2.448 ns.

This formula is applicable for K \times N < 95. For longer periods, the K \times N=95 value can be used. This steady value can be approximated by: D_{Nmax} [ns] = \pm (1.5 + 600 / (K \times f_{MC})).



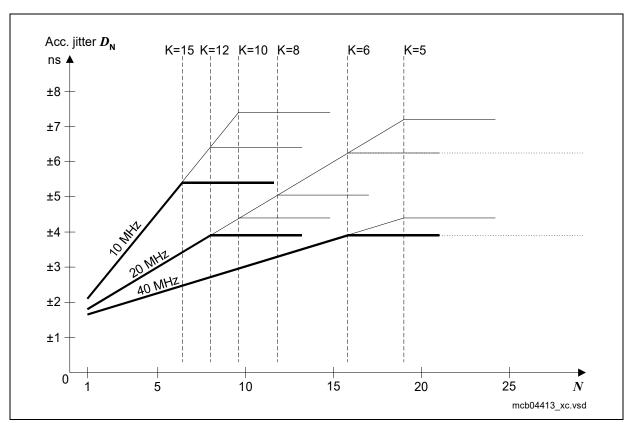


Figure 4-3 Approximated Accumulated PLL Jitter

Note: The bold lines indicate the minimum accumulated jitter which can be achieved by selecting the maximum possible output prescaler factor K.

Different frequency bands can be selected for the VCO, so the operation of the PLL can be adjusted to a wide range of input and output frequencies:

Table 4-7 VCO Bands for PLL Operation

PLL_CLC.VCOSEL	VCO Frequency Range	Base Frequency Range 1)
00	400 500 MHz	250 320 MHz
01	500 600 MHz	300 400 MHz
10	600 700 MHz	350 480 MHz
11	Reserved ²⁾	

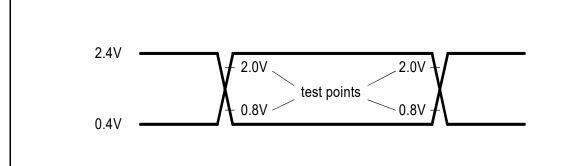
¹⁾ Base Frequency Range is the free running operation frequency of the PLL, when no input clock is available.

²⁾ This option cannot be used.



4.3.3 AC Characteristics

(Operating Conditions apply)



AC inputs during testing are driven at 2.4V for a logic "1" and 0.4V for a logic "0". Timing measurements are made at V_{IHmin} for a logic "1" and V_{ILmax} for a logic "0".

Figure 4-4 Input/Output Waveforms for AC Tests - for GPIO, Dedicated and EBU pins



4.3.4 Input Clock Timing

(Operating Conditions apply)

Parameter		Symbol	Limits		Unit
			min	max	
Oscillator clock frequency	with PLL	$f_{\rm OSC}$ SR	4	25	MHz
Input clock frequency driving at XTAL1	with PLL	foscdd SR	-	40	MHz
Input Clock Duty Cycle (t ₁ /t ₂)	•	SR	45	55	%

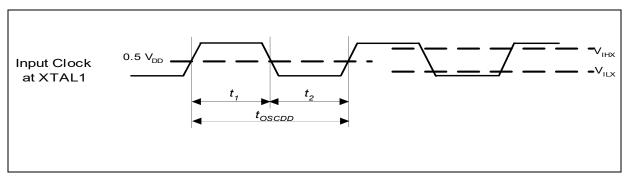


Figure 4-5 Input Clock Timing



4.3.5 Port Timing

(Operating Conditions apply; C_L=50pF)

Parameter	Sym	bol	ol Limits		Unit
			min	max	
Port data valid from TRCLK ¹⁾ up to 120 MHz ²⁾	t_1	CC	_	13	ns

¹⁾ Port data is output with respect to the FPI clock. The TRCLK is used as a reference here since the FPI clock is not available as an external pin and TRCLK is same frequency as CPU clock. Port lines maintain their states for at least 2 CPU clocks.

²⁾ 120 MHz is verified by design/characterization.

