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3.3-V PHASE-LOCK LOOP CLOCK DRIVER WITH 3-STATE OUTPUTS

FEATURES

- Low Output Skew for Clock-Distribution and **Clock-Generation Applications**
- Operates at 3.3-V V_{CC}
- **Distributes One Clock Input to Six Outputs**
- One Select Input Configures Three Outputs to Operate at One-Half or Double the Input Frequency
- No External RC Network Required
- **On-Chip Series Damping Resistors**
- External Feedback Pin (FBIN) Is Used to **Synchronize the Outputs to the Clock Input**
- Application for Synchronous DRAM, **High-Speed Microprocessor**
- **TTL-Compatible Inputs and Outputs**
- Outputs Drive 50- Ω Parallel-Terminated **Transmission Lines**
- State-of-the-Art *EPIC-IIB™* BiCMOS Design **Significantly Reduces Power Dissipation**
- Distributed V_{CC} and Ground Pins Reduce **Switching Noise**
- Packaged in Plastic 28-Pin Shrink **Small-Outline Package**

(TOP VIEW) AV_{CC} 28 AV_{CC} AGND 12 27 AGND CLKIN []3 26 FBIN SEL []4 25 TEST <u>OE</u> []₅ 24 CLR GND [6 23 V_{CC} 1Y1 **[**]₇ 22 2Y1 V_{CC} **[**8 21 GND GND []9 20 V_{CC} 1Y2 []₁₀ 19 2Y2 18 GND V_{CC} 11 GND [] 12 17 VCC 1Y3 []₁₃ 16 2Y3 15 GND V_{CC} [] 14

DB PACKAGE

DESCRIPTION

The CDC2536 is a high-performance, low-skew, low-jitter clock driver. It uses a phase-lock loop (PLL) to precisely align, in both frequency and phase, the clock output signals to the clock input (CLKIN) signal. It is specifically designed for use with synchronous DRAMs and popular microprocessors operating at speeds from 50 MHz to 100 MHz or down to 25 MHz on outputs configured as half-frequency outputs. The CDC2536 operates at 3.3-V V_{CC} and is designed to drive a 50-W transmission line. The CDC2536 also provides on-chip series-damping resistors, eliminating the need for external termination components.

The feedback (FBIN) input is used to synchronize the output clocks in frequency and phase to the input clock (CLKIN). One of the six output clocks must be fed back to FBIN for the PLL to maintain synchronization between CLKIN and the outputs. The output used as the feedback pin is synchronized to the same frequency as CLKIN.

The Y outputs can be configured to switch in phase and at the same frequency as CLKIN. The select (SEL) input configures three Y outputs to operate at one-half or double the CLKIN frequency, depending on which pin is fed back to FBIN (see Tables 1 and 2). All output signal duty cycles are adjusted to 50% independent of the duty cycle at the input clock.

Output-enable (OE) is provided for output control. When OE is high, the outputs are in the high-impedance state. When OE is low, the outputs are active. TEST is used for factory testing of the device and can be use to bypass the PLL. TEST should be strapped to GND for normal operation.

Unlike many products containing PLLs, the CDC2536 does not require external RC networks. The loop filter for the PLL is included on-chip, minimizing component count, board space, and cost.



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EPIC-IIB is a trademark of Texas Instruments.



Because it is based on PLL circuitry, the CDC2536 requires a stabilization time to achieve phase lock of the feedback signal to the reference signal. This stabilization time is required following power up and application of a fixed-frequency, fixed-phase signal at CLKIN, as well as following any changes to the PLL reference or feedback signals. Such changes occur upon change of SEL, enabling the PLL via TEST, and upon enable of all outputs via $\overline{\text{OE}}$.

The CDC2536 is characterized for operation from 0°C to 70°C.

DETAILED DESCRIPTION OF OUTPUT CONFIGURATIONS

The voltage-controlled oscillator (VCO) used in the CDC2536 has a frequency range of 100 MHz to 200 MHz, twice the operating frequency of the CDC2536 outputs. The output of the VCO is divided by two and by four to provide reference frequencies with a 50% duty cycle of one-half and one-fourth the VCO frequency. SEL determines which of the two signals is buffered to each bank of device outputs.

One device output must be externally wired to FBIN to complete the PLL. The VCO operates such that the frequency of the output matches that of CLKIN. In the case that a VCO/2 output is wired to FBIN, the VCO must operate at twice the CLKIN frequency resulting in device outputs that operate at either the same or one-half the CLKIN frequency. If a VCO/4 output is wired to FBIN, the device outputs operate at the same or twice the CLKIN frequency.