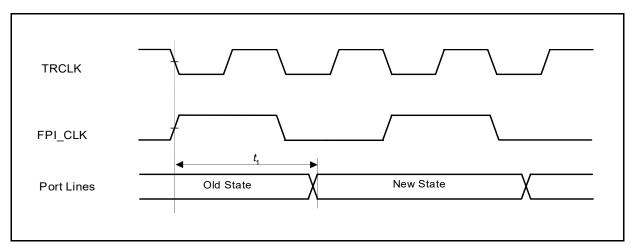


Figure 4-6 Port Timing



4.3.6 Timing for JTAG Signals

(Operating Conditions apply; C_L=50pF)

Parameter	Symbol	Lir	Limits		
		min	max		
TCK clock period	$t_{TCK}SR$	50	_	ns	
TCK high time	t_1 SR	10	1-	ns	
TCK low time	t_2 SR	29	<u> </u>	ns	
TCK clock rise time	t_3 SR	_	0.4	ns	
TCK clock fall time	t_4 SR	_	0.4	ns	

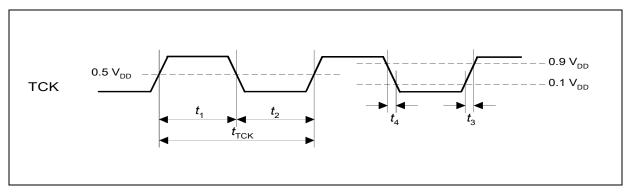


Figure 4-7 TCK Clock Timing



Parameter	meter Symbol		Lin	Unit	
			min	max	
TMS setup to TCK _F	t_1	SR	7.85	_	ns
TMS hold to TCK _/	t_2	SR	3.0	_	ns
TDI setup to TCK _r	t_1	SR	10.9	_	ns
TDI hold to TCK _r	t_2	SR	3.0	_	ns
TDO valid output from TCK ~_	t_3	CC	_	10.7	ns
TDO high impedance to valid output from TCK 🦜	t_4	CC	_	23.0	ns
TDO valid output to high impedance from TCK ~	t_5	CC	_	26.0	ns

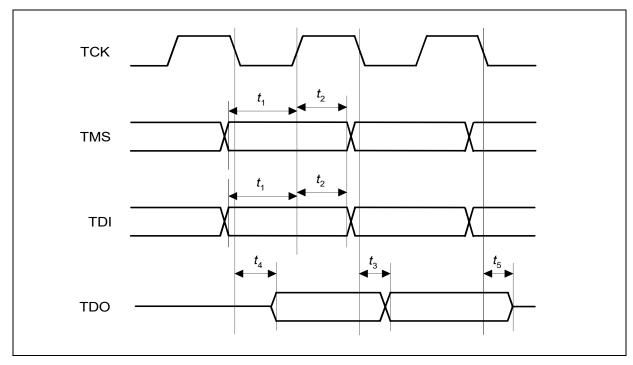


Figure 4-8 JTAG Timing



4.3.7 Timing for OCDS Trace and Breakpoint Signals

(Operating Conditions apply; $C_L(TRCLK) = 25 \text{ pF}$, $C_L = 50 \text{ pF}$)

Parameter	Symbol		Limits		Unit
			min	max	
BRK_OUT valid from TRCLK _r	t_1	CC	_	5.2	ns
OCDS2_STATUS[4:0] valid from TRCLK _r	<i>t</i> ₁	CC	0	5	ns
OCDS2_INDIR_PC[7:0] valid from TRCLK	t_1	CC	0	5	ns
OCDS2_BRKPT[2:0] valid from TRCLK	t_1	CC	0	5	ns

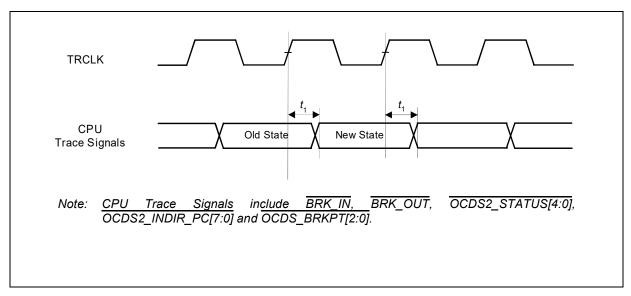


Figure 4-9 OCDS Trace Signals Timing



4.3.8 EBU Timings

4.3.8.1 SDCLKO Output Clock Timing

(Operating Conditions apply; CL = 50 pF)

Parameter	Symbol	Limits ¹⁾		Limits ²⁾		Unit
		min	max	min	max	
SDCLKO period	t ₁ CC	10	_	8.3	_	ns
SDCLKO high time	t ₂ CC	3	_	2.5	_	ns
SDCLKO low time	t ₃ CC	3	_	2.5	_	ns
SDCLKO rise time	t ₄ CC	_	2.5	_	2.5	ns
SDCLKO fall time	t ₅ CC	_	2.5	_	2.5	ns

¹⁾ The parameters are applicable for PC100 SDRAM access and the maximum SDCLKO is up to 100 MHz.

4.3.8.2 BFCLKO Output Clock Timing

(Operating Conditions apply; $C_{l} = 50 \text{ pF}$)

Parameter	Sy	Symbol Limit 1)		nit ¹⁾	Lin	Unit	
			min	max	min	max	
Clock period	t_1	CC	20	_	16.7	_	ns
BFCLKO high time	t_2	CC	6.6	_	7.5	_	ns
BFCLKO low time	t_3	CC	6.6	_	7.5	_	ns
BFCLKO rise time	t_4	CC	_	3.5	_	3.5	ns
BFCLKO fall time	t_5	CC	_	2.5	_	2.5	ns

¹⁾ The CPU runs at 150 MHz and the Burst Flash runs at divided by 3 clock.

²⁾ The parameters are applicable for PC133 SDRAM access and the maximum SDCLKO is up to 120 MHz.

²⁾ The CPU runs at 120 MHz and the Burst Flash runs at divided by 2 clock.



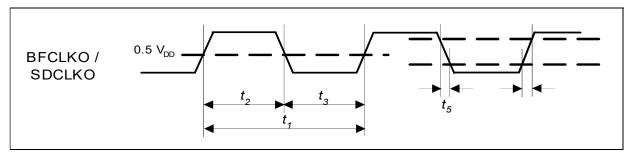


Figure 4-10 EBU Clock Output Timing

4.3.8.3 Timing for SDRAM Access Signals

(Operating Conditions apply; $C_L = 50 \text{ pF}^{1}$)

Parameter	Symbol		Lim	its ²⁾	Limits ³⁾		Unit
			min	max	min	max	
SDCLKO period	t_1	CC	10	_	8.3	_	ns
CKE output valid time from SDCLKO _r	t_1	CC	-	8.0	_	6.8	ns
CKE output hold time from SDCLKO _r	t_2	CC	0	_	0.8	_	ns
Address output valid time from SDCLKO	t_3	CC	_	8.0	_	6.8	ns
Address output hold time from SDCLKO	<i>t</i> ₄	CC	1.0	_	0.8	_	ns
CSx, RAS, CAS, RD/WR, BC(3:0) output valid time from SDCLKO _√	<i>t</i> ₅	CC	_	8.0	_	6.8	ns
CSx, RAS, CAS, RD/WR, BC(3:0) output hold time from SDCLKO _✓	<i>t</i> ₆	CC	1.0	_	8.0	_	ns
AD(31:0) output valid time from SDCLKO _<	<i>t</i> ₇	CC	-	8.0	_	6.8	ns
AD(31:0) output hold time from SDCLKO	<i>t</i> ₈	CC	1.0	_	0.8	_	ns
AD(31:0) input setup time to SDCLKO _r	t_9	SR	4.0	_	2.9	_	ns
AD(31:0) input hold time from SDCLKO 🖌	<i>t</i> ₁₀	SR	3.0	_	3.0	_	ns

¹⁾ If application conditions other than 50 pf capacitive load are used, then the proper correlation factor should be used for your specific application condition. For design team, the load should be set according to the system requirement.

²⁾ The parameters are applicable for PC100 SDRAM access and the maximum SDCLKO is up to 100 MHz.

³⁾ The parameters are applicable for PC133 SDRAM access and the maximum SDCLKO is up to 120 MHz.