Output Configuration A

Output configuration A is valid when any output configured as a $1\times$ frequency output in Table 1 is fed back to FBIN. The input frequency range for CLKIN is 50 MHz to 100 MHz when using output configuration A. Outputs configured as $1/2\times$ outputs operate at half the CLKIN frequency, while outputs configured as $1\times$ outputs operate at the same frequency as CLKIN.

Table 1. Output Configuration A

INPUT	OUTP	PUTS
SEL	1/2× FREQUENCY	1× FREQUENCY
L	None	All
Н	1Yn	2Yn

Output Configuration B

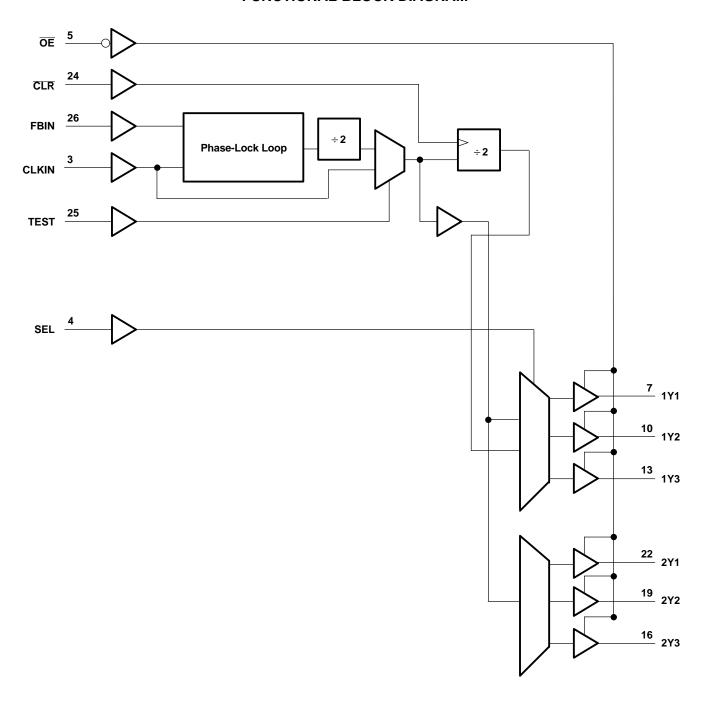
Output configuration B is valid when any output configured as a $1\times$ frequency output in Table 2 is fed back to FBIN. The input frequency range for CLKIN is 25 MHz to 50 MHz when using output configuration B. Outputs configured as $1\times$ outputs operate at the CLKIN frequency, while outputs configured as $2\times$ outputs operate at double the frequency of CLKIN.

Table 2. Output Configuration B

INPUT	OUTP	UTS
SEL	1× FREQUENCY	2× FREQUENCY
Н	1Yn	2Yn
L	All	None



FUNCTIONAL BLOCK DIAGRAM





FUNCTIONAL BLOCK DIAGRAM (continued)

Terminal Functions

TERM	/INAL		
NAME	NO.	I/O	DESCRIPTION
CLKIN	3	I	Clock input. CLKIN provides the clock signal to be distributed by the CDC2536 clock-driver circuit. CLKIN provides the reference signal to the integrated PLL that generates the clock output signals. CLKIN must have a fixed frequency and fixed phase for the phase-lock loop to obtain phase lock. Once the circuit is powered up and a valid CLKIN signal is applied, a stabilization time is required for the PLL to phase lock the feedback signal to its reference signal.
CLR	24	ı	CLR is used for testing purposes only. Connect CLR to GND for normal operation.
FBIN	26	ı	Feedback input. FBIN provides the feedback signal to the internal PLL. FBIN must be hardwired to one of the six clock outputs to provide frequency and phase lock. The internal PLL adjusts the output clocks to obtain zero phase delay between the FBIN and differential CLKIN inputs.
ŌĒ	5	I	Output enable. $\overline{\text{OE}}$ is the output enable for all outputs. When $\overline{\text{OE}}$ is low, all outputs are enabled. When $\overline{\text{OE}}$ is high, all outputs are in the high-impedance state. Since the feedback signal for the PLL is taken directly from an output, placing the outputs in the high-impedance state interrupts the feedback loop; therefore, when a high-to-low transition occurs at $\overline{\text{OE}}$, enabling the output buffers, a stabilization time is required before the PLL obtains phase lock.
SEL	4	ı	Output configuration select. SEL selects the output configuration for each output bank (e.g. 1×, 1/2×, or 2×).(see Tables 1 and 2).
TEST	25	ı	TEST is used to bypass the PLL circuitry for factory testing of the device. When TEST is low, all outputs operate using the PLL circuitry. When TEST is high, the outputs are placed in a test mode that bypasses the PLL circuitry. TEST should be grounded for normal operation.
1Y1-1Y3	7, 10, 13	0	These outputs are configured by SEL to transmit one-half or one-fourth the frequency of the VCO. The relationship between the CLKIN frequency and the output frequency is dependent on SEL. The duty cycle of the Y output signals is nominally 50%, independent of the duty cycle of the CLKIN signal. Each output has an internal series resistor to dampen transmission-line effects and improve the signal integrity at the load.
2Y1-2Y3	22, 19, 16	0	These outputs transmit one-half the frequency of the VCO. The relationship between the CLKIN frequency and the output frequency is dependent on the frequency of the output being fed back to FBIN. The duty cycle of the Y output signals is nominally 50%, independent of the duty cycle of the CLKIN signal. Each output has an internal series resistor to dampen transmission-line effects and improve the signal integrity at the load.