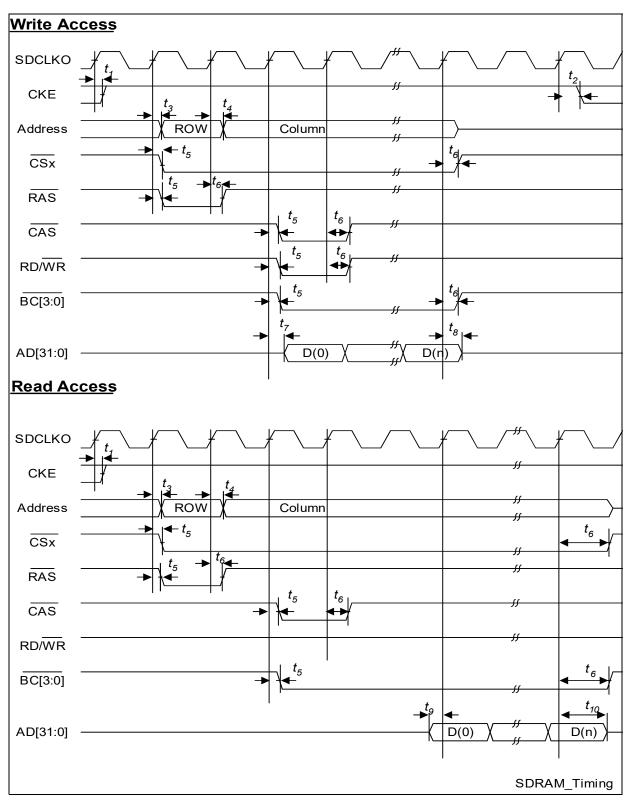


Figure 4-11 SDRAM Access Timing



4.3.8.4 Timing for Burst Flash Access Signals

(Operating Conditions apply; $C_L = 50 \text{ pF}$)

Parameter	ameter Symbol		Limits		Unit
			min	max	
Address output valid time from BFCLKO	<i>t</i> ₁	CC	_	11.0	ns
Address output hold time from BFCLKO _r	t_2	CC	10.0	_	ns
CSx output valid time from BFCLKO _√	<i>t</i> ₃	CC	_	9.0	ns
RD output valid time from BFCLKO	t_4	CC	_	10.0	ns
ADV output valid time from BFCLKO	<i>t</i> ₅	CC	_	10.0	ns
ADV output hold time from BFCLKO	<i>t</i> ₆	CC	3.0	_	ns
BAA output valid time from BFCLKO	<i>t</i> ₇	CC	_	10.0	ns
BAA output hold time from BFCLKO _r	<i>t</i> ₈	CC	3.0	_	ns
AD(31:0) input setup time to BFCLKO _r	t_9	SR	5.0	_	ns
AD(31:0) input hold time from BFCLKO	t ₁₀	SR	3.0	_	ns
WAIT input setup time to BFCLKO _	t ₁₁	SR	5.0	_	ns
WAIT input hold time from BFCLKO _	t ₁₂	SR	3.0	_	ns



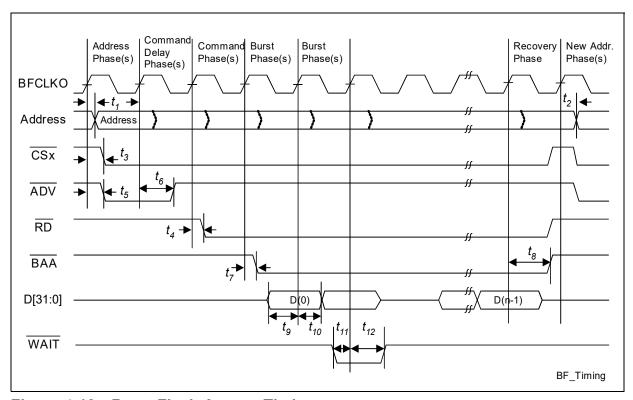


Figure 4-12 Burst Flash Access Timing

Note: Output delays are always referenced to BFCLKO. The reference clock for input characteristics depends on bit BFCON.FDBKEN.

BFCON.FDBKEN = 0: BFCLKO is the input reference clock.

BFCON.FDBKEN = 1: BFCLKI is the input reference clock (EBULMB clock feedback enabled).



4.3.8.5 Timing for Demultiplexed Access Signals

(Operating Conditions apply; $C_L = 50 \text{ pF}$) 1)

Parameter		nbol	Limits		Unit
			min	max	
CSx, RD/WR, RD, MR/W, BC(3:0) output valid time from output clock _√	t_1	CC	_	9	ns
CSx, RD/WR, RD, MR/W, BC(3:0) output hold time from output clock _√		CC	0.0	_	ns
Address output valid time from output clock 🖌	<i>t</i> ₃	CC	_	9	ns
Address output hold time from output clock _/	<i>t</i> ₄	CC	0.0	_	ns
WAIT input setup time to output clock _r	<i>t</i> ₇	SR	12	_	ns
WAIT input hold time from output clock _r	<i>t</i> ₈	SR	3	_	ns
AD(31:0) output valid time from output clock _/	<i>t</i> ₉	CC	_	9	ns
AD(31:0) output hold time from output clock _r	t ₁₀	CC	0.0	_	ns
AD(31:0) input setup time to output clock _r		SR	1.3	_	ns
AD(31:0) input hold time from output clock \checkmark	t ₁₂	SR	3	_	ns
RMW output valid time from output clock _r	t ₁₃	CC	-	8	ns
RMW output hold time from output clock _r	t ₁₄	CC	1.3	-	ns
AD(31:0) output hold time from RD/WR _r	t ₁₆	CC	0	-	ns

¹⁾ The purpose for characterization of Asynchronous access is to provide the performance of all of the signals to user. User can decide whether an extra cycle is needed or not based on above parameters to generate signals with correct timing sequence. It is user's responsibility to program the correct phase length according to the memory/peripheral device specification and EBU specification.



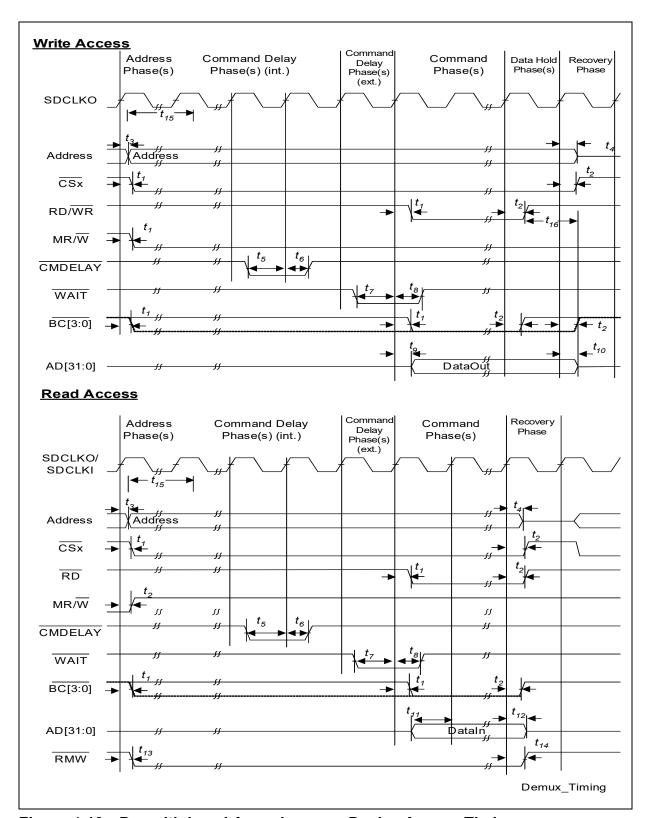


Figure 4-13 Demultiplexed Asynchronous Device Access Timing



4.3.8.6 Timing for Multiplexed Access Signals

(Operating Conditions apply; $C_L = 50 \text{ pF})^{1)}$

Parameter		nbol	Limits		Unit
			min	max	
ALE, \overline{CSx} , RD/ \overline{WR} , \overline{RD} , MR/ \overline{W} , $\overline{BC(3:0)}$ output valid time from output clock \mathcal{I}	t_1	CC	_	9	ns
ALE, \overline{CSx} , \overline{RD} , \overline{MR} , \overline{MR} , \overline{BC} output hold time from output clock \mathcal{I}	t_2	CC	0.0	_	ns
AD(31:0) output valid time from output clock 🔟	t_3	CC	_	9	ns
AD(31:0) output hold time from output clock _r	<i>t</i> ₄	CC	0.0	_	ns
AD(31:0) input setup time to output clock _r	t_5	SR	1.4	_	ns
AD(31:0) input hold time from output clock _r	<i>t</i> ₆	SR	3	_	ns
WAIT input setup time to output clock _r	t ₉	SR	12	_	ns
WAIT input hold time from output clock _r	t ₁₀	SR	3	_	ns
RMW output valid time from output clock _r		CC	_	8	ns
RMW output hold time from output clock _r	t ₁₂	CC	1.3	_	ns
ALE width		CC	8.5	_	ns
AD(31:0) output hold time from RD/WR _	t ₁₄	CC	0	_	ns

¹⁾ The purpose for characterization of Asynchronous access is to provide the performance of all of the signals to user. User can decide whether an extra cycle is needed or not based on above parameters to generate signals with correct timing sequence. It is user's responsibility to program the correct phase length according to the memory/peripheral device specification and EBU Specification.