ABSOLUTE MAXIMUM RATINGS

over operating free-air temperature range (unless otherwise noted)(1)

	UNIT
Supply voltage range, V _{CC}	-0.5 V to 4.6 V
Input voltage range, V _I (see ⁽²⁾)	-0.5 V to 7 V
Voltage range applied to any output in the high state or power-off state, V_0 (see $^{(2)}$)	-0.5 V to 5.5 V
Current into any output in the low state, I _O	24 mA
Input clamp current, $I_{IK}(V_I < 0)$	-20 mA
Output clamp current, I _{OK} (V _O < 0)	-50 mA
Maximum power dissipation at $T_A = 55^{\circ}C$ (in still air) (see $^{(3)}$)	0.68 W
Storage temperature range, T _{stg}	-65°C to 150°C

⁽¹⁾ Stresses beyond those listed under absolute maximum ratings may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under recommended operating conditions is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

⁽²⁾ The input and output negative-voltage ratings may be exceeded if the input and output clamp-current ratings are observed.

⁽³⁾ The maximum package power dissipation is calculated using a junction temperature of 150°C and a board trace length of 75 mils. For more information, refer to the *Package Thermal Considerations* application note in the 1994 *ABT Advanced BiCMOS Technology Data Book*, literature number SCBD002B.



RECOMMENDED OPERATING CONDITIONS (SEE (1))

		MIN	MAX	UNIT
V _{CC}	Supply voltage	3	3.6	V
V _{IH}	High-level input voltage	2		V
V _{IL}	Low-level input voltage		0.8	V
VI	Input voltage	0	5.5	V
I _{OH}	High-level output current		12	mA
I _{OL}	Low-level output current		12	mA
T _A	Operating free-air temperature	0	70	°C

⁽¹⁾ Unused inputs must be held high or low to prevent them from floating.

ELECTRICAL CHARACTERISTICS

over recommended operating free-air temperature range, T_A = 25°C (unless otherwise noted)

PARAMETER		TEST CONDITIONS		MIN	MAX	UNIT
V _{IK}	V _{CC} = 3 V,	I _I = -18 mA			1.2	V
V _{OH}	$V_{CC} = MIN \text{ to } MAX^{(1)},$	I _{OH} = -100 μA		V _{CC} - 0.2		V
	$V_{CC} = 3 V$,	$I_{OH} = -12 \text{ mA}$		2		
\ <u>\</u>	$V_{CC} = 3 V$,	$I_{OL} = 100 \mu A$			0.2	V
V _{OL}	$V_{CC} = 3 V$,	$I_{OL} = 12 \text{ mA}$			0.8	V
	$V_{CC} = 0 \text{ or } MAX^{(1)},$	V _I = 3.6 V			±10	
l ₁	$V_{CC} = 3.6 \text{ V},$	$V_I = V_{CC}$ or GND			±1	μA
I _{OZH}	V _{CC} = 3.6 V,	V _O = 3 V			10	μΑ
I _{OZL}	V _{CC} = 3.6 V,	V _O = 0			10	μΑ
			Outputs high		2	
I _{cc}	$V_{CC} = 3.6 \text{ V}, I_{O} = 0, V_{I} = V_{CC}$	_C or GND	Outputs low		2	mA
			Outputs disabled		2	
C _i	$V_I = V_{CC}$ or GND				6	pF
C _o	$V_O = V_{CC}$ or GND				9	pF

⁽¹⁾ For conditions shown as MIN or MAX, use the appropriate value specified under recommended operating conditions.

TIMING REQUIREMENTS

over recommended ranges of supply voltage and operating free-air temperature

			MIN	MAX	UNIT
f	Clock frequency	When VCO is operating at four times the CLKIN frequency	25	50	MHz
^I clock	ock Clock frequency	When VCO is operating at double the CLKIN frequency	50	100	IVITZ
	Duty cycle, CLKIN		40%	60%	
	Duty cycle, CLKIN Stabilization time ⁽¹⁾	After SEL		50	
		After OE ↓		50	
	Stabilization time (1)	After power up		50	μs
	Duty cycle, CLKIN	After CLKIN		50	

⁽¹⁾ Time required for the integrated PLL circuit to obtain phase lock of its feedback signal to its reference signal. For phase lock to be obtained, a fixed-frequency, fixed-phase reference signal must be present at CLKIN. Until phase lock is obtained, the specifications for propagation delay and skew parameters given in the switching characteristics table are not applicable.



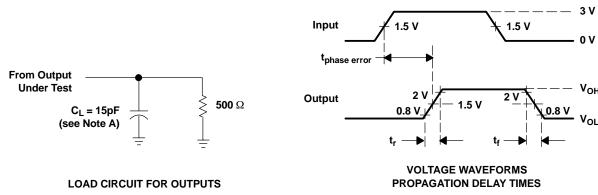
SWITCHING CHARACTERISTICS

over recommended ranges of supply voltage and operating free-air temperature, $C_L = 15 \text{ pF}$ (see $^{(1)}$ and Figure 1, Figure 2 and Figure 3)

PARAMETER	FROM (INPUT)	TO (OUTPUT)	MIN	MAX	UNIT
f _{max}			100		MHz
Duty cycle		Y	45%	55%	
t _{phase error} (2)	CLKIN↑	Y↑	500	+500	ps
t _{jitter} (RMS)	CLKIN↑	Y↑		200	ps
t _{sk(0)} (2)				0.5	ns
t _{sk(pr)} (2)				1	ns
t _r				1.4	ns
t _f				1.4	ns

- (1) The specifications for parameters in this table are applicable only after any appropriate stabilization time has elapsed.
- (2) The propagation delay, t_{phase error}, is dependent on the feedback path from any output to FBIN. The t_{phase error}, t_{sk(o)}, and t_{sk(pr)} specfications are only valid for equal loading of all outputs.

PARAMETER MEASUREMENT INFORMATION

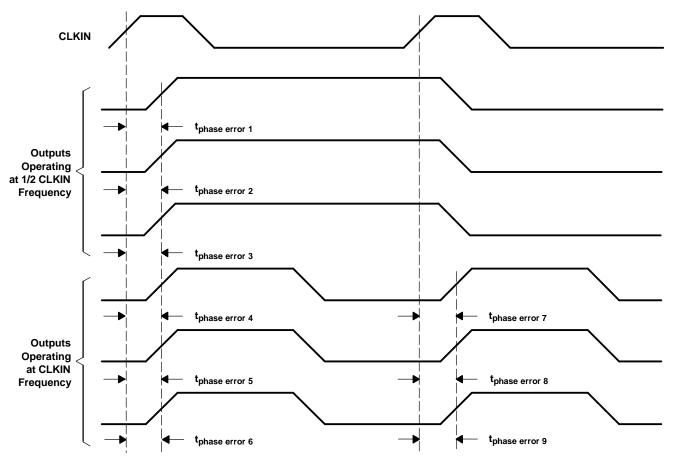


- A. NOTES: . C_L includes probe and jig capacitance.
- B. All input pulses are supplied by generators having the following characteristics: PRR \leq 100 MHz, $Z_O = 50 \Omega$, $t_r \leq$ 2.5 ns, $t_r \leq$ 2.5 ns.
- C. The outputs are measured one at a time with one transition per measurement.