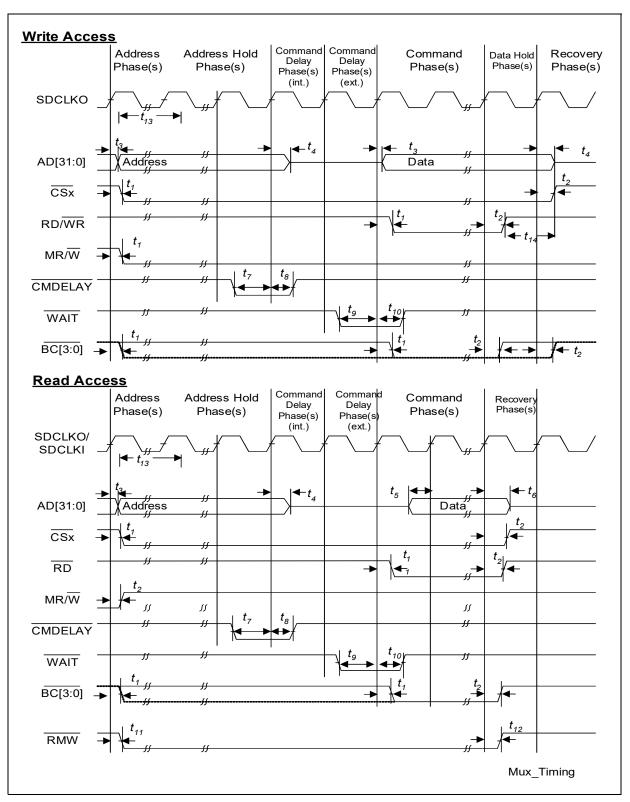


Figure 4-14 Write Access in Multiplexed Access



4.3.9 Peripheral Timings

4.3.9.1 Timing for Ethernet Signals

(Operating Conditions apply; $C_L = 50 \text{ pF}$)

Parameter		nbol	Lir	nits	Unit
			min	max	
ETXCLK period (10 Mbit/sec Ethernet)	<i>t</i> ₁	SR	400.0	_	ns
ETXCLK high time (10 Mbit/sec Ethernet)	t_2	SR	140	260	ns
ETXCLK low time (10 Mbit/sec Ethernet)	t_3	SR	140	260	ns
ETXCLK period (100 Mbit/sec Ethernet)	<i>t</i> ₁	SR	40.0	_	ns
ETXCLK high time (100 Mbit/sec Ethernet)	t_2	SR	14	26	ns
ETXCLK low time (100 Mbit/sec Ethernet)	t_3	SR	14	26	ns
ERXCLK period (10 Mbit/sec Ethernet)	<i>t</i> ₁	SR	400.0	_	ns
ERXCLK high time (10 Mbit/sec Ethernet)	t_2	SR	140	260	ns
ERXCLK low time (10 Mbit/sec Ethernet)	t_3	SR	140	260	ns
ERXCLK period (100 Mbit/sec Ethernet)	t_1	SR	40.0	_	ns
ERXCLK high time (100 Mbit/sec Ethernet)	t_2	SR	14	26	ns
ERXCLK low time (100 Mbit/sec Ethernet)	<i>t</i> ₃	SR	14	26	ns
ERXD(3:0) input setup to ERXCLK	t_4	SR	10.0	_	ns
ERXD(3:0) input hold from ERXCLK	<i>t</i> ₅	SR	_	10.0	ns
ERXDV input setup to ERXCLK	t ₄	SR	10.0	_	ns
ERXDV input hold from ERXCLK	<i>t</i> ₅	SR	_	10.0	ns
ERXER input setup to ERXCLK	t ₄	SR	10.0	_	ns
ERXER input hold from ERXCLK	t_5	SR	_	10.0	ns
ETXD(3:0) output valid from ETXCLK	<i>t</i> ₆	CC	_	25.0	ns
ETXEN output valid from ETXCLK	<i>t</i> ₆	CC	_	25.0	ns
ETXER output valid from ETXCLK	<i>t</i> ₆	CC	_	25.0	ns
EMDC clock period	<i>t</i> ₇	CC	400.0	_	ns
EMDC high time	t ₈	CC	160	_	ns
EMDC low time	<i>t</i> ₉	CC	160	_	ns
EMDIO input setup to EMDC (sourced by STA)	t ₁₀	SR	10.0	_	ns



Parameter	Symbol	l Limits		Unit
		min	max	
EMDIO input hold from EMDC (sourced by STA)	<i>t</i> ₁₁ SR	_	10.0	ns
EMDIO output valid from EMDC (sourced by PHY)	t ₁₂ CC	_	300.0	ns

Note: Any other parameters which are not stated here, please refer to ANSI/IEEE Std 802.3, Section 22.3.

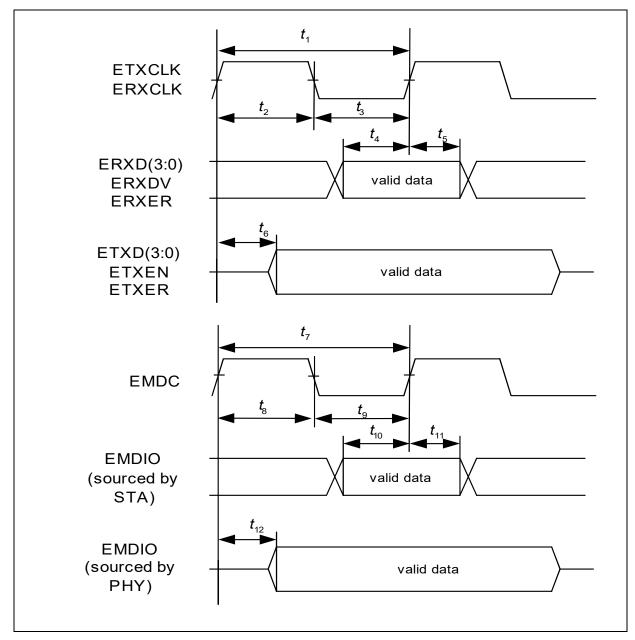


Figure 4-15 Ethernet Timing



4.3.9.2 SSC Master Mode Timing

(Operating Conditions apply; $C_L = 50 \text{ pF}$)

Parameter		Symbol	Limit Values		Unit
			min.	max.	
SCLK clock period	t_0	CC	2*T _{SSC} 1)	_	ns
MTSR/SLSOx delay from SCLK 🖍	t_1	CC	0	8	ns
MRST setup to SCLK 🌂	t_2	SR	10	_	ns
MRST hold from SCLK 🤟	t_3	SR	5	_	ns

¹⁾ $T_{SSCmin} = T_{SYS} = 1/f_{SYS}$. When $f_{SYS} = 120$ MHz, $t_0 = 16.7$ ns

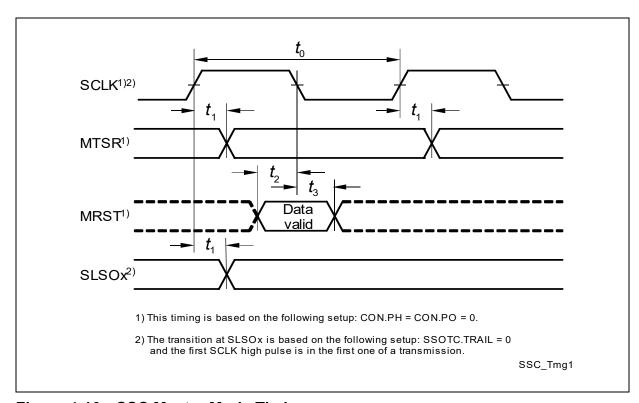


Figure 4-16 SSC Master Mode Timing



4.3.9.3 MLI Interface Timing

(Operating Conditions apply; $C_L = 50 \text{ pF}$)

Parameter Symbo		Limit	Unit	
		min.	max.	
TCLK/RCLK clock period	t ₀ CC/SR	2*T _{MLI} 1)	_	ns
MLI outputs delay from TCLK 🖌	t ₅ CC	0	8	ns
MLI inputs setup to RCLK 飞	t ₆ SR	4	_	ns
MLI inputs hold to RCLK 飞	t ₇ SR	4	_	ns
RREADY output delay from TCLK 飞	t ₈ CC	0	8	ns