Figure 1. Load Circuit and Voltage Waveforms



PARAMETER MEASUREMENT INFORMATION (continued)

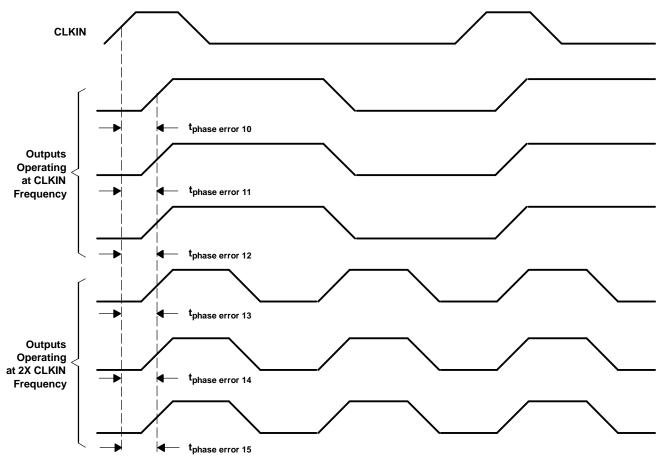


- A. NOTES: . Output skew, $t_{sk(0)},$ is calculated as the greater of:
 - The difference between the fastest and slowest of $t_{phase\ error\ n}$ (n = 1, 2, . . . 6)
 - The difference between the fastest and slowest of $t_{phase\ error\ n}$ (n = 7, 8, 9)
- B. Process skew, $t_{sk(pr)}$, is calculated as the greater of:
 - The difference between the maximum and minimum $t_{phase\ error\ n}$ (n = 1, 2, . . . 6) across multiple devices under identical operating conditions
 - The difference between the maximum and minimum t_{phase error n} (n = 7, 8, 9) across multiple devices under identical operating conditions

Figure 2. Waveforms for Calculations of $t_{sk(p)}$ and $t_{sk(pr)}$



PARAMETER MEASUREMENT INFORMATION (continued)



- A. NOTES: . Output skew, $t_{sk(0)}$, is calculated as the greater of:
 - The difference between the fastest and slowest of $t_{phase\ error\ n}$ (n = 10, 11, . . . 15)
- B. Process skew, $t_{sk(pr)}$, is calculated as the greater of:
 - The difference between the maximum and minimum t_{phase error n} (n = 10, 11, ... 15) across multiple devices under identical operating conditions.

Figure 3. Waveforms for Calculation of $t_{\text{sk(o)}}$ and $t_{\text{sk(pr)}}$





6-Feb-2020

PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan	Lead/Ball Finish	MSL Peak Temp	Op Temp (°C)	Device Marking	Samples
CDC2536DB	ACTIVE	SSOP	DB	28	50	Green (RoHS & no Sb/Br)	NIPDAU	Level-1-260C-UNLIM	0 to 70	CDC2536	Samples
CDC2536DBG4	ACTIVE	SSOP	DB	28	50	Green (RoHS & no Sb/Br)	NIPDAU	Level-1-260C-UNLIM	0 to 70	CDC2536	Samples
CDC2536DBR	ACTIVE	SSOP	DB	28	2000	Green (RoHS & no Sb/Br)	NIPDAU	Level-1-260C-UNLIM	0 to 70	CDC2536	Samples
CDC2536DBRG4	ACTIVE	SSOP	DB	28	2000	Green (RoHS & no Sb/Br)	NIPDAU	Level-1-260C-UNLIM	0 to 70	CDC2536	Samples

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) RoHS: TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (CI) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

- (3) MSL, Peak Temp. The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.
- (4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.
- (5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.
- (6) Lead/Ball Finish Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead/Ball Finish values may wrap to two lines if the finish value exceeds the maximum column width.

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PACKAGE OPTION ADDENDUM

6-Feb-2020

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PACKAGE MATERIALS INFORMATION

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TAPE AND REEL INFORMATION





	Dimension designed to accommodate the component width
В0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*All dimensions are nominal

Device	Package Type	Package Drawing			Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
CDC2536DBR	SSOP	DB	28	2000	330.0	16.4	8.1	10.4	2.5	12.0	16.0	Q1

www.ti.com 26-Feb-2019



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
CDC2536DBR	SSOP	DB	28	2000	350.0	350.0	43.0



SMALL OUTLINE PACKAGE



NOTES:

- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.

 2. This drawing is subject to change without notice.

 3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not
- exceed 0.15 mm per side.
- 4. This dimension does not include interlead flash. Interlead flash shall not exceed 0.25 mm per side.
- 5. Reference JEDEC registration MO-150.



SMALL OUTLINE PACKAGE



NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.

7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.



SMALL OUTLINE PACKAGE



NOTES: (continued)

- 8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
- 9. Board assembly site may have different recommendations for stencil design.



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