¹⁾ $T_{MLImin} = T_{SYS} = 1/f_{SYS}$. When $f_{SYS} = 120$ MHz, $t_0 = 16.7$ ns

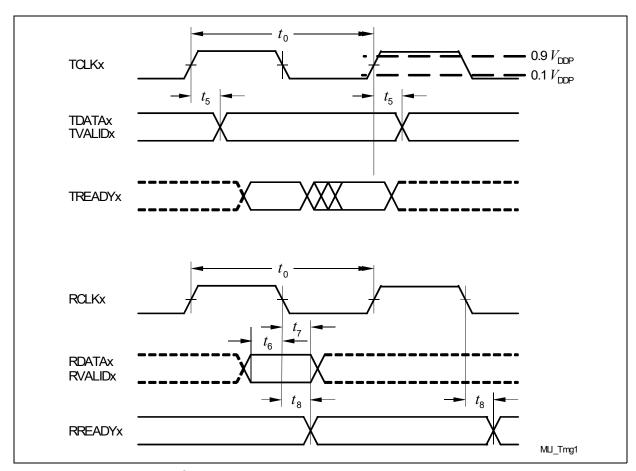


Figure 4-17 MLI Interface Timing

Note: The generation of RREADYx is in the input clock domain of the receiver. The reception of TREADYx is asynchronous to TCLKx.



4.3.9.4 Timing for USB Transceiver Signals

(Operating Conditions apply; $C_L = 50 \text{ pF}$)

Parameter	Symbol	Limits		Unit
		min	max	
Full speed mode rise time	t _{FR} CC	4	20	ns
Full speed mode fall time	t_{FF} CC	4	20	ns

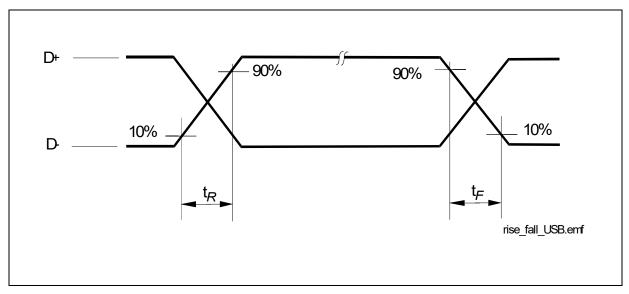


Figure 4-18 AC Testing: Input, Output Waveforms



4.4 Package Outline

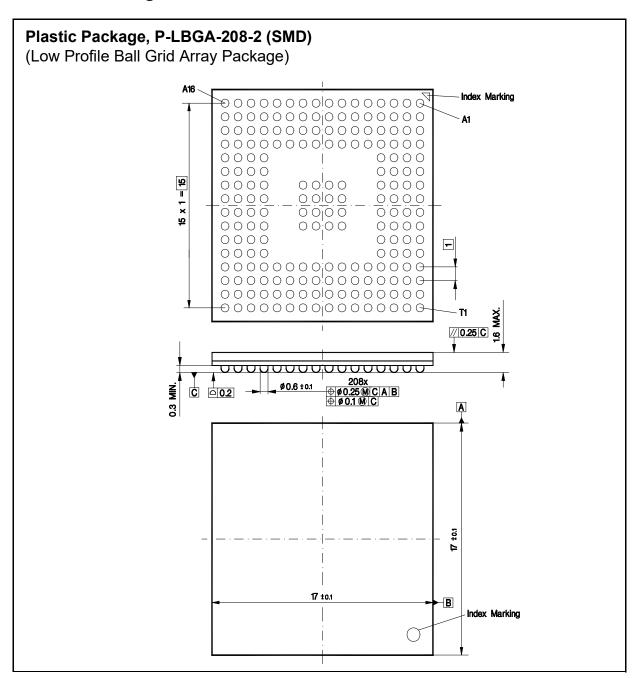


Figure 4-19 P-LBGA-208-2 Package

You can find all of our packages, sorts of packing and others in our Infineon Internet Page "Products": http://www.infineon.com/products.

SMD = Surface Mounted Device

Dimensions in mm

 $w\ w\ w\ .\ i\ n\ f\ i\ n\ e\ o\ n\ .\ c\ o\ m$

